



RF accelerating structures, Lecture 10

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RF Accelerating Structures

Outline:

- 10. SRF Cavity Design Approaches for Linear Accelerators;**
- 11. Architecture of SRF accelerators;**
- 12. SRF around the world;**
- 13. APPENDIXES**

Chapter 10.

SRF Cavity Design Approaches for Linear Accelerators.

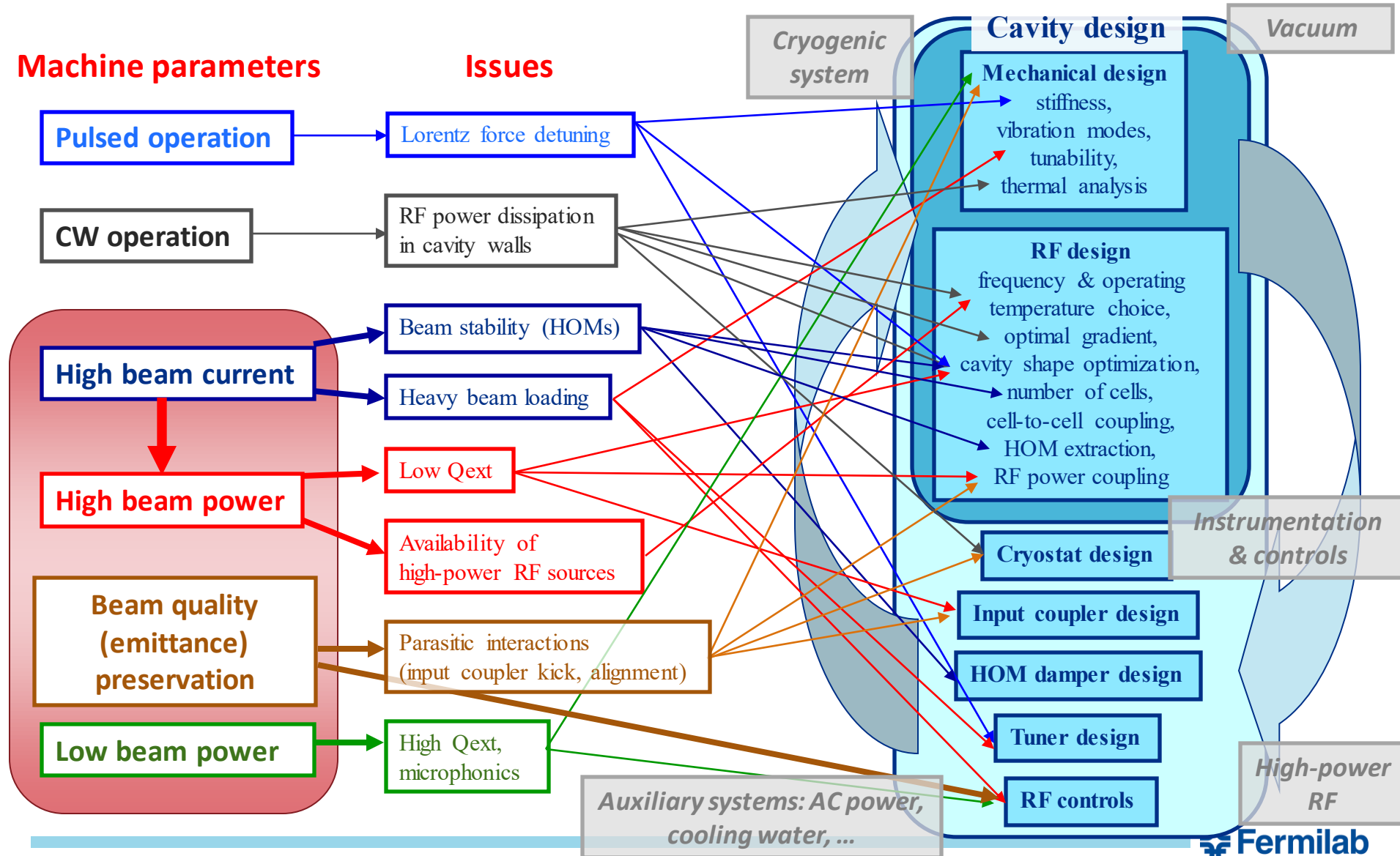
- a. The cavity RF design: approaches, tools;
- b. SRF cavity mechanical design;
- c. The cavity component design – RF couplers and tuners;
- d. Cryo-module design: issues and approaches.

SRF system design interdependences

Cryomodule design

Machine parameters

Issues



SRF cavities design issues:

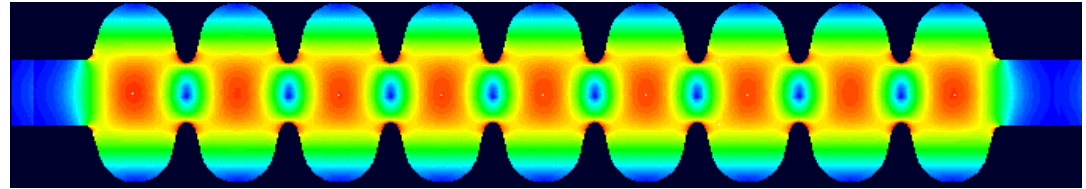
- Acceleration efficiency
 - max R/Q & min surface field enhancement factors (electric & magnetic)
- High Order Modes (HOMs) dumping
 - incoherent effect (loss factors, cryogenic losses)
 - coherent effects (emittance dilution, cryo-losses)
 - collective effects (transverse & longitudinal beam instabilities)
- Operation with small beam current
 - narrow cavity bandwidth & microphonics
- Field Emission
 - multipactor & dark current
- High Gradient pulsed operation
 - Lorentz force detuning
- Input Power Coupler
 - CW operation (min RF loss & static heat load)
- Beam Instrumentation
 - Cold Beam Position Monitor (low & high relativistic beam)

Tools for RF cavity simulations:

I. Field calculations:

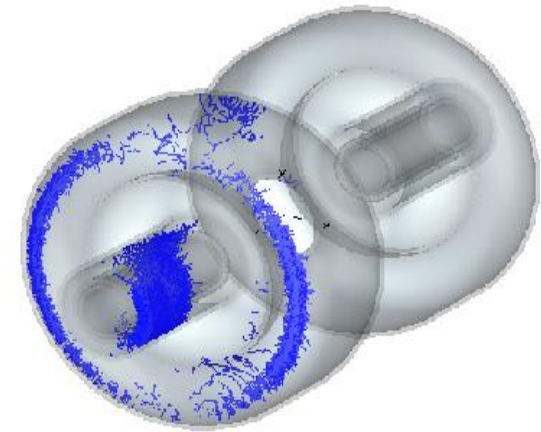
- Spectrum, (r/Q) , G , β (coupling)
- Field enhancement factors

- HFSS (3D);
- CST (3D);
- Omega-3P (3D);
- Analyst (3D);
- Superfish (2D)
- SLANS (2D, high precision of the field calculation).



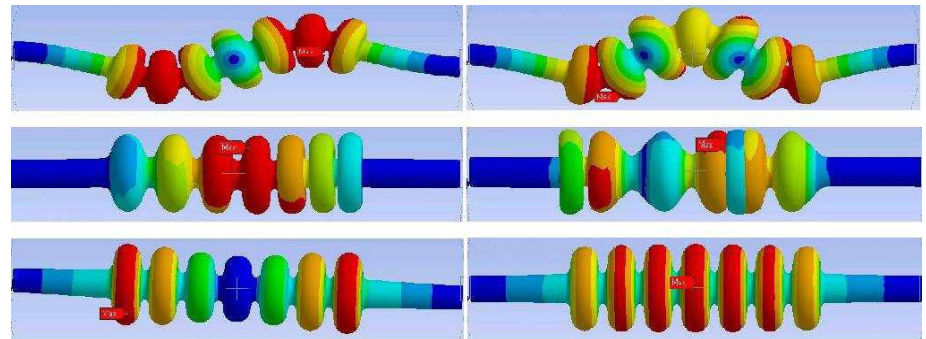
II. Multipactoring (2D, 3D)

- Analyst;
- CST (3D);
- Omega-3P



III. Wakefield simulations (2D, 3D):

- GdfidL;
- PBCI;
- ECHO.



IV. Mechanical simulations:

Lorenz force and Lorenz factor,
Vibrations,
Thermal deformations.

a. ANSYS

Cavity Simulation Workflow

- 1. Creation of the project 3D model**
 - drawing in the CST GUI (takes time, full-parametrization, easy modification)
 - geometry import from 3rd parties CADs (quick, need special license, limited parametrization, potential mesh problem)
- 2. Choosing a proper solver**
 - depends on the problem, available hardware, simulation time ...
- 3. Setting boundary conditions**
 - frequency, symmetries, ports, materials, beam excitation, temperature, ...
- 4. Checking the mesh quality**
 - generate and visualize the mesh, set initial mesh size, create sub-volumes and modify models if needed, mesh fine-tuning (curvature order, surface approximation)
- 5. Solver fine-tuning**
 - direct or iterative, parallelization, special settings, ...
- 6. Running first simulation**
 - check the results, set postprocessing steps, tune & modify the mesh, ...
- 7. Setting optimization**
 - set parameters sweep, define the goal function, simplify the model

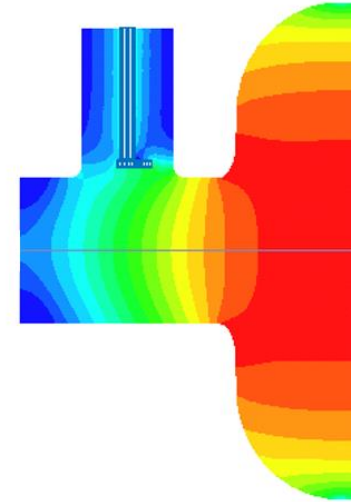
Design approaches

- Aperture choice:

- Smaller aperture → smaller field enhancement factors, higher R/Q;

- Limitations:

- beam losses,
- field flatness,
- mechanical stability,
- surface processing,
- Q_{load} (coupling to the main coupler)
- HOMs (trapped modes)
- Wakes (electron accelerators)



SNS (805 MHz): $2a=86\text{mm}$ ($\beta=0.61$), $2a/\lambda = 0.23$

$2a=98\text{mm}$ ($\beta=0.81$), $2a/\lambda = 0.26$

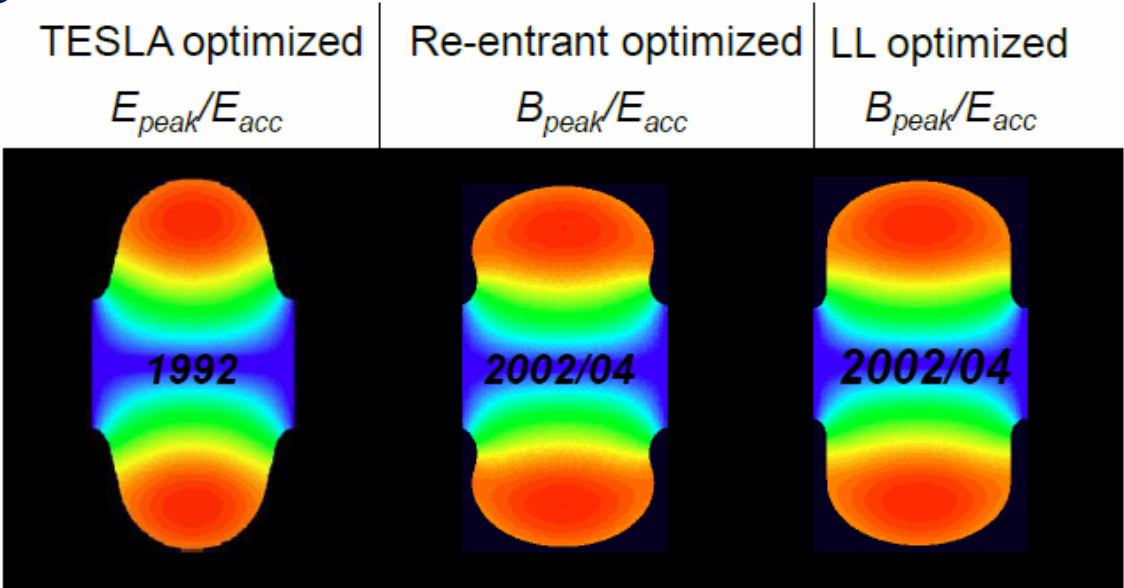
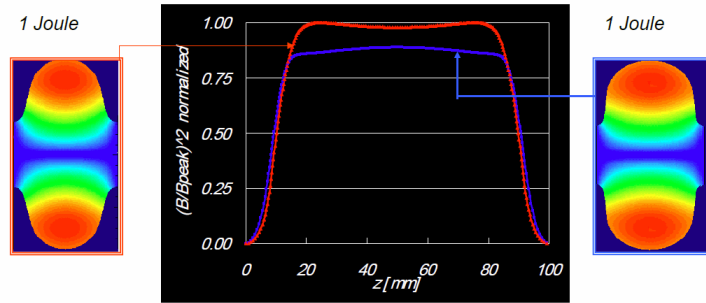
HIPPI (704 MHz): $2a=80\text{mm}$ ($\beta=0.47$), $2a/\lambda = 0.19$

PIP II (650 MHz): $2a=83\text{mm}$ ($\beta=0.61$), $2a/\lambda = 0.18$

$2a=100\text{ mm}$ ($\beta=0.9$), $2a/\lambda = 0.22$

Design approaches

Cavity shape:



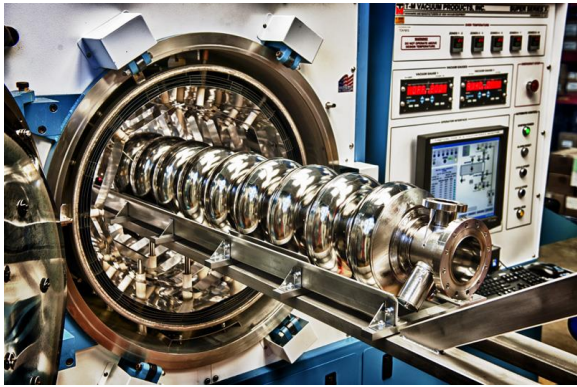
		TESLA optimized	Re-entrant optimized	LL optimized
r_i	[mm]	35	30	30
k_{cc}	[%]	1.9	1.56	1.52
E_{peak}/E_{acc}	-	1.98	2.30	2.36
B_{peak}/E_{acc}	[mT/(MV/m)]	4.15	3.57	3.61
R/Q	[Ω]	113.8	135	133.7
G	[Ω]	271	284.3	284
$R/Q \cdot G$	[$\Omega \cdot \Omega$]	30840	38380	37970
$k_{\perp} (\sigma_z=1mm)$	[V/pC/cm ²]	0.23	0.38	0.38
$k_{\parallel} (\sigma_z=1mm)$	[V/pC]	1.46	1.75	1.72

Electromagnetic optimization

The Scope of EM-Mechanical Design (a cavity +He vessel)

- Minimize a sensitivity to microphonics due to He pressure fluctuations (df/dP) and mechanical vibrations
- Minimize a Lorentz Force Detuning (LFD) coefficient
- To keep the stiffness and tuning sensitivity at suitable level to allow for tuning.
- Keep provision for slow and fast tuner integration.
- Enough strength to withstand atmospheric pressure
- “Dressed” cavity has to be qualified in 5 different load conditions by stress analysis
 1. Warm Pressurization
 2. Cold operation at maximum pressure
 3. Cool down and tuner extension
 4. Cold operation at maximum pressure and LHe weight
 5. Upset condition – Insulating and beam vacuum failure

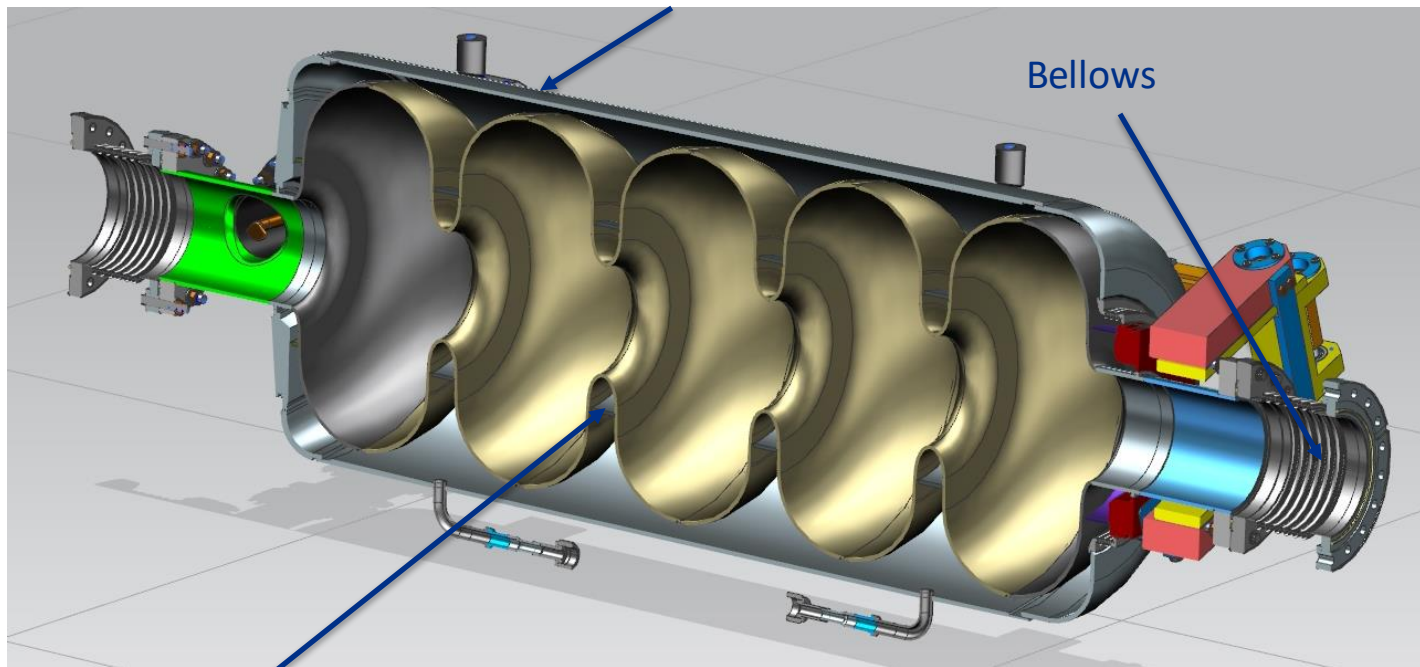
Design approaches. Mechanical stiffness



Bare cavity

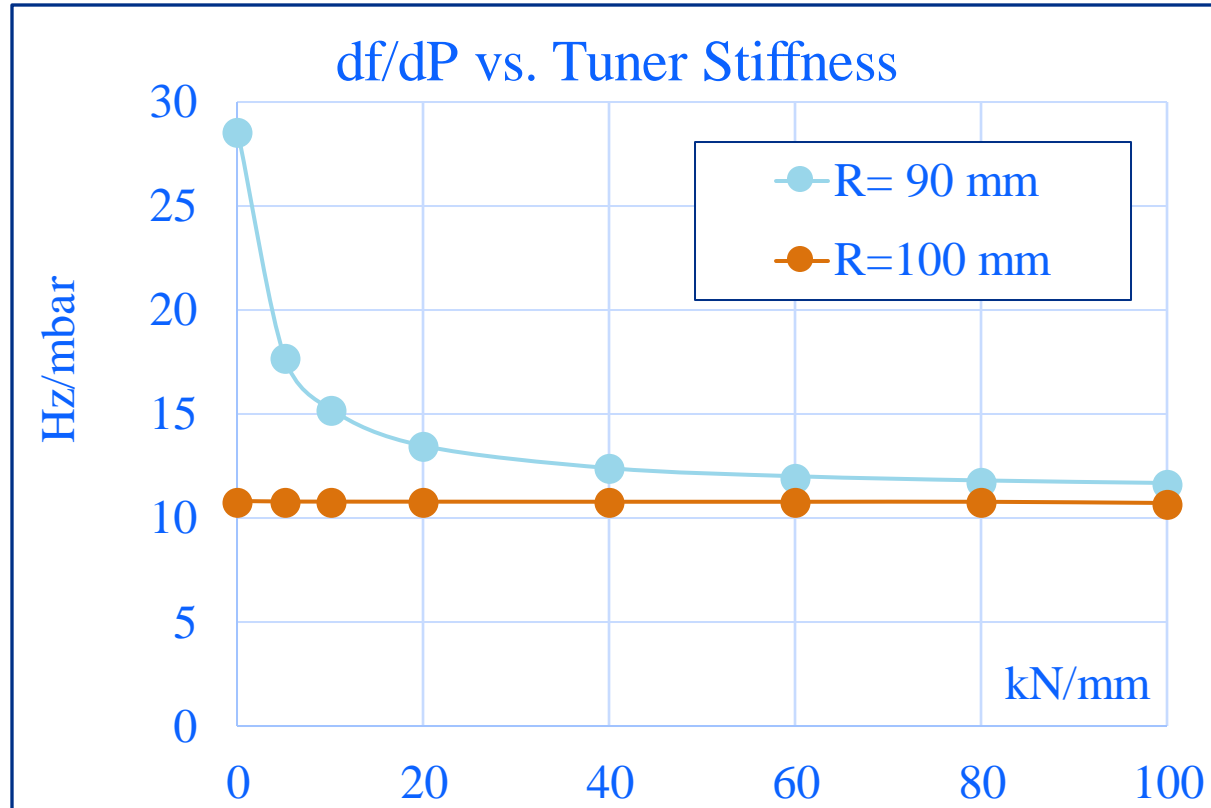


Dressed cavity



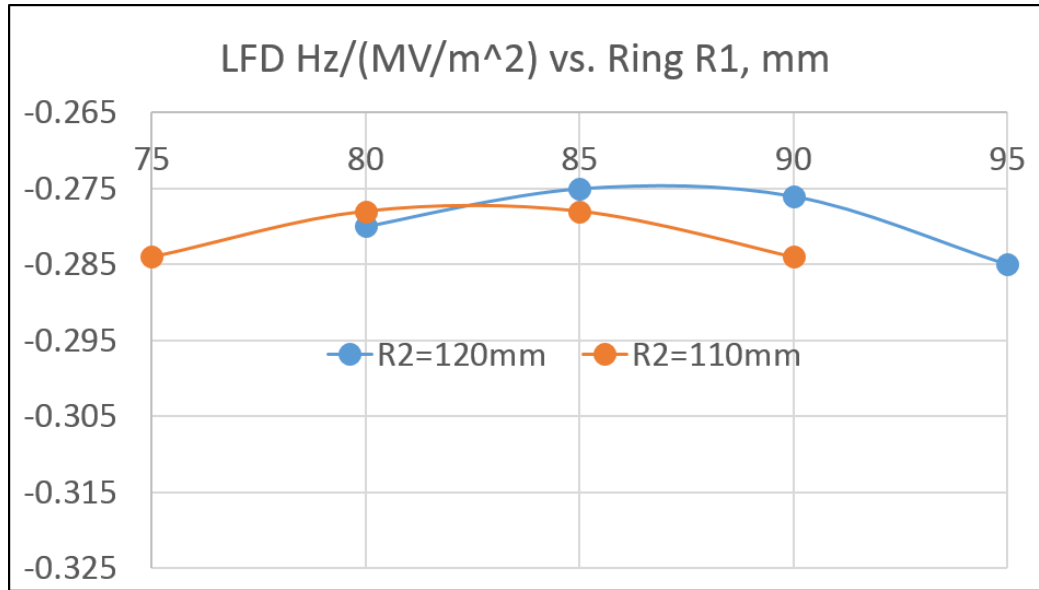
Stiffening rings

Design approaches, df/dP optimization

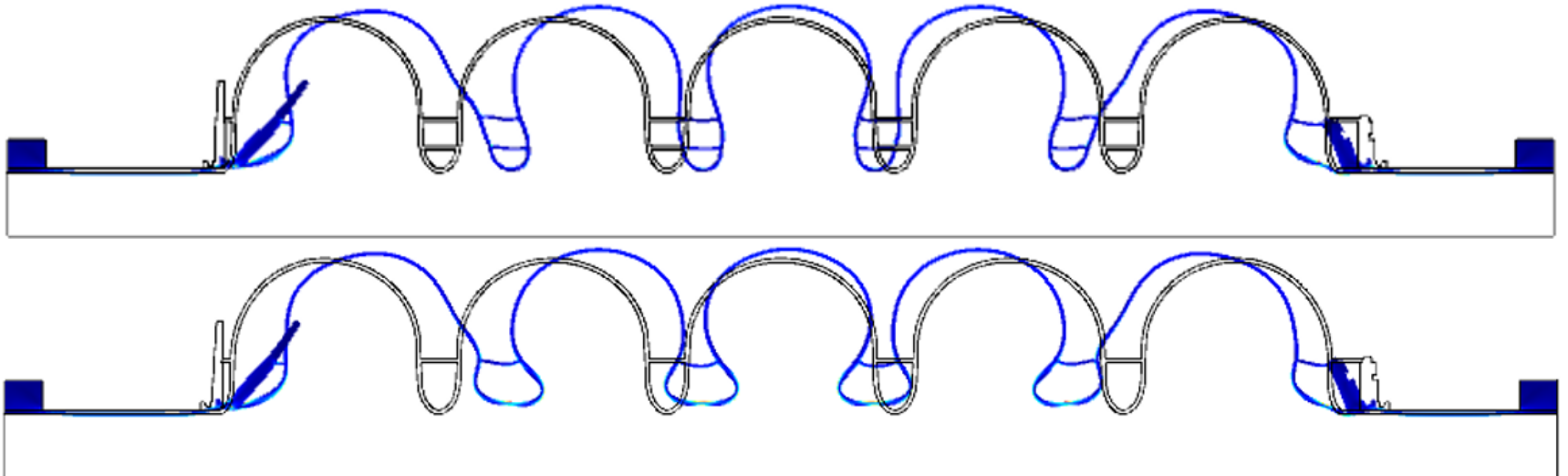


df/dP for stiffening ring $R = 90$ mm vs. 100 mm
Bellows radius of $OD=125$ mm

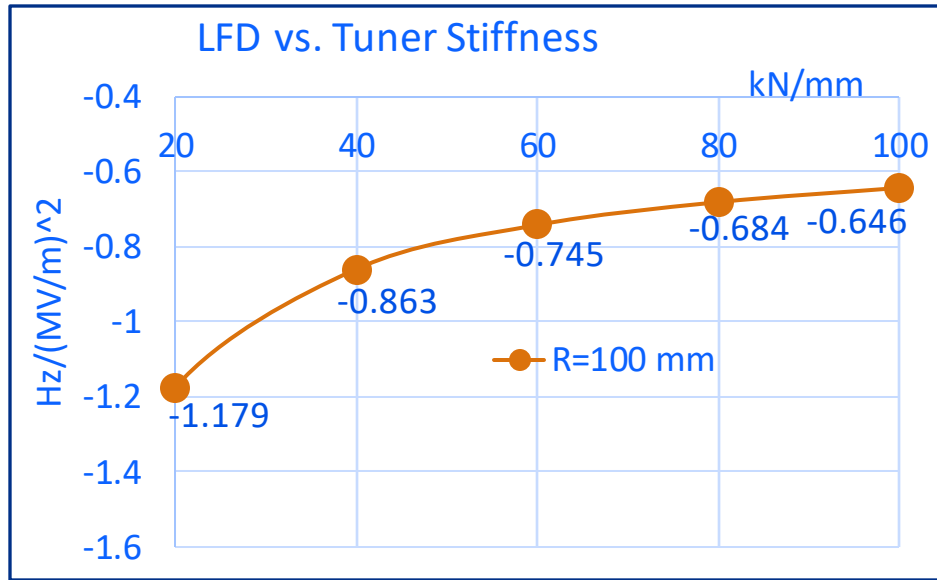
Design approaches, LFD minimization.



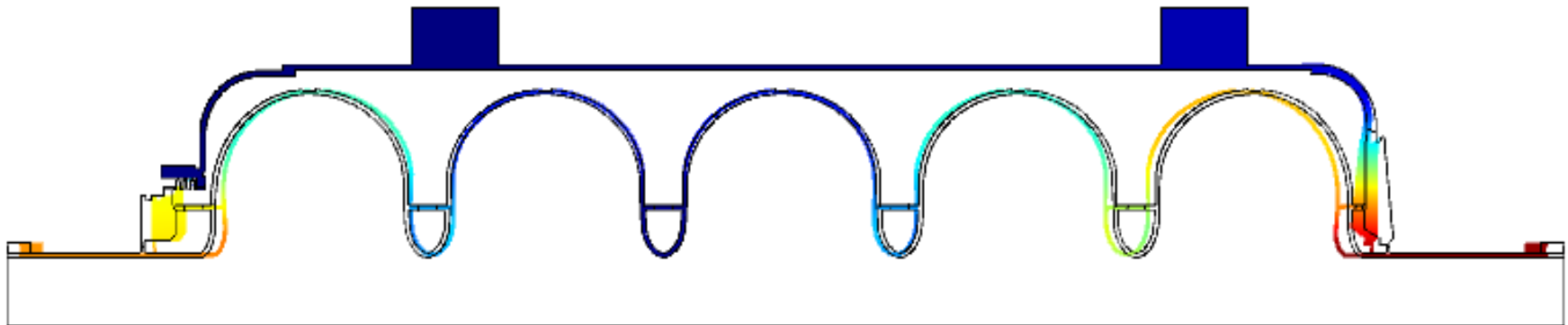
For 2 rings LFD~-0.275 Hz/(MV/m)²
For 1 ring LFD~-0.38 Hz/(MV/m)².



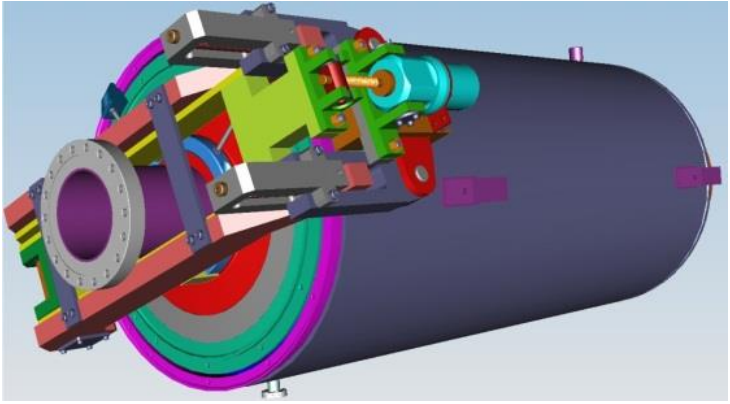
Design approaches, LFD optimization



LFD for dressed cavity with R=100 mm stiffening rings



Design approaches, an elliptical cavity tuner options

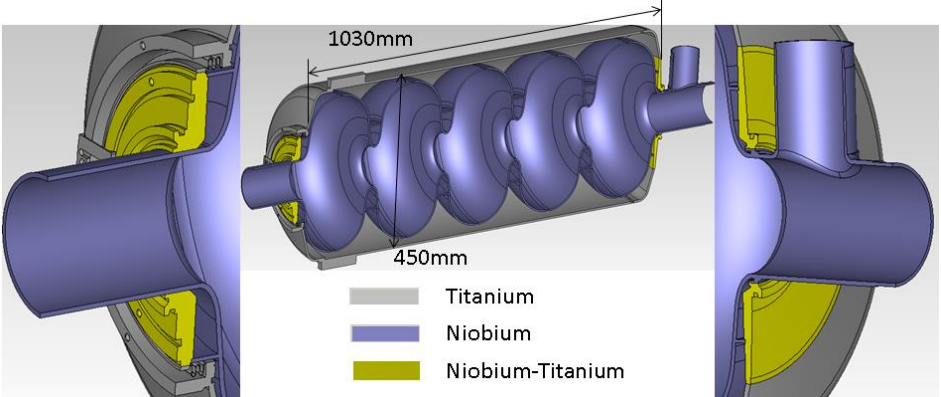


Blade Tuner – scaled ILC:

- High df/dP ,
- Insufficient tuning efficiency;

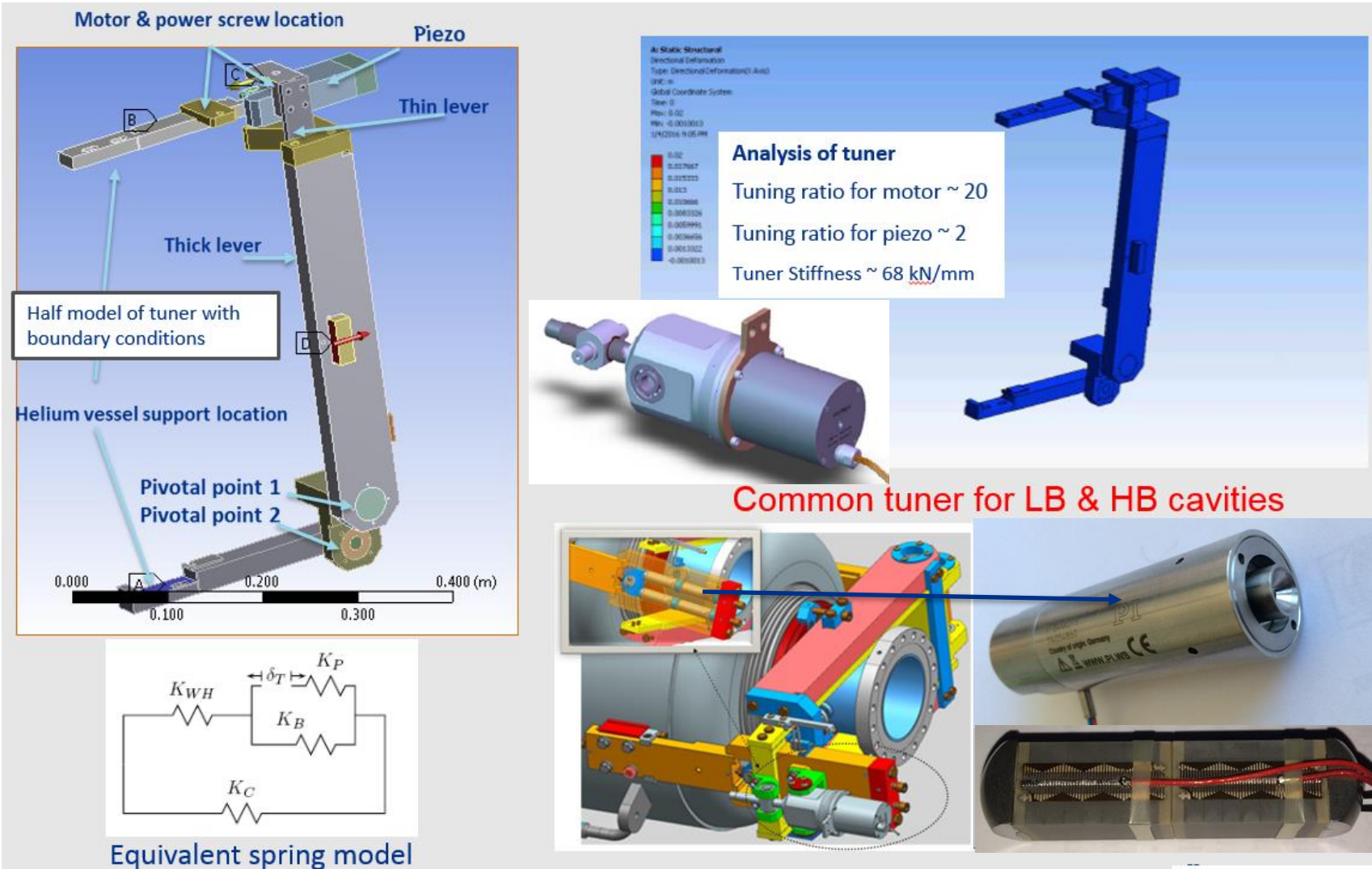
Lever Tuner design:

- Low df/dP ,
- Mechanical resonances > 60 Hz;
- Good tunability;
- Less expensive.



Design approaches, the tuner

- Coarse tuner (step motor) – for the cavity tuning after cooldown, range ~ 100 kHz
- Fine tuner (piezo) – for microphonics and LFD compensation, range ~ 1 kHz

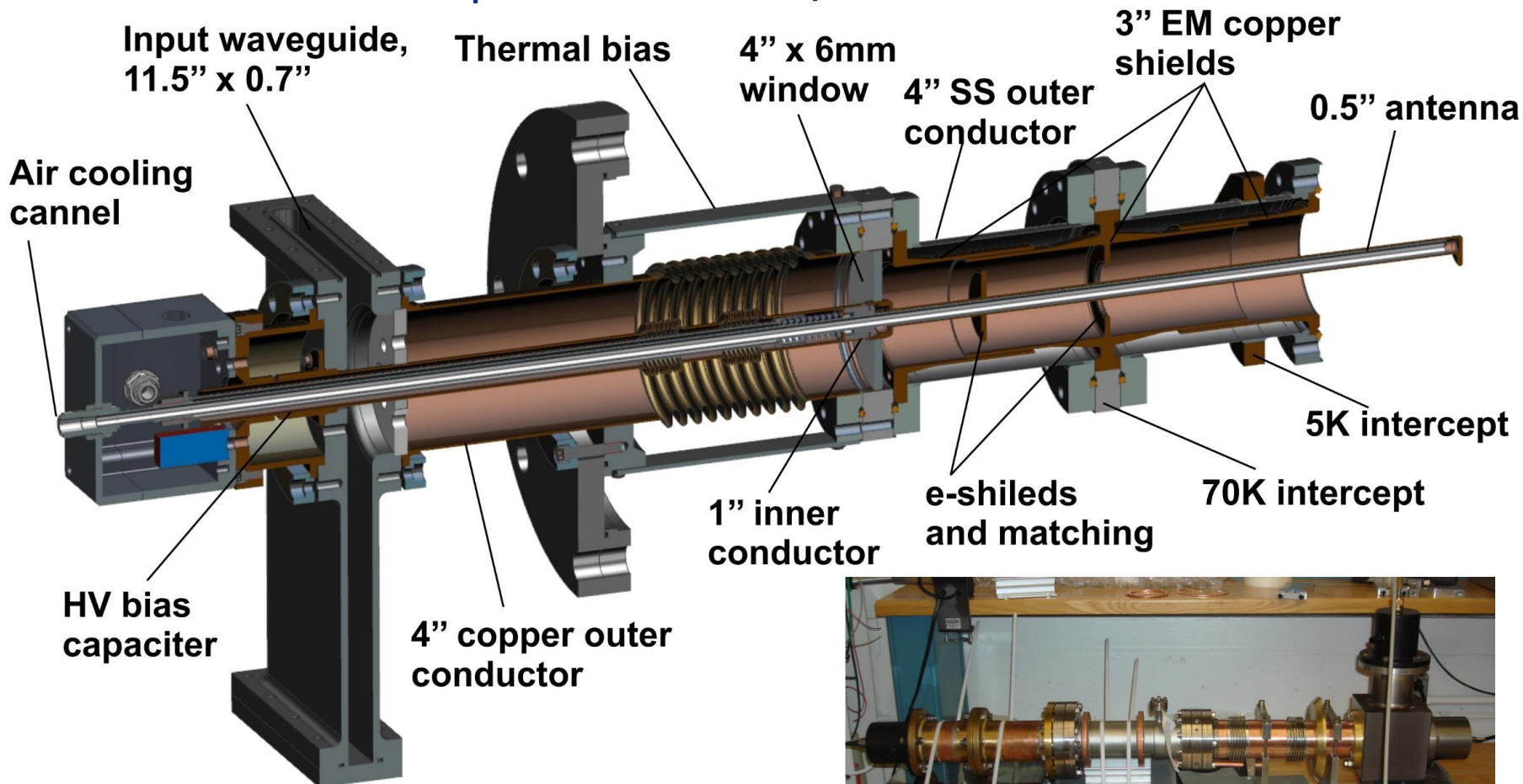


Couplers for SRF cavities:

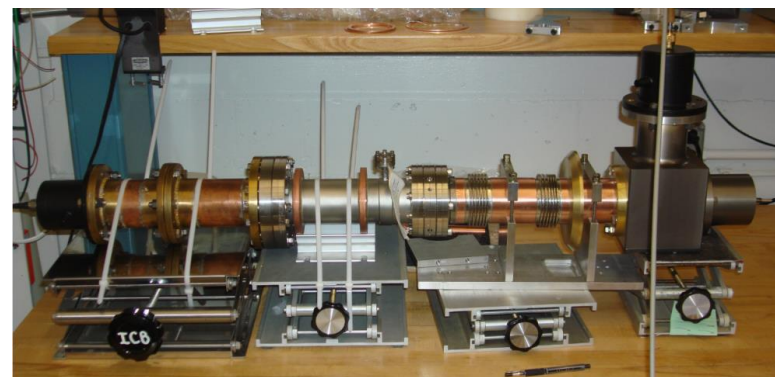
- The couplers should transfer RF power to the cavity operating at cryo temperature
 - High pulsed power (hundreds of kW for pulsed operation - SNS, ESS) → good electric strength
 - High average power (up to 100 kW in CW – PIP II) → low dynamic losses, sufficient cooling (air, He)
 - Low static losses
 - Good vacuum properties
- Issues:
 - MP in the vacuum part → DC bias
 - Copper plating (chips)
- Design choices:
 - One window versus two windows
 - Type of cooling

Couplers for SRF cavities:

Coupler for PIP II LB/HB 650

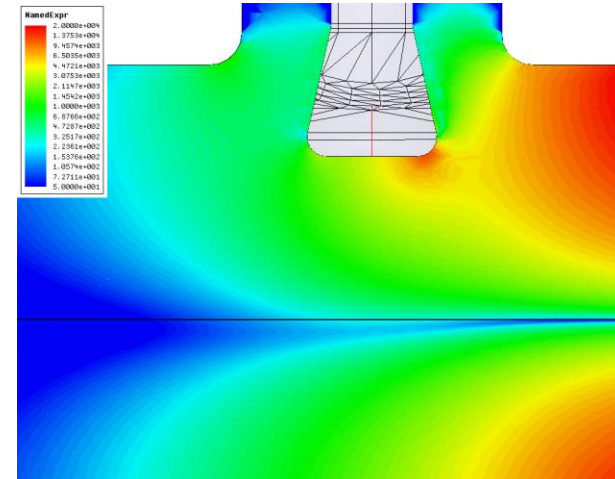
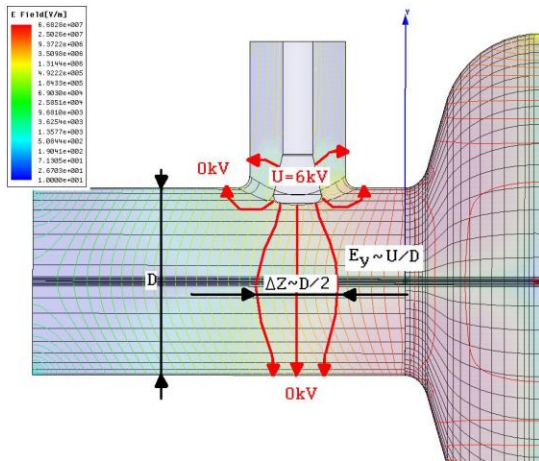


Average power up to 100 kW



RF kick caused by the input and HOM couplers.

Simple estimations of the transverse fields caused by the main coupler:



RF voltage:

$U = (2PZ)^{1/2}$, Z —coax impedance;

for $P=300$ kW and $Z \approx 70$ Ohms

$U \approx 6$ kV

Transverse kick caused by the couplers acts on a bunch the same direction for all the RF cavities of the linac.

Real part may be compensated by the linac feedback system;

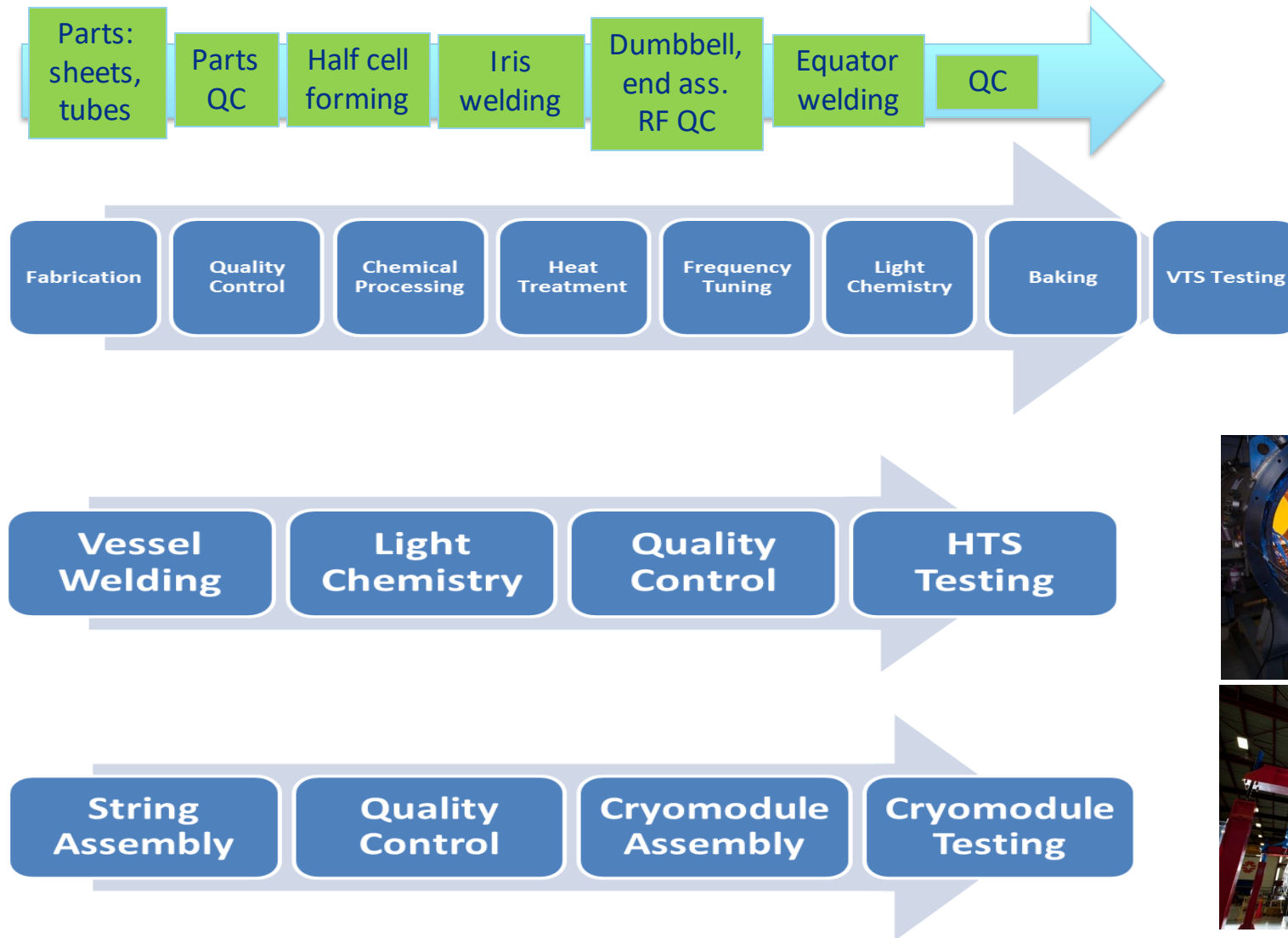
Imaginary part gives the beam emittance dilution (here β is beta-function, σ is the bunch length, and U_0 is the initial beam energy):

$$\gamma \mathcal{E} \approx \gamma(z_{\max}) y_{\max} y'_{\max} = \frac{\pi^2 v^2 E^2 \sigma^2 \beta^3 \gamma_0}{\lambda_{RF}^2 U_0^2}$$

Transverse kick:

$$v = \frac{\Delta p_y c}{\Delta U_{acc}} \approx \frac{U}{2U_{acc}} = \frac{6kV}{2 \times 30MV} = 10^{-4}$$

SRF cavity production technology



Material quality control

Typical Technical Specification to Niobium Sheets (For XFEL Cavities)

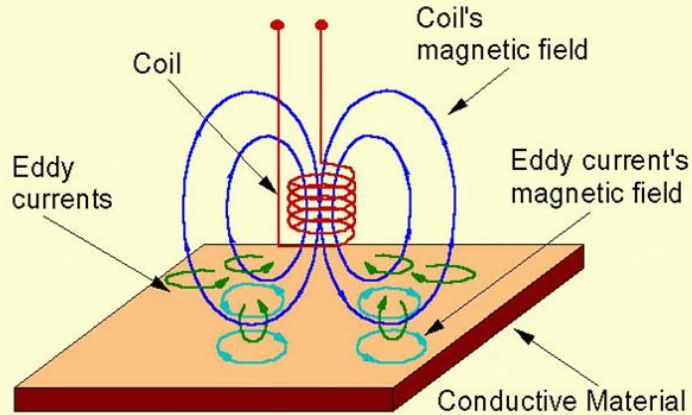
Concentration of impurities in ppm*				Mechanical properties	
Ta	≤ 500	H	≤ 2	RRR**	≥ 300
W	≤ 70	N	≤ 10	Grain size	≈ 50 μm
Ti	≤ 50	O	≤ 10	Yield strength, $\sigma_{0.2}$	50 < $\sigma_{0.2}$ < 100 N/mm² (Mpa)
Fe	≤ 30	C	≤ 10	Tensile strength	> 100 N/mm² (Mpa)
Mo	≤ 50			Elongation at break	30 %
Ni	≤ 30			Vickers hardness HV 10	≤ 60

*ppm=Parts per Million

**The Residual Resistance Ratio “RRR” is the ratio of resistance at 300K (room temperature) to the resistance at 10K.

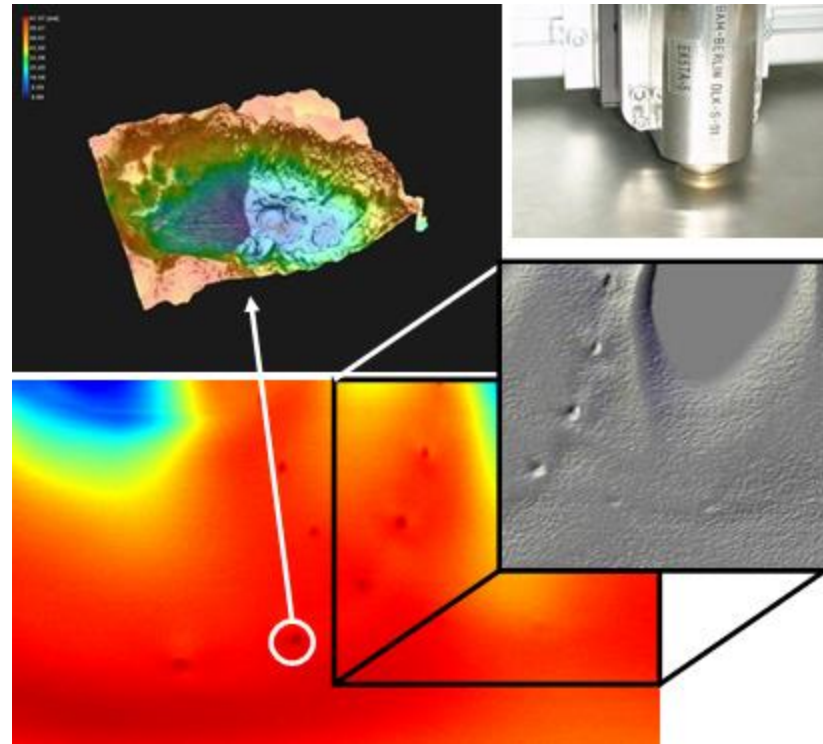
Material quality control

Eddy current scanning



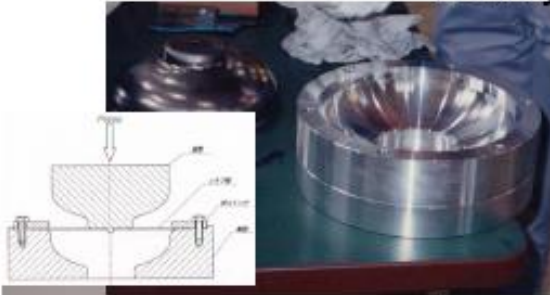
Disks are cut from high purity niobium sheet and eddy current scanned for pits, scratches or inclusions of foreign materials

Discs with inclusions of foreign materials or damage are rejected



Elliptical cavity production

Fabrication: Conventional fabrication (deep drawing and EB welding of fine grain Nb). Experiences of ca. 20 years of industrial cavity fabrication are available



Half cells are produced by deep drawing.

Dumb bells are formed by electron beam welding.



After proper cleaning eight dumb bells and two end group sections welded by electron beam together

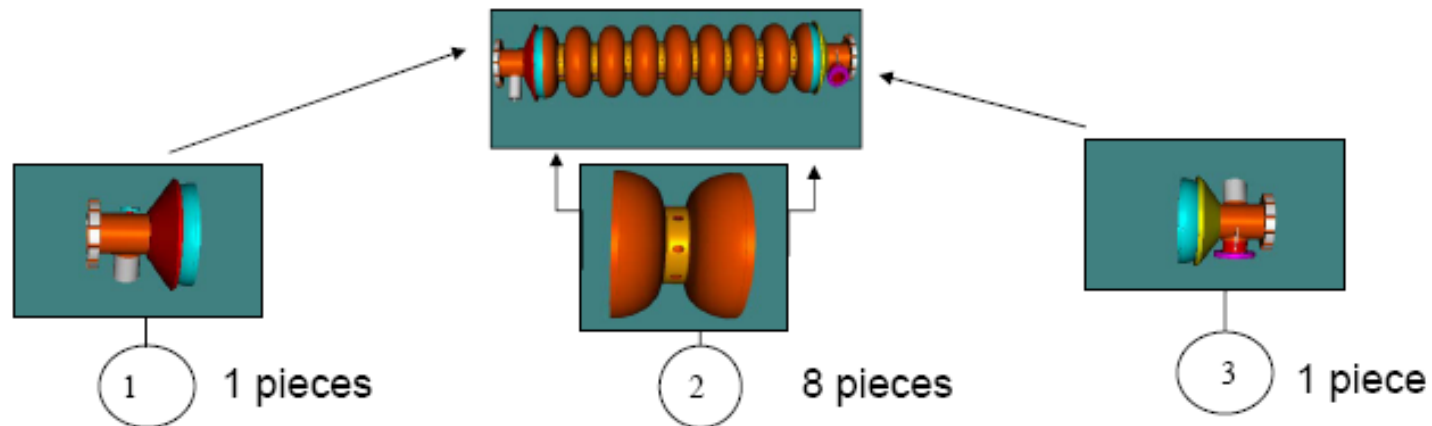


Important: clean conditions on all steps
shape accuracy, preparation and EB welding

Elliptical cavity production

Cavity welding: the general way

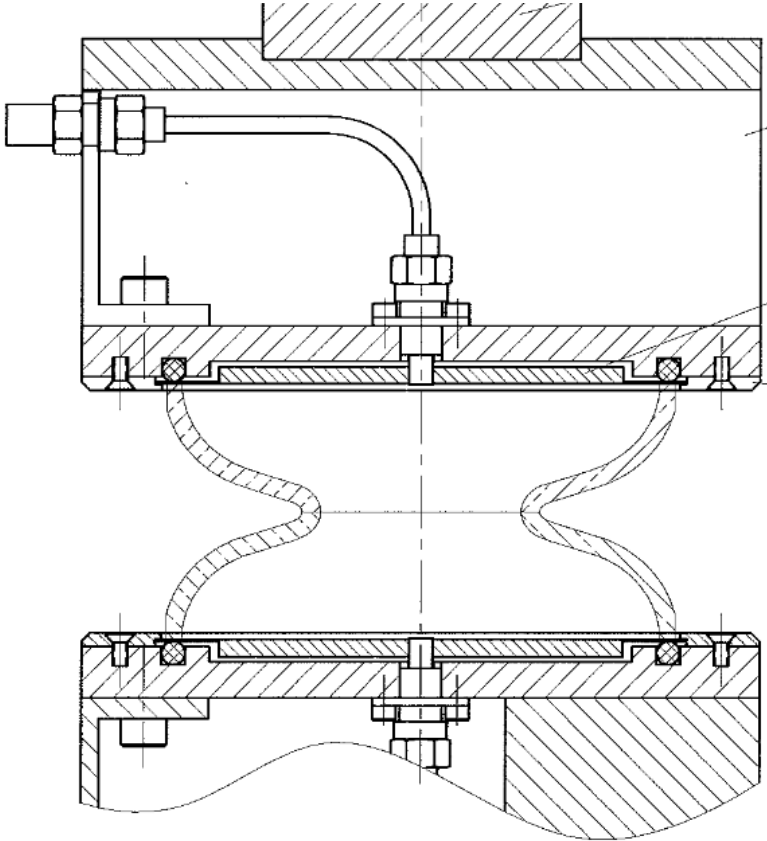
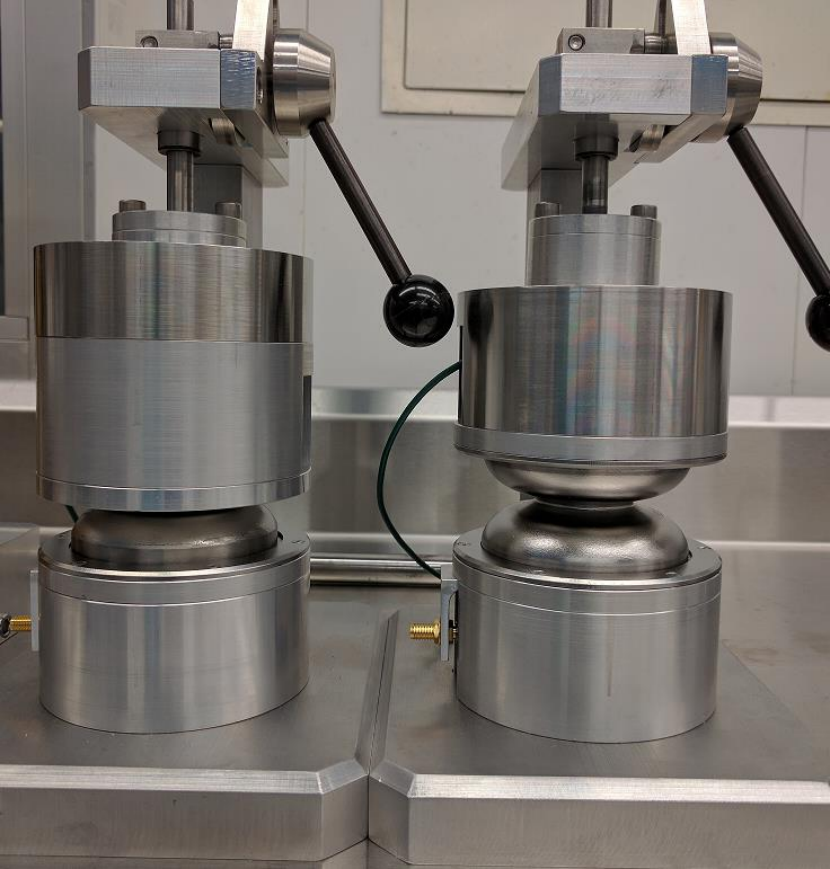
There are differences of welding processes in industry



1. Degreasing and rinsing of parts
2. Drying under clean condition
3. Chemical etching at the welding area (Equator)
4. Careful and intensive rinsing with ultra pure water
5. Dry under clean conditions
6. Install parts to fixture under clean conditions
7. Install parts into electron beam (eb) welding chamber
(no contamination on the weld area allowed)
8. Install vacuum in the eb welding chamber $\leq 1E-5$ mbar
9. Welding and cool down of Nb to $T < 60$ C before venting
10. Leak check of weld



Elliptical cavity production



MEASURING OF DUMBELL

3.9 GHz half cells and dumbbell measurement fixture

Elliptical cavity production

Cavity production steps:

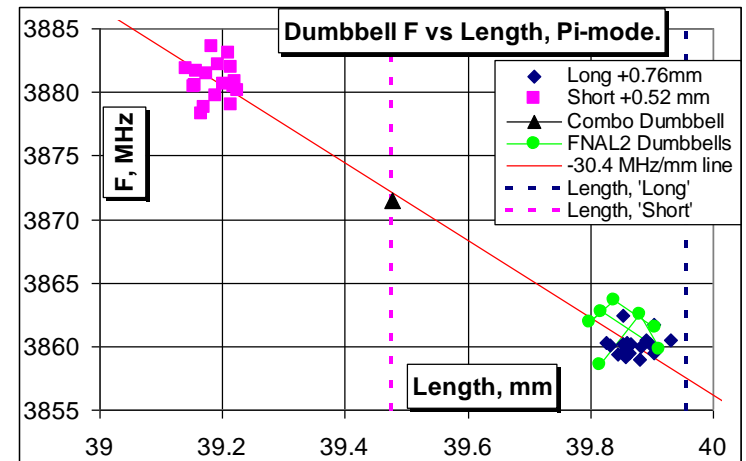
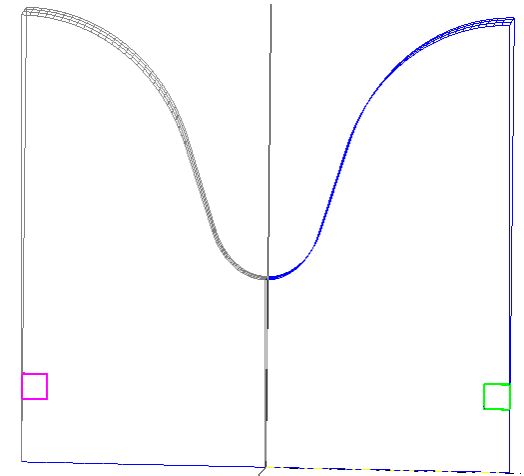
- Eddy current scanning of Nb shits.
- Cut disk blanks with hole in the center
- Flow forming of half-cell and trimming iris and equator area with extra length for tuning and welding shrinkage compensation. No extra length for a tuning in mid-cells. If pass visual inspection:
- Frequency and length measurements. Sensitivity of the frequency to extra length is 14 MHz/mm at iris and -55 MHz/mm at equator.
- EB welding of two half-cell at iris to form dumbbell. Partial penetration welding from both sides. If pass visual inspection:
- Frequency and length measurements of the dumbbells. Both mode frequencies F_0 and F_{pi} measured 3 times: 1) without perturbation F_0 and F_1 , 2) with perturbation in 1st half cell F_{01} and F_{11} 3) with perturbation in 2nd half cell F_{02} and F_{12} . Difference of the frequencies of two half cell can be calculated from these data:

$$dF = F_2 - F_1 = (F_{01} - F_{11} + F_{12} - F_{02}) / (F_{01} + F_{11} - F_{02} - F_{12}) * k * F_0$$

Where $k \sim 4(F_{pi} - F_0) / (F_{pi} + F_0)$, for a 3rd harmonic cavity $k \sim 0.08$ MHz

-Trimming calculations:

- Equator trimming
- Equator welding
- Mechanical and RF QC of the new cavity.
- Bulk BCP and 800C baking,
- RF tuning of the cavity



Multi-cell cavity field flatness tuning

FNAL elliptical 9 cell cavity tuning procedure. This technique based on bead-pull measurements of field distribution on operating (π -mode). Amplitudes of E-field in the center of each cell used for frequency of individual cells.

Normalized field distribution is uniform, $A_i=1$ for $i=1,2, \dots 8, 9$, if frequency of each cell are same. When frequency of the cell #n is shifted by dF_n field distribution will change by dA_i .

$$dA_i = K_{in} * dF_n$$

Perturbation of frequency of each will change field distribution:

Let us solve this equation to find frequency perturbation from field distribution:

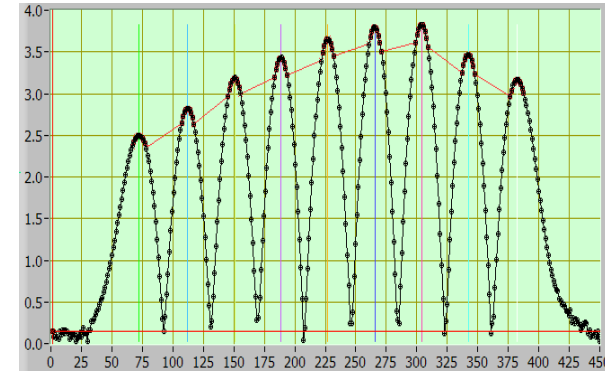
$$dA = K * dF \Rightarrow K^{-1} * dA = K^{-1} K * dF = dF$$

Where sensitivity coefficients matrix K calculated from HFSS simulations.

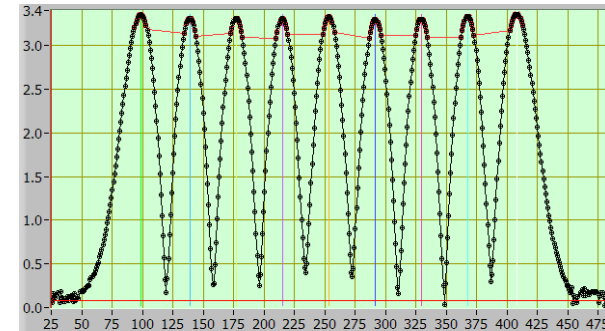
During RF tuning of the cavity, we need to tune its operating mode frequency F_0 . Also, we can not measure individual cell frequency but can measure F_0 . Tuning of cell #n by dF_n shifts also cavity frequency by $dF_0 \sim dF_n / 9$. If design frequency is F_0 tuning of the cell should be done by shifting operating mode frequency by:

$$dF_0 = (F_0 - F_0 - dF_n) / 9$$

This technique works best when field flatness of the cavity is close to ideal. Because it linear and based on small perturbations. Tuning is better to start with most perturbed cell. If field flatness still not acceptable the additional tuning cycle should be done.



Before tuning. FF 65%, slope +28 %



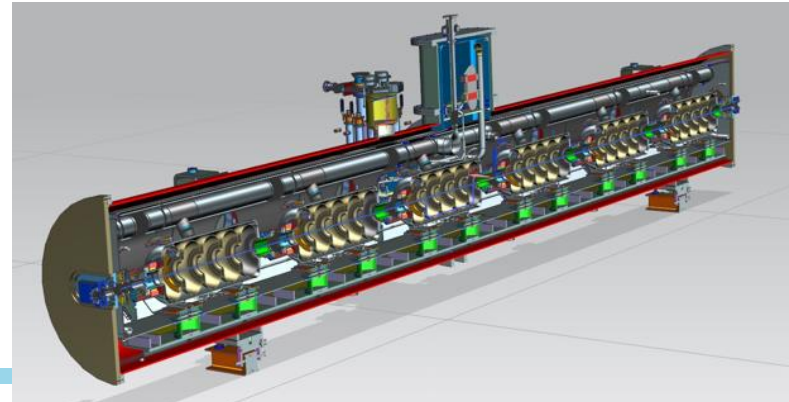
After tuning. FF 98%, slope +0.64 %

ICL Cavity Tuning Machine

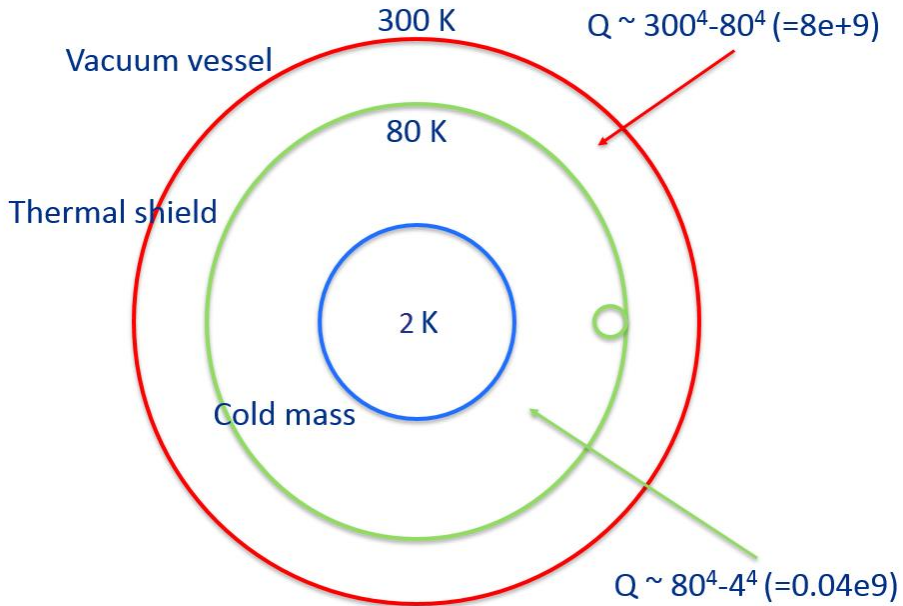
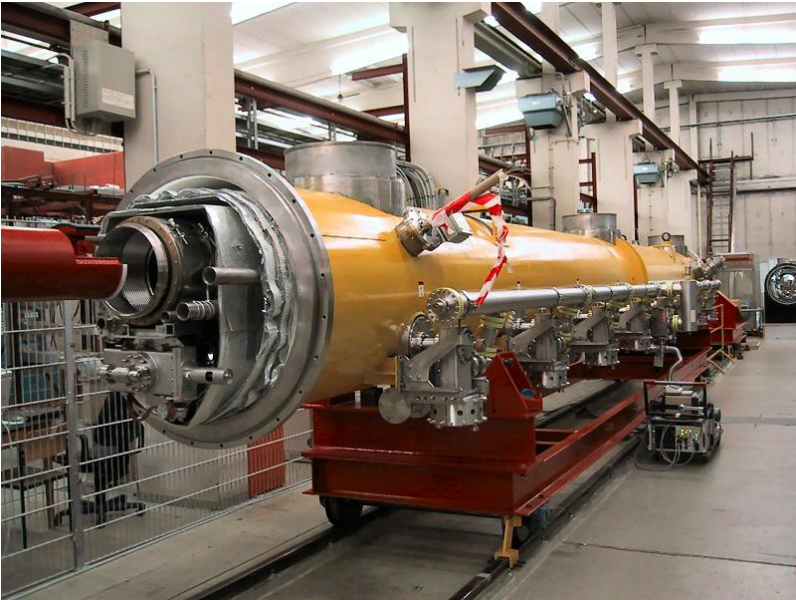


Vacuum vessels

- The outermost cryostat component that:
 - Contains the insulating vacuum.
 - Serves as the major structural element to which all other systems are attached to the accelerator tunnel floor.
 - Serves as a pressure containment vessel in the event of a failure in an internal cryogen line.
- The design for internal and external pressure are addressed by the ASME Boiler and Pressure Vessel Code, Section VIII, Divisions 1 and 2 and specific workplace codes.
- Insulating vacuum is generally in the $1e10^{-6}$ torr range, but can be as low as $1e10^{-4}$. The lower the better.



Vacuum vessel and thermal shields

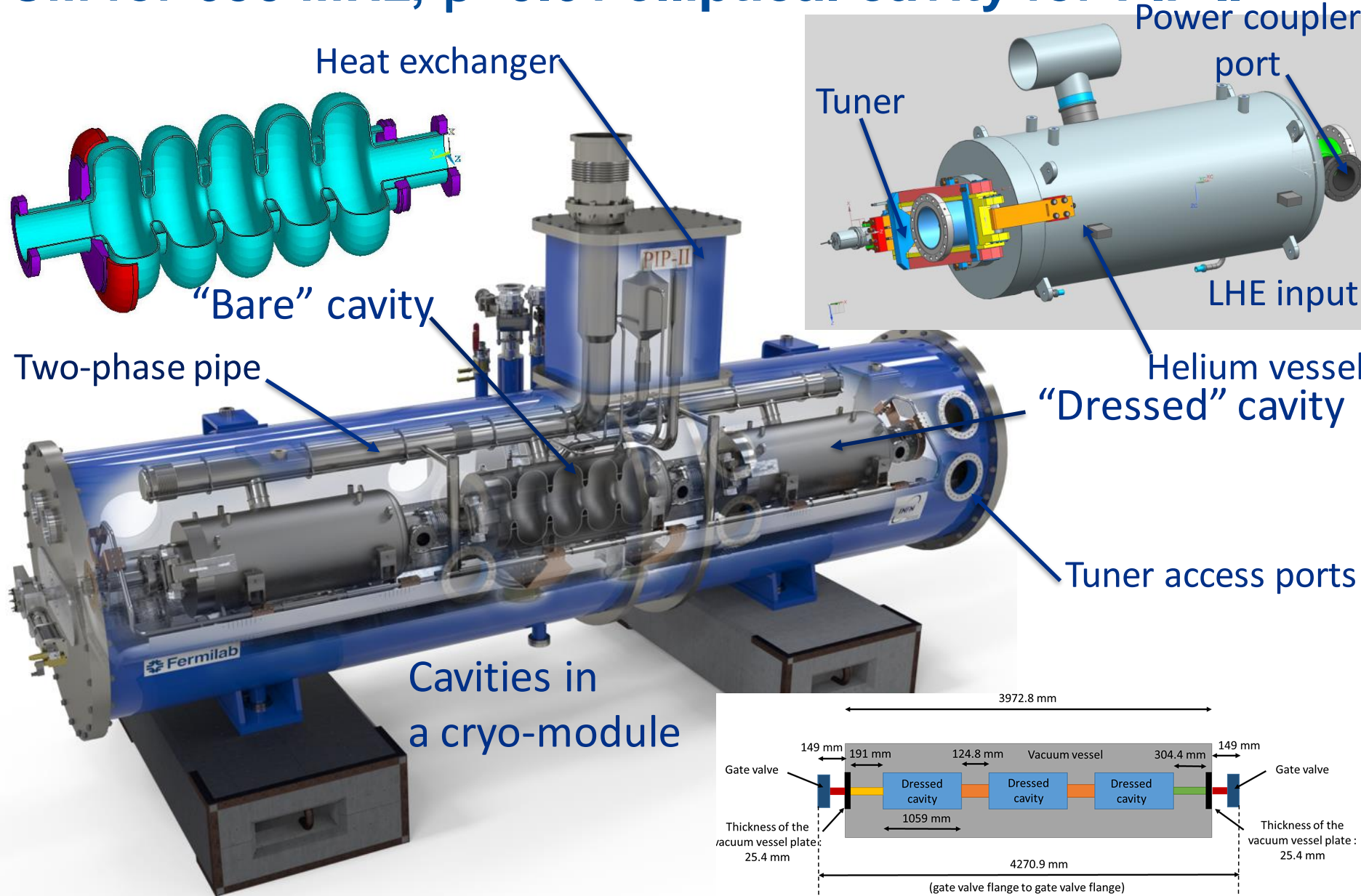


Insulation

- Multi-layer insulation (MLI) reflects radiation heat transfer back toward its source.
- Usually mounted on the outside of the colder surface, e.g. the thermal shield or cold mass.
- Consists of alternating layers of reflector and spacer material:
 - Reflector is usually double-aluminized mylar sheets 6-12 μm thick aluminum-coated on both sides with a minimum of 300 \AA .
 - Spacer is usually a polyester net, fiberglass net or other similar material compatible with the environment.
 - The reflector can be perforated to facilitate pumpout.
- The number of layers varies but is usually from 30-60 layers on a thermal shield nominally at 80 K and 10-15 layers on a lower temperature shield or cold mass.
- It must be in vacuum – $1\text{e}10^{-4}$ torr or lower.
- To estimate the total heat load due to radiation and residual gas conduction, realistic values are $\sim 1.5 \text{ W/m}^2$ at 80 K and $\sim 0.15 \text{ W/m}^2$ at 4.5 K.



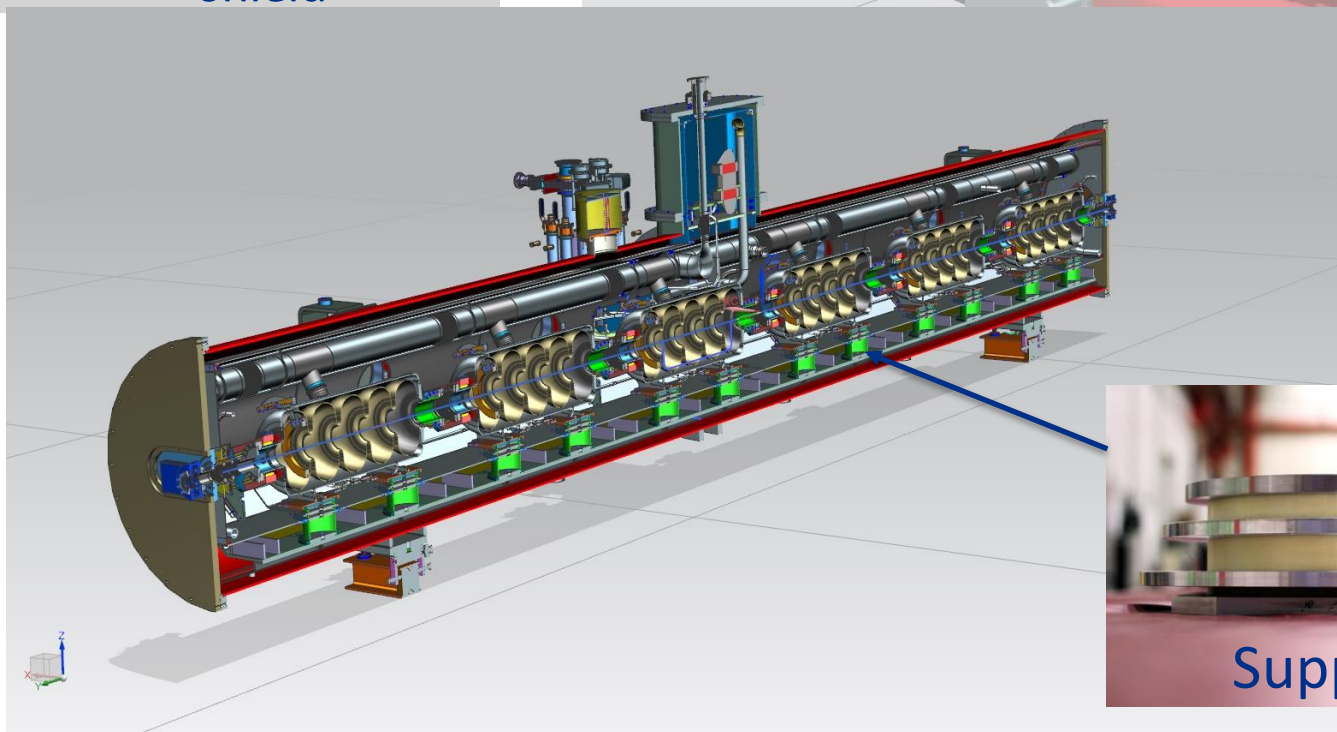
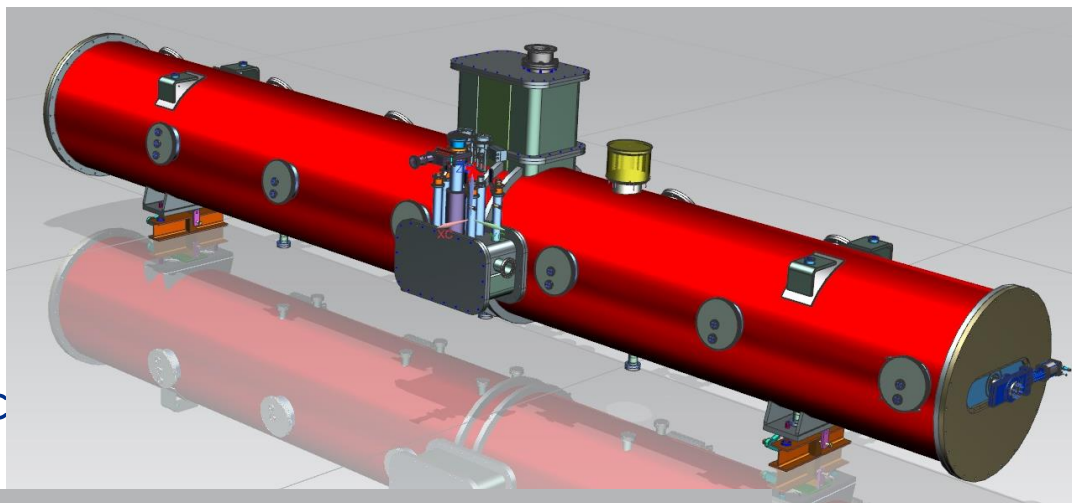
CM for 650 MHz, $\beta=0.61$ elliptical cavity for PIP II



CM for 650 MHz, $\beta=0.9$ elliptical cavity for PIP II

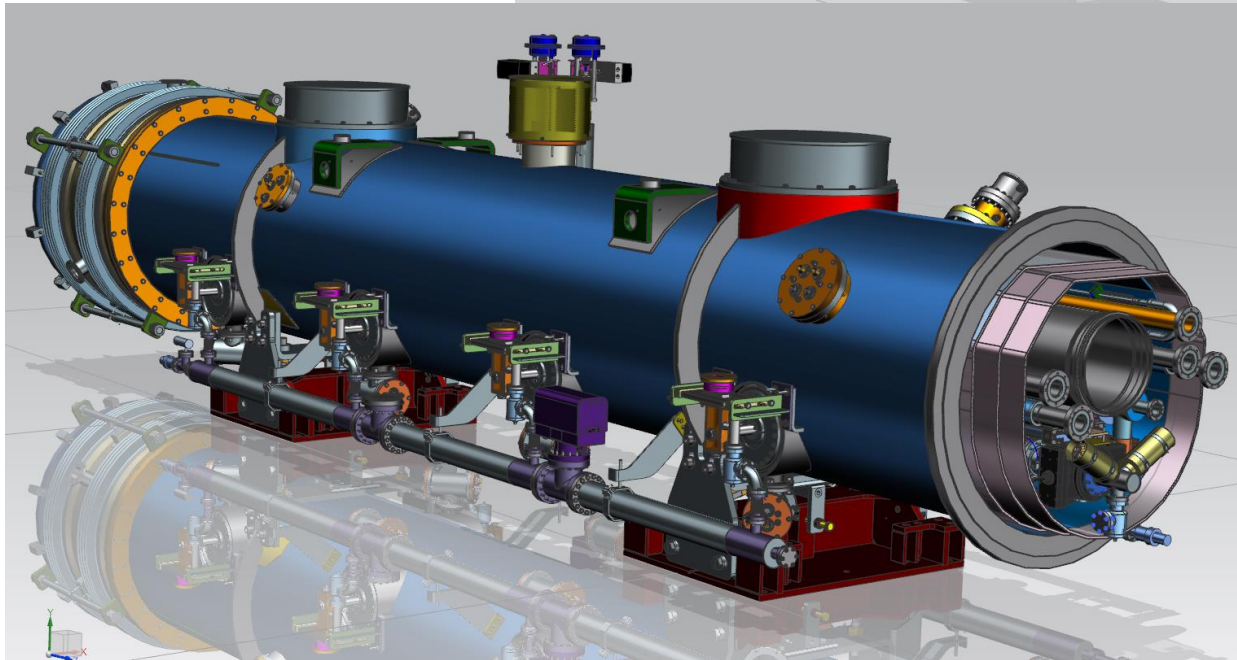
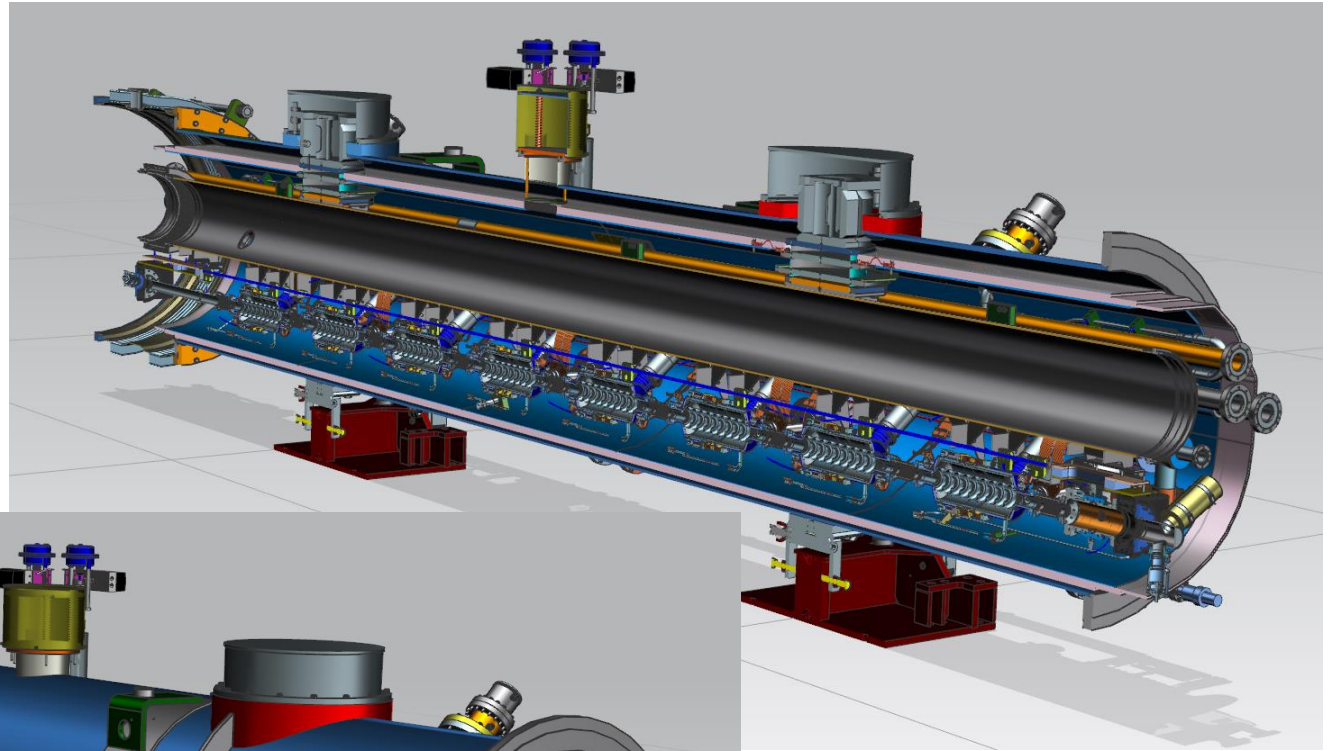
Tuner

Internal magnetic shield



Supports

CM for 3.9 GHz elliptical cavity for LCLS II



Cavities supported by High-power couplers

Elliptical Cryomodules – CEA/IPNO

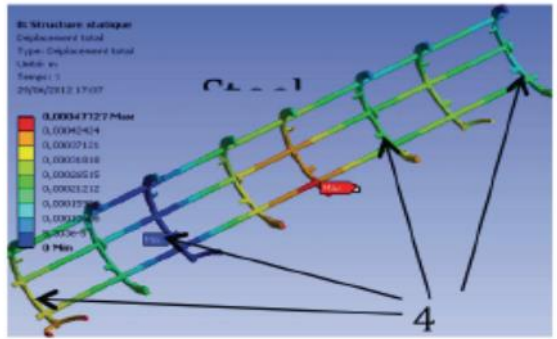
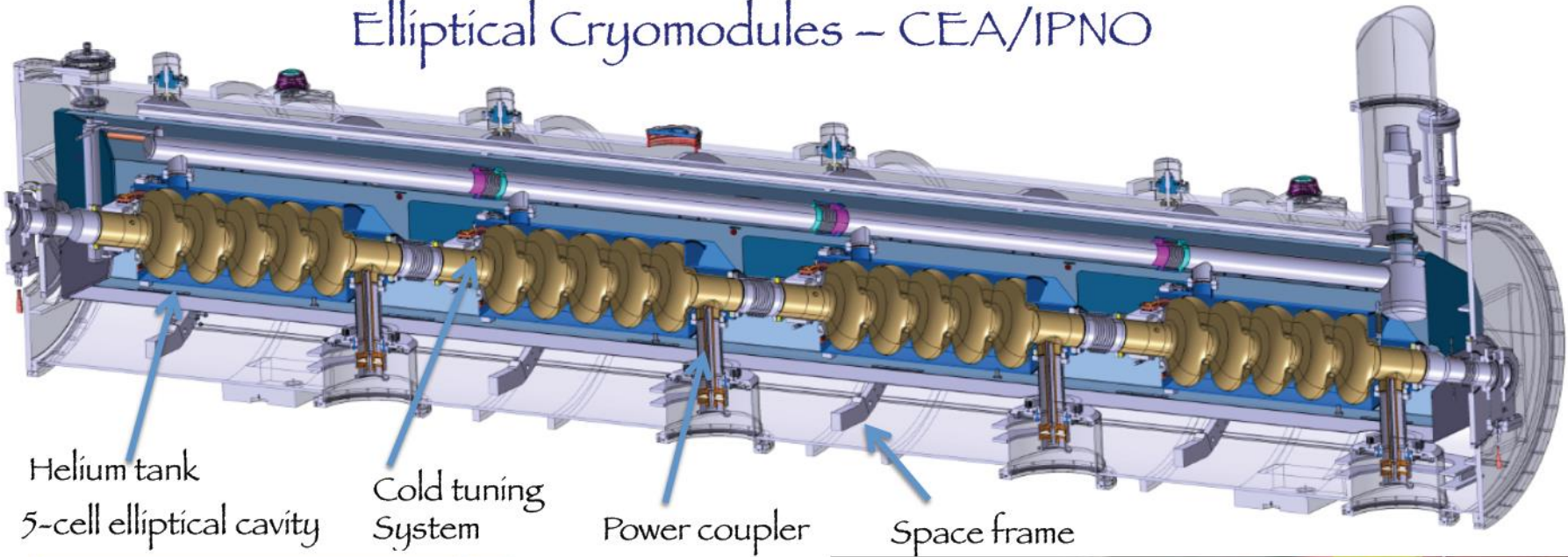
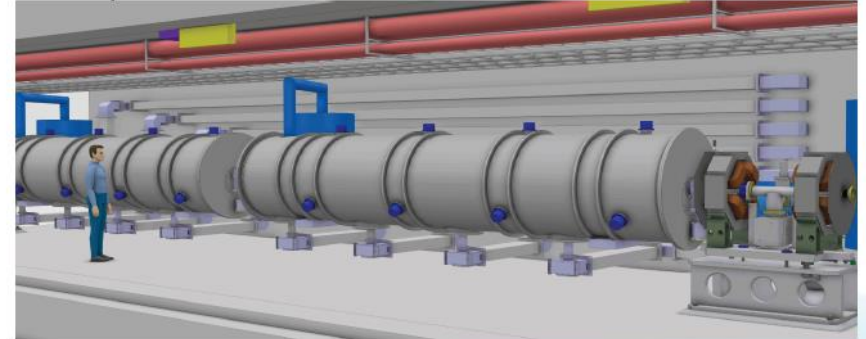
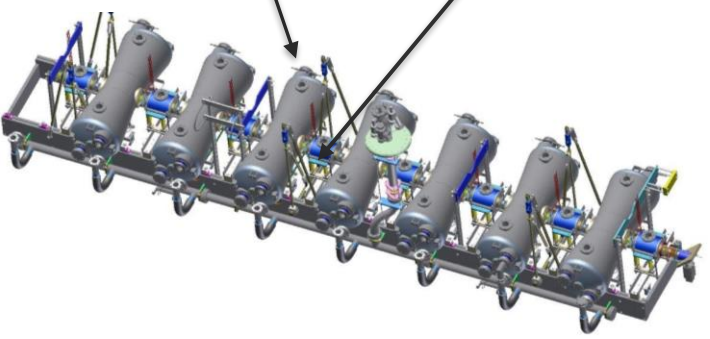


Figure 4.120: Helium vessel with hanging rod

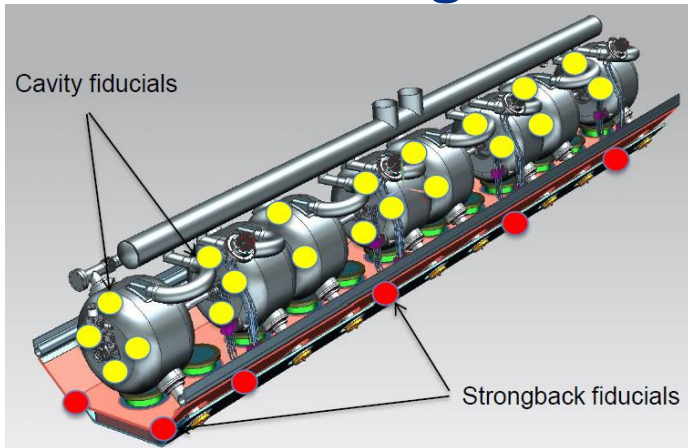


CM for 325 MHz HWR cavities for PIP II (developed by ANL)

cavities
solenoids

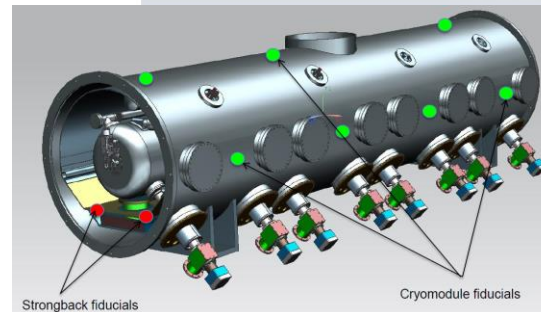
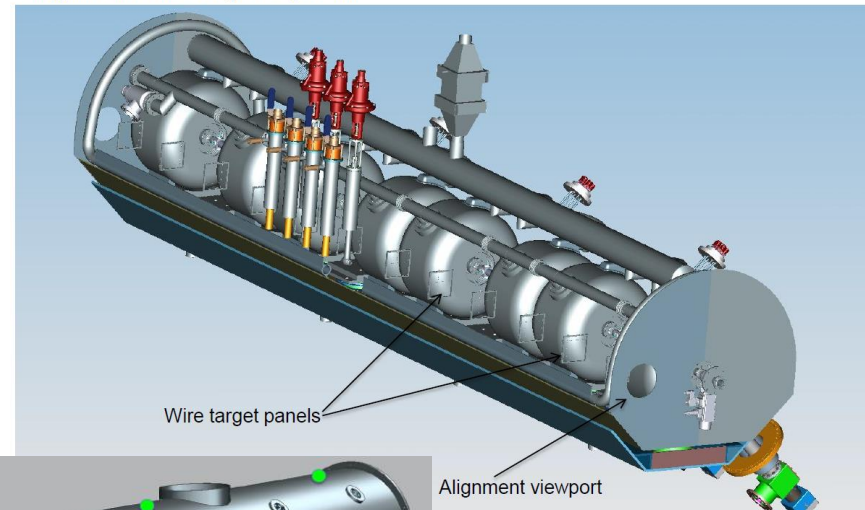


Cavities alignment



- Special alignment wire target panels (4 per each component, 1 upstream and 1 downstream on each left-right side) – used for relative optical measurement with respect to external reference points
- Each panel has two intersecting wires forming a crosshair along the diagonal, creating 24 configurations

Alignment wire targets system



Transverse cavity alignment, mm RMS $< \pm 1$
Angular alignment, mrad RMS $\leq \pm 10$

- The cavity string is aligned in warm state with an offset to compensate for thermal shrinking.
- Laser Trackers and precision optical and electronic levels is used as instrumentation for the alignment of cavity string
- After completion of cavity string alignment and transfer to the strongback referencing points, the cavities and solenoids temporary fiducials, as well as the fiducials on both sides of the strongback, can be removed

□ Beam – base alignment (BPM, HOMs)

Cryogenics and thermoacoustic oscillations

- LCLS-II is a low-beam-current SRF linac → cavities have very narrow bandwidth, ~ 10 Hz.
- During acceptance testing of the prototype LCLS-II cryomodule, unexpectedly high level of microphonics was encountered preventing stable operation of the cavities in a GDR mode.
- The problem was traced to thermoacoustic oscillations in the supply JT valves.
- Thermoacoustic oscillations generally occur in long gas-filled tubes with a large temperature gradient.
- Acoustic modes couple to mass transport up and down column especially well when gas density is strongly tied to temperature. E.g. warm gas from the top of a valve column moving to the cold bottom contracts, reducing pressure at warm region, driving the now cold gas back.
- These oscillations are generally important for the tremendous heat leaks they can represent, not microphonics.



- Low pressure operation consistently eliminated icing on the supply valves (JT, bypass)
- Indicates **suppression** of thermo-acoustic oscillations

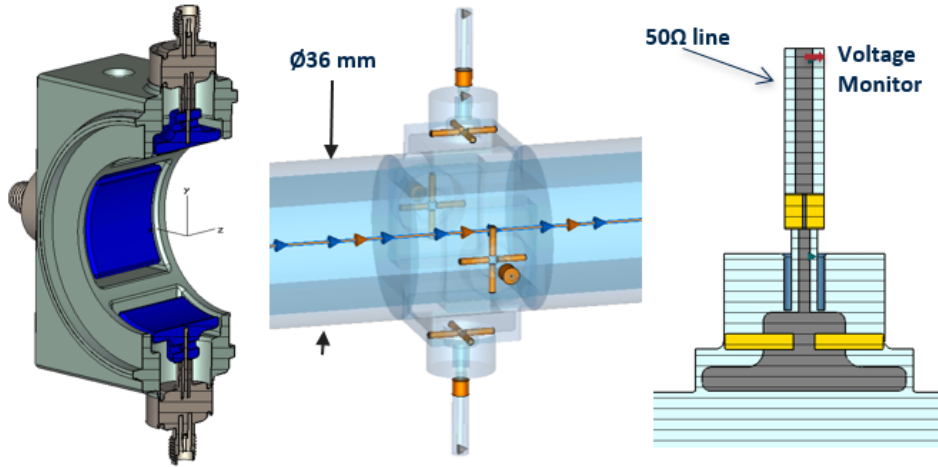
J. Holzbauer, "1.3 GHz Microphonics measurement and mitigations," MRCW18

Beam Diagnostics

- *Beam Position and Phase Measurement System (button –type or stripline BPMs)*
- *Beam Loss Measurement System (ionization chambers).*
- *Beam Intensity Measurement System (DC current transformers and beam toroids)*
- *Beam Transverse Profile Measurement System (traditional wire scanner or photo-disassociation of H^- by laser radiation)*
- *Beam Transverse Emittance Measurements (Allison-Type Emittance Scanners or Laser-Based Emittance Scanners)*

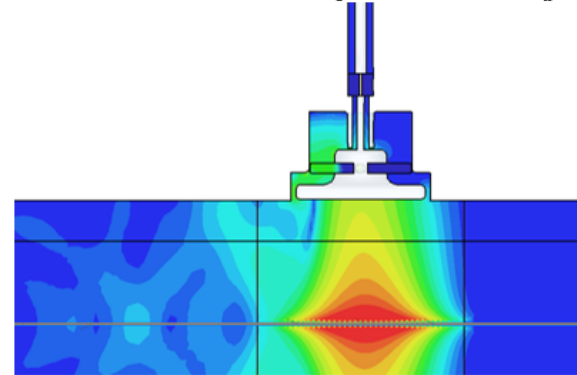
“Don’t try to save on the beam diagnostics!”

Beam Diagnostics in CM, beam position monitors (BPM)



BPM signals produced by the 4 mm rms bunch

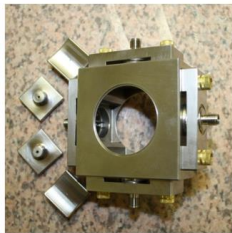
Instant Electric Field induced by the 4 mm bunch ($\beta = 1$)



BPM signals spectral densities for various bunch lengths

- Hermetic feedthrough welding is finished.
- Next step = fit the electrodes and electron beam weld them into position.

BPMs are mounted to the focusing elements:



BPMs for HWR cryomodule, prototype (top) and production (bottom) units



BPMs for SSR1 cryomodule



ring pickup

Summary:

- ❑ RF design of the cavity is based on
 - the accelerator operation regime – pulsed or CW;
 - the beam power and energy;
 - the beam quality requirements.
- ❑ RF cavity parameter optimization includes:
 - frequency,
 - RT versus SRF
 - operating temperature choice for SRF,
 - optimal gradient,
 - cavity shape optimization,
 - number of cells,
 - cell-to-cell coupling,
 - HOM extraction,
 - RF power coupling
- ❑ RF linac is self-consistent system and its subsystem are interconnected; therefore, the RF cavity design is an iterative process.
- ❑ RF cavity design includes:
 - RF parameter optimization;
 - MP analysis
 - Mechanical optimization.

Summary (cont):

- ❑ The SRF cavity component design includes:
 - the input power design;
 - the cavity tuner design;
 - The He vessel design.
- ❑ The SRF cavity manufacturing process contains a lot of operations and requires high technological culture:
 - material quality control;
 - cell manufacturing and pre-tune;
 - final assembly;
 - surface processing;
 - welding into the He vessel;
 - component assembly;
 - Cavity string assembly;
 - cryo-module assembly;
 - alignment
- ❑ The cryo-module:
 - Contains the insulating vacuum.
 - Serves as the major structural element to which all other systems are attached to the accelerator tunnel floor.
 - Serves as a pressure containment vessel in the event of a failure in an internal cryogen line.

Chapter 11.

Architecture of SRF accelerators.

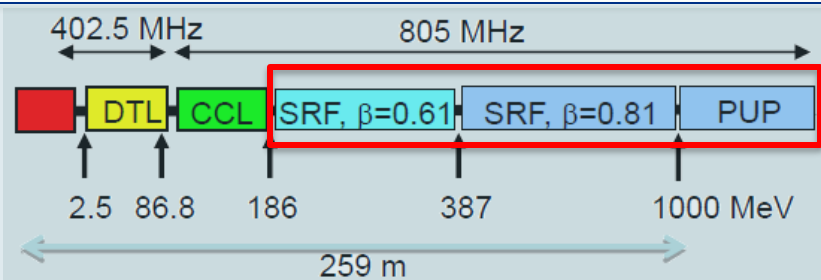
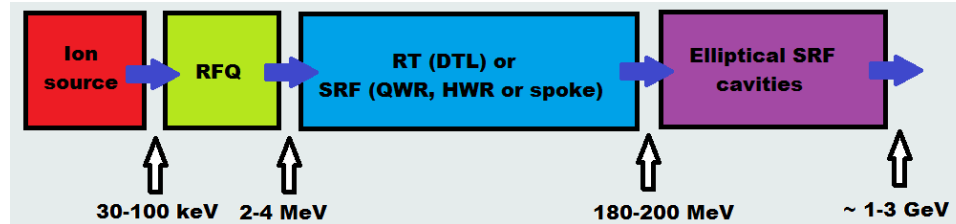
a. proton/ion SRF linacs:

- RT or SRF front end?
- choice of beamline elements;
- lattice design.

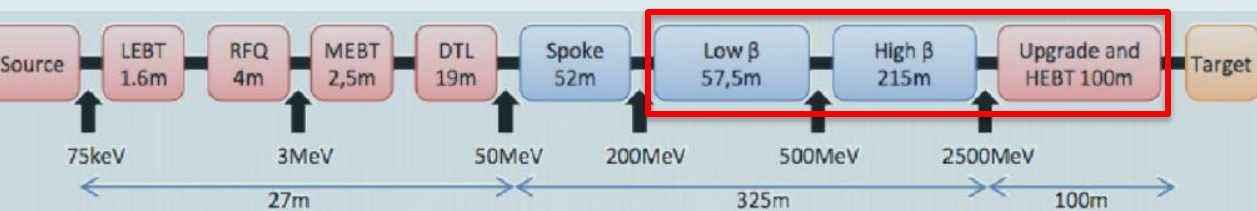
b. electron SRF linacs.

Architecture of a GeV-range proton SRF accelerator:

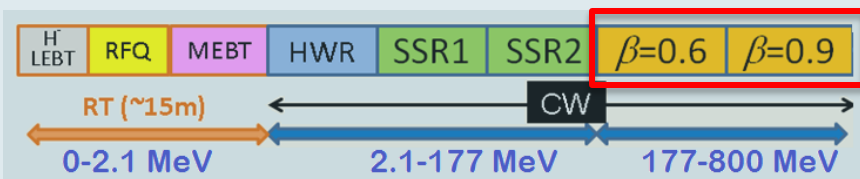
Layout of typical modern proton SRF accelerator.



SNS (ORNL): H^- , 1 GeV, 6% DF, 1.44 MW to accumulator ring
In operation



ESS (Lund): p^+ , 2.5 GeV, 4% DF, 5 MW to target
Under construction



PIP II (FNAL): H^- , 800 MeV, up to 100% DF, up to 1.6 MW
Design

Linac Design Philosophy:

❑ RT or SRF frontend?

- For low duty factor RT frontend (up to ~200 MeV) may be used
- For high DF or CW SRF is necessary from the beginning

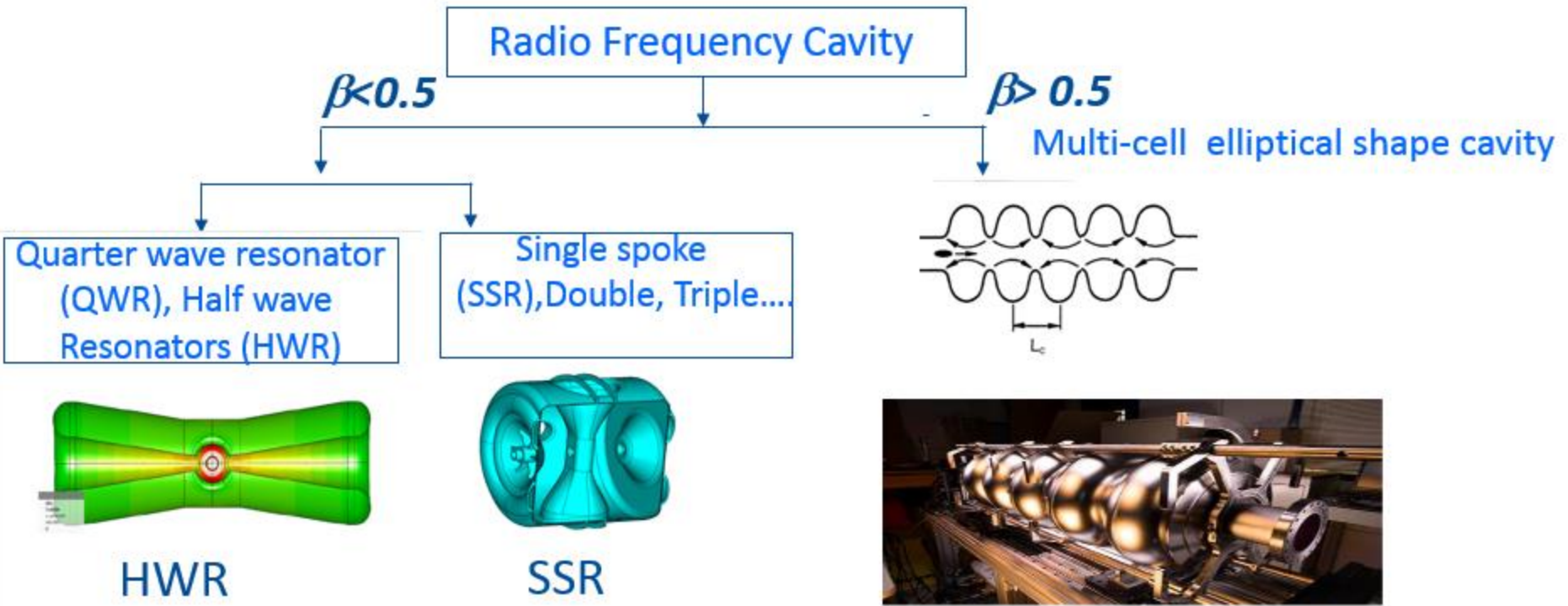
❑ Choice of beam line elements

- Accelerating RF Cavities
- Focusing Magnets

❑ Lattice Design

- Focusing Period
- Transition Energy between Sections

RF cavities:



- Lower RF frequency provides better interaction with beam.
- RF defocusing factor is inversely proportional to frequency.
- Lower frequency implies larger RF bucket and hence larger longitudinal acceptance.

RF cavities:

□ The frequency choices for multi-cell:

- Cavity length is about the same for the same β_G (the same number of couplers, tuners, etc). Typical length $\sim 0.8-1$ m depending on β_G (from iris to iris)
- Lower frequencies \rightarrow bigger size, higher cost, more difficult handling, microphonics but: lower losses per unit length (smaller R/Q , but lower R_s); larger aperture (current interception), smaller beam defocusing; smaller number of cells and therefore, smaller a/λ , smaller K_m and K_e and smaller numbers of cavity types.
- Typically , they use 650 – 800 MHz, and 5-7 cells/cavity:
 - SNS: 804 MHz, 6 cells/cavity (in operation)
 - ESS: 704 MHz, 5 cells/cavity (under construction)
 - PIP II: 650 MHz, 5 cells/cavity (under development)

□ The frequency choices for the front end:

- Subharmonics of the main frequency.

□ Acceleration gradient choice (high DF, CW):

- Quench, $B_{\text{peak}} \approx 70-80$ mT
- Field emission, $E_{\text{peak}} \approx 40$ MV/m
- Thermal breakdown typically is not an issue for proton linacs.

RF cavities:

- Selection of the maximum accelerating gradients in cavities are made on the basis of :
 - Peak surface magnetic field
 - Peak surface Electrical field
- Choices of peak magnetic fields are derived from:
 - Dynamics heat load due to accelerating mode
 - Cavity quenching.
- Choices of peak surface field is made to avoid field emission

CW Linac assumptions:

- 162.5 MHz: $H_{pk} < 50\text{mT}$
- 325 MHz: $H_{pk} < 60\text{mT}$
- 650 MHz: $H_{pk} < 70\text{mT}$
- $E_{pk} < 40 \text{ MV/m}$.

Accelerating Gradient in PIP-II Linac

	HWR	SSR1	SSR2	LB650	HB650
Gradient (MV/m)	9.7	10	11.4	15.9	17.8

Focusing elements:

- ❖ Normal conducting magnets are cheaper but superconducting magnets are:
 - Compact in size
 - Provide intense magnetic field with low power consumption.
- ❖ Low energy part of SRF linac typically has solenoidal focusing:
 - ❖ Provide radial focusing
- ❖ Intermediate and high energy section of linac use normal conducting doublet focusing.
 - ❖ Simplify cavity magnetic shielding requirements
- ❖ Correctors are built in each magnets.
- ❖ Solenoidal and doublet focussing keeps the beam round in transverse planes.

- *Focusing magnets in each section*

<u>Section</u>	HWR	SSR1	SSR2	LB650	HB650
<u>Magnet</u>	S	S	S	FD	FD

S – solenoid, FD – doublet (F : focusing and D: Defocusing quadrupole).

Focusing elements:

□ Solenoid:

Solenoid focal length f :
(non-relativistic case, T is a particle kinetic energy, $T=mv^2/2$.)

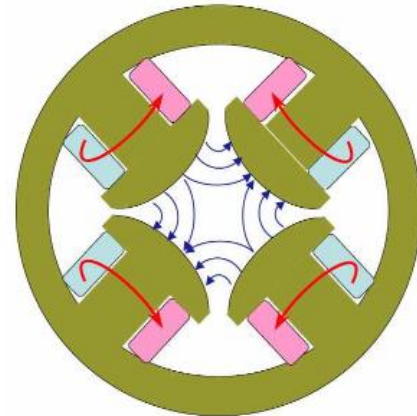
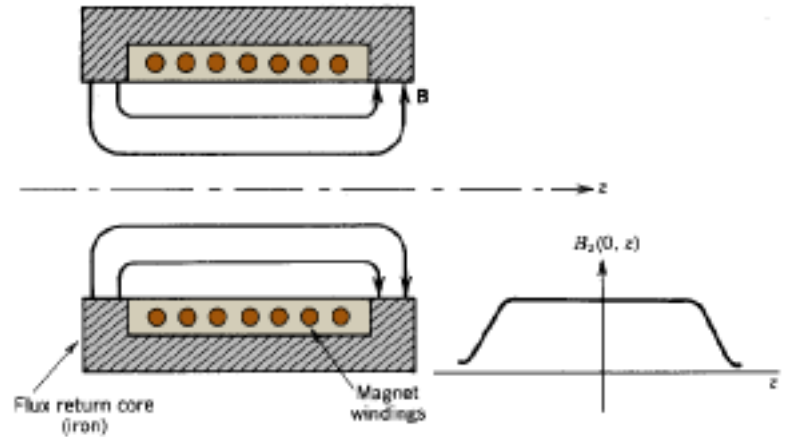
$$\frac{1}{f} = \frac{q^2}{8Tm} \int B_z^2 dz$$

- Focal length is proportional to β^2 ;
- Focal length is inversed proportional to $B_z^2 L$, L is the solenoid length;
- Therefore, solenoid can be used for low β ($\beta < 0.5$). For higher β quad is used.

□ Quadrupole lens:

Quad focal length f :

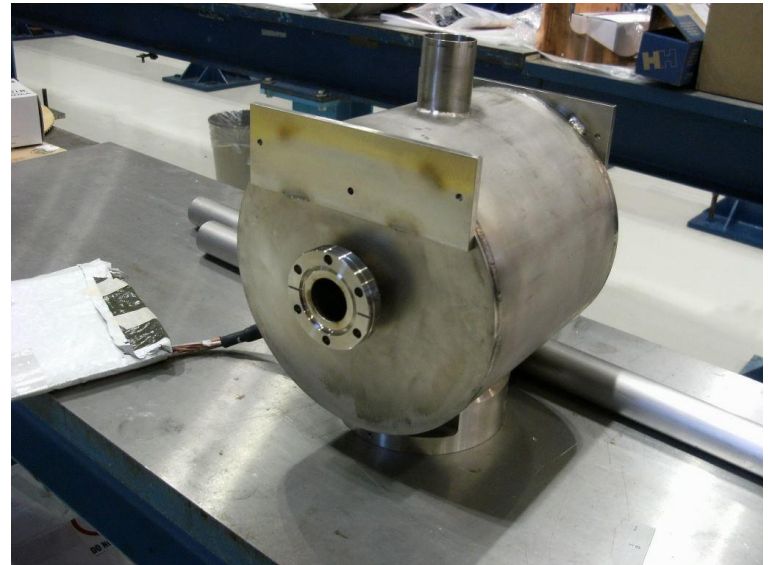
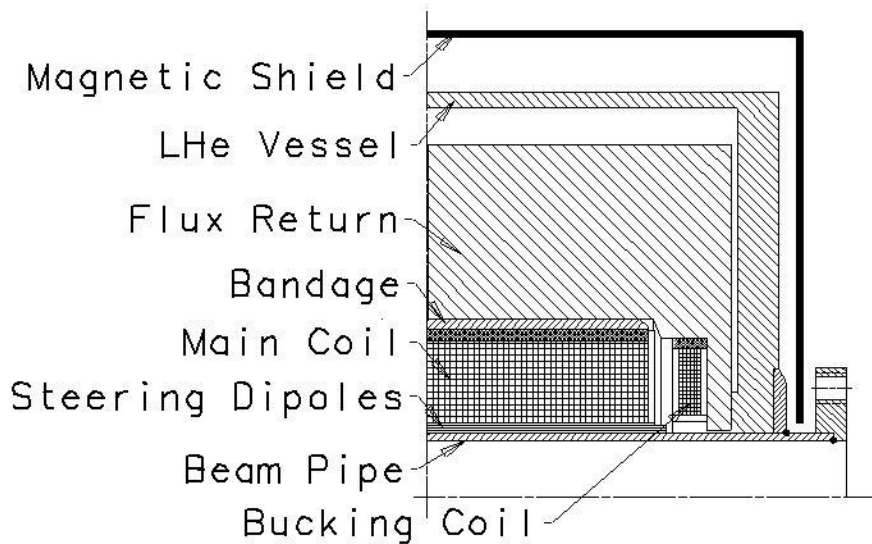
$$\frac{1}{f} = \frac{qB'L}{\gamma\beta mc^2}$$



Focusing elements:

For low section SC solenoids are used.

- Simple and inexpensive;
- Filed up to 6-8 T;
- SRF cavity should have < 10 mT on the SRF cavity surface: remnant solenoid field should be compensated
- Solenoid contains correction coils (steering dipoles)
- Alignment (typically < 0.3 - 0.5 mm, < 5 mrad tilt);
- Quench protection;
- Leads

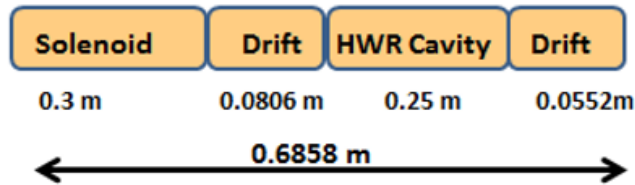


Lattice Design: Focusing Periods

- Length of the focusing period is kept short, especially in the low energy section where beam is non-relativistic and non-linear force may be significant.

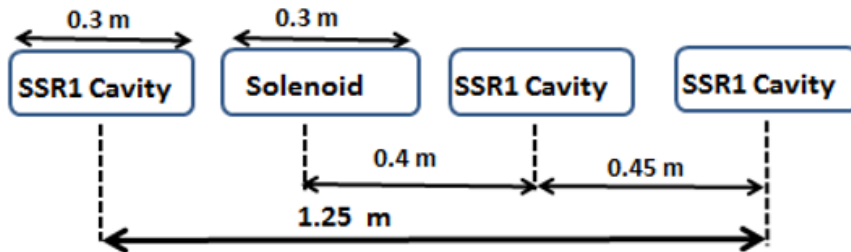
Cryomodule Arrangement

HWR Section :



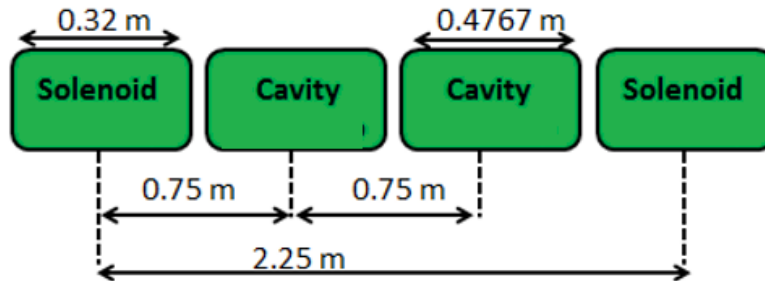
SC-SC-SC-SC-SC-SC-SC-SC

SSR1 Section :



CSC-CSC-CSC-CSC

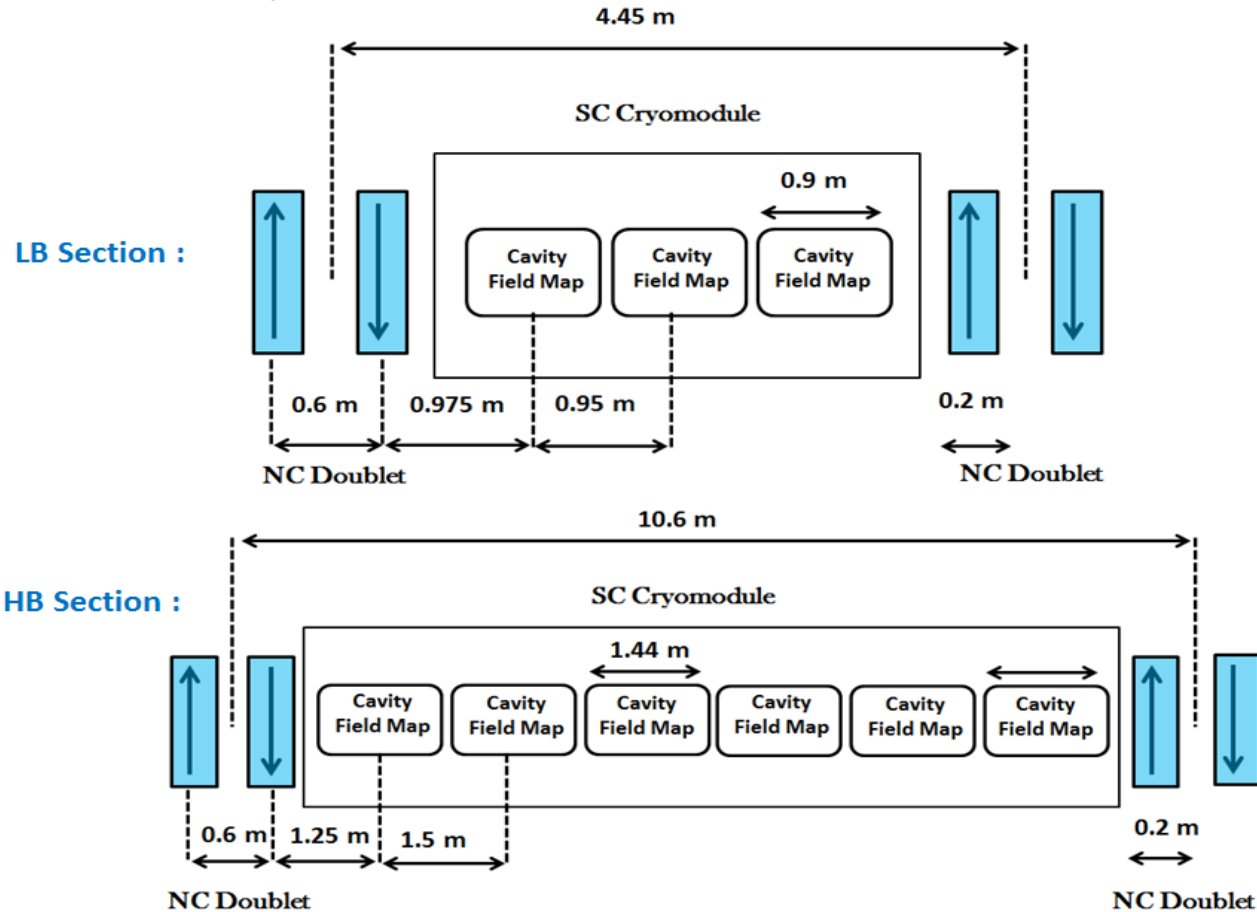
SSR2 Section :



SCC-SCC-SC

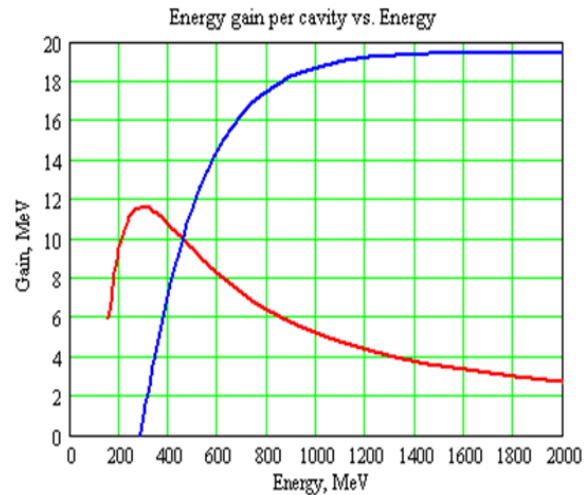
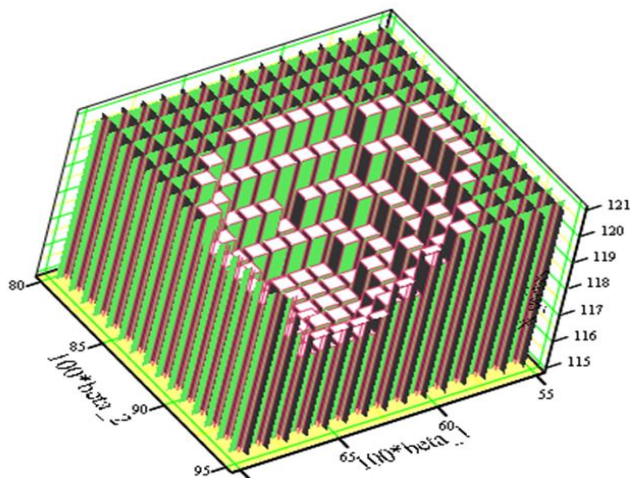
Lattice Design: Focusing Period in High Energy Section

- Frequency jump from 325 MHz to 650 MHz at LB650 MHz section
- Solenoidal focusing is replaced with quadrupole doublet.
- Same family of doublet is used in both LB650 and HB650 sections.



Transition Energy between Sections :

- Transition Energy between Sections (type cavity change). Optimization in order to minimize the number of cavities.
- Beam matching between sections and cryomodules are achieved using elements of each side of transitions. Avoiding abrupt changes in beam envelopes to reduce possibility of halo formations.
- Adiabatic variation in phase advance along linac. Reduces possibility of beam mismatch.

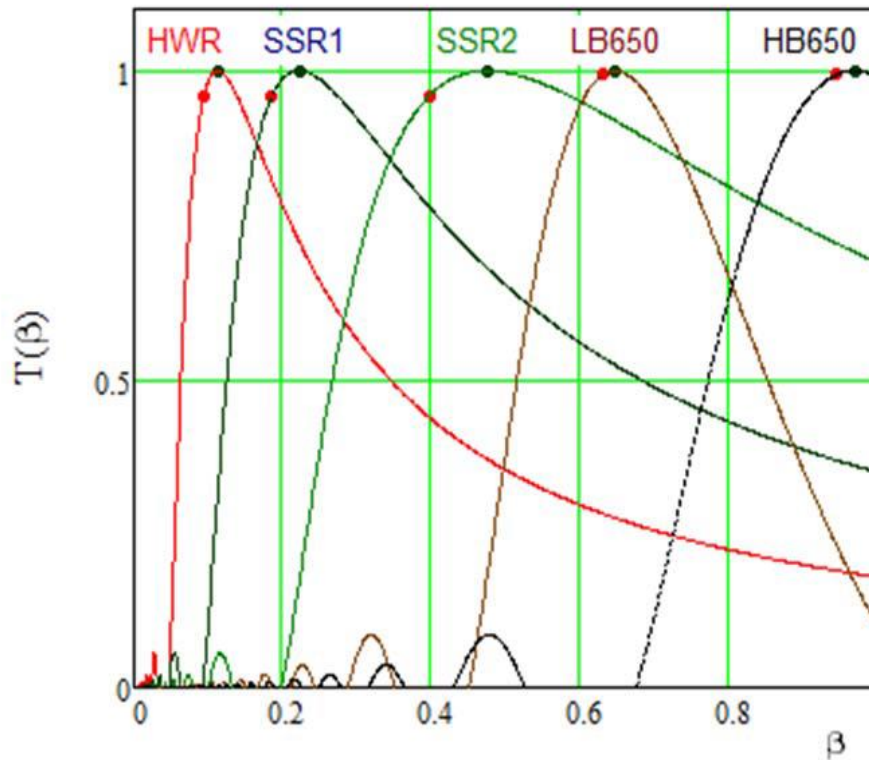


Number of cavities required for acceleration from 185 to 800 MeV versus cavity beta in the LB650 and HB650 sections (left) and the energy gain per cavity versus particle energy (right) for LB650 (red curve) and HB650 (blue curve) cavities.

Architecture of a GeV-range proton SRF accelerator:

Correct selections of transitional energy provide better optimization of real estate gradient and reduction in total number of beam line elements.

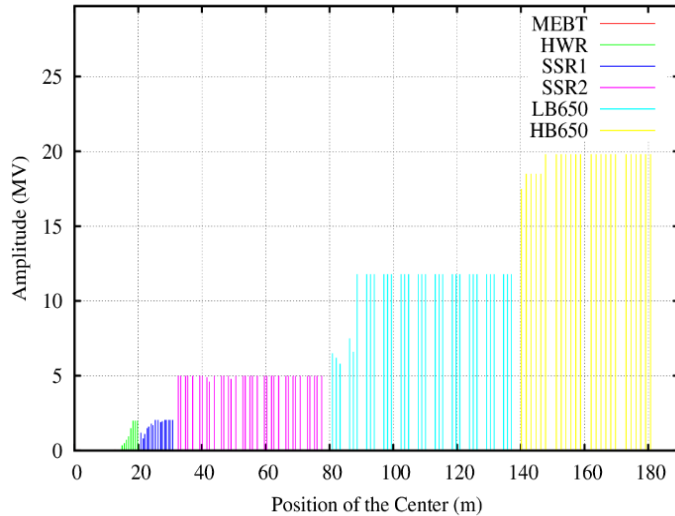
Transition time factor v/s beta



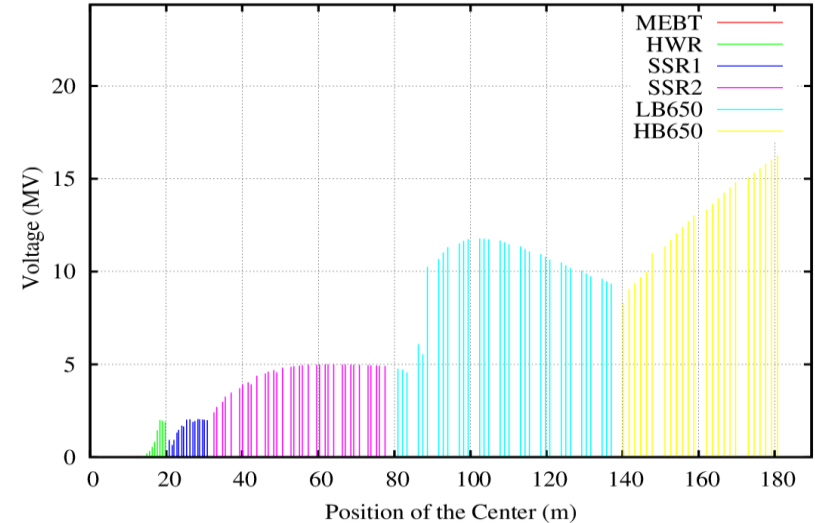
Sections	Initial Energy (MeV)	Design Beta	Beta range
HWR	2.1	0.094	0.067 -0.147
SSR1	10.3	0.186	0.147-0.266
SSR2	35	0.398	0.266-0.55
LB 650	185	0.61	0.55-0.758
HB 650	500	0.92	0.758-0.842

Acceleration voltage distribution

Voltage Amplitude in Cavities

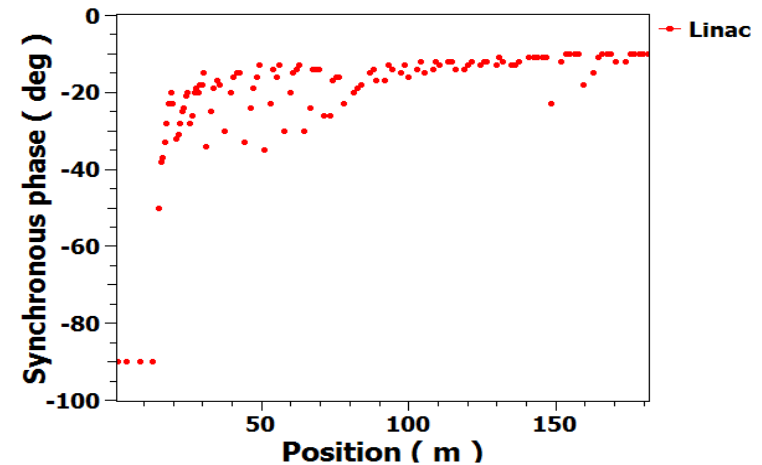


Voltage gain by beam



- Maximum Energy gain in PIP-II SC cavities

	HWR	SSR1	SSR2	LB650	HB650
Max. Egain (MeV)	2	2.05	5	11.9	19.9



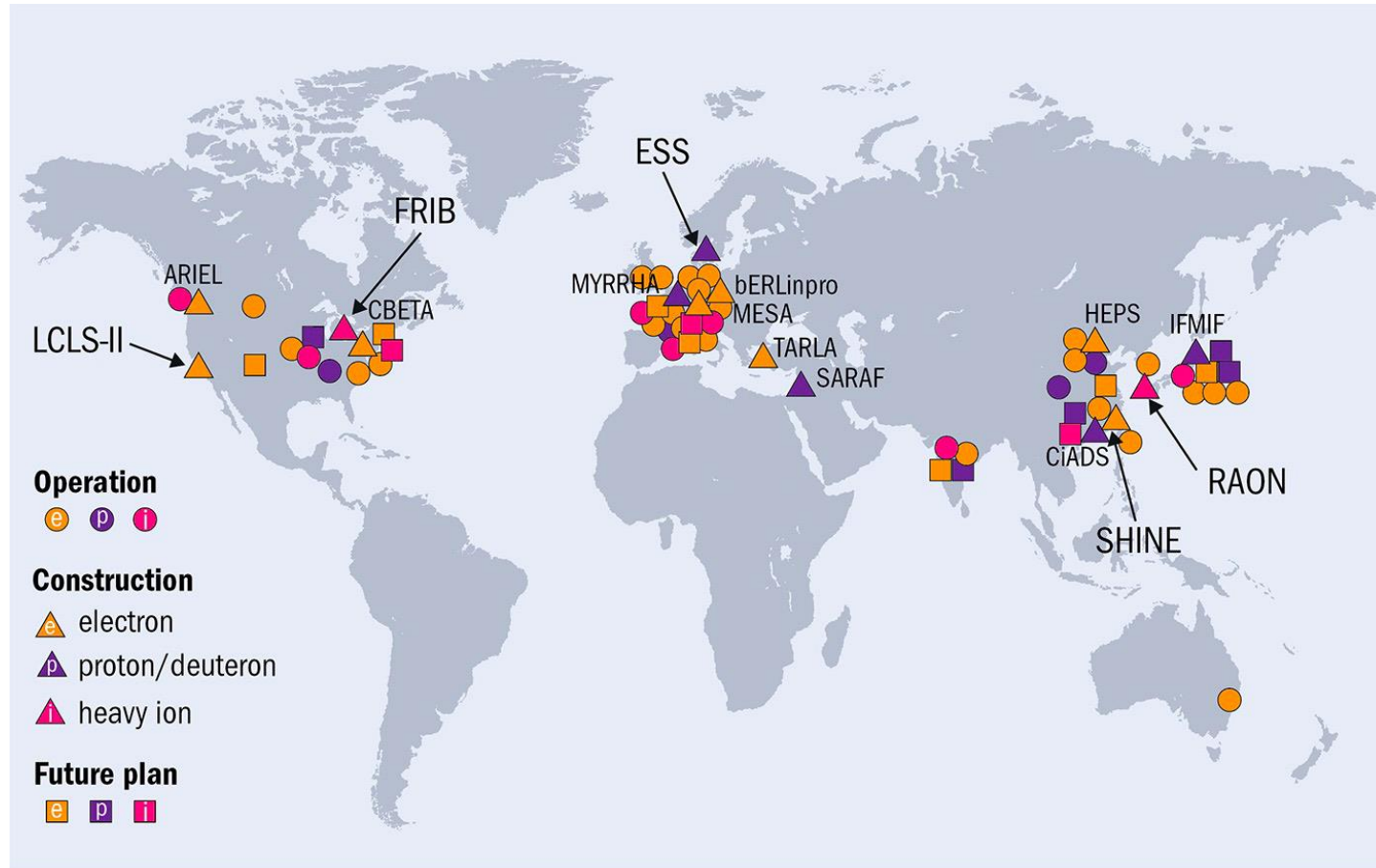
Summary:

- ❑ Architecture of a big SRF linac is determined by:
 - accelerated particles – electrons, protons or ions;
 - accelerator operation regime – pulsed or CW;
 - accelerator parameters – energy and power.
- ❑ For a proton accelerator the choice of the front end – RT or SRF – depends on the operation regime, pulsed or CW.
- ❑ The frequencies and cavity types for a proton or an ion accelerator should be determined;
- ❑ The types of the focusing elements should be selected.
- ❑ The lattice should be designed, which provides
 - acceleration
 - focusing
 - bunching.
- ❑ Break points between the section with different cavity types should be optimized;
- ❑ The sections should be matched to each other to provide required beam quality.

Chapter 12.

SRF around the world

SRF around the world



Global view Distribution of superconducting particle accelerators using SRF structures for electrons (orange), protons (purple) and heavy ions (pink). More than 30 SRF accelerators are in operation (circles), approximately 15 are presently under construction (triangles) and more than 10 future projects are under consideration (squares). *Credit: CERN*

Big SRF Accelerator Facilities:

Linac	Laboratory	Application	Acc. Particle	Operation	Status
SNS	ORNL, USA	Neutron Source	H⁻	pulsed	Operation
ESS	ESS, Sweden	Neutron Source	p	pulsed	Construction
CIADS	IMP, China	ADS	p	CW	R&D
ISNS	Indore, India	Neutron Source	p	pulsed	R&D
ADSS	BARC, India	ADS	p	CW	R&D
PIP II	FNAL, USA	Neutrino/Muons	H ⁻	CW/pulsed	R&D
FRIB	MSU, USA	Nucellar physics	Ions	CW	Commissioned
RAON	RISP, S.Korea	Nucellar physics	Ions	CW	Construction
CEBAF	JLAB, USA	Nucellar physics	e⁻	CW	Operation
XFEL	DESY, Germany	FEL	e⁻	pulsed	Operation
SHINE	SINAP, China	FEL	e ⁻	CW	Construction
LCLS II	SLAC, USA	FEL	e ⁻	CW	Construction

New large SRF accelerator installations

CEBAF Upgrade - JLAB Upgrade 6.5 GeV => 12 GeV electrons	80 cavities	Electrons
XFEL – Hamburg, Germany 17.5 GeV electrons – Pulsed X-ray FEL	840 cavities	
LCLS-II (+ LCLS-II-HE) – SLAC 4 GeV electrons – CW X-ray FEL	296 (+184) cavities	
SPIRAL-II – France 30 MeV, 5 mA protons -> Heavy Ions	28 cavities	Ions
FRIB – MSU 400 kW, heavy ion beams for nuclear astrophysics	340 cavities	
RISP - Daejeon, South Korea 400 kW, heavy ion beams for nuclear astrophysics	340 cavities	
ESS – Sweden 1 – 2 GeV, 5 MW Pulsed spallation source	150 cavities	Protons
PIP-II – Fermilab 800 MV High intensity proton linac for neutrino beams	115 cavities	

Coming up: *SHINE* in China, *EIC* at BNL, *ILC* in Japan, *FCC-ee/FCC-hh* at CERN, *CEPC-SPPC* in China, *Accelerator complex upgrade to 2.4 MW* at Fermilab

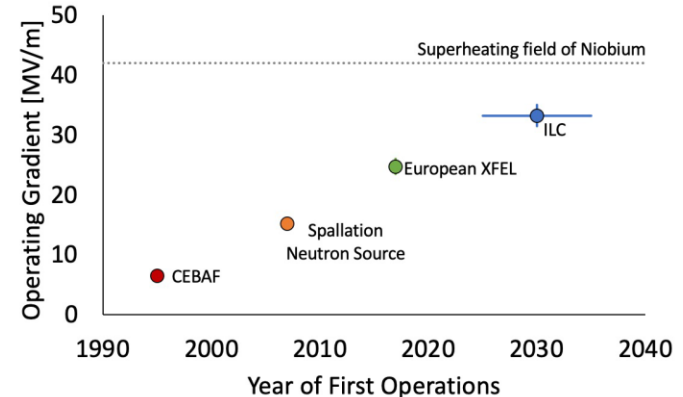
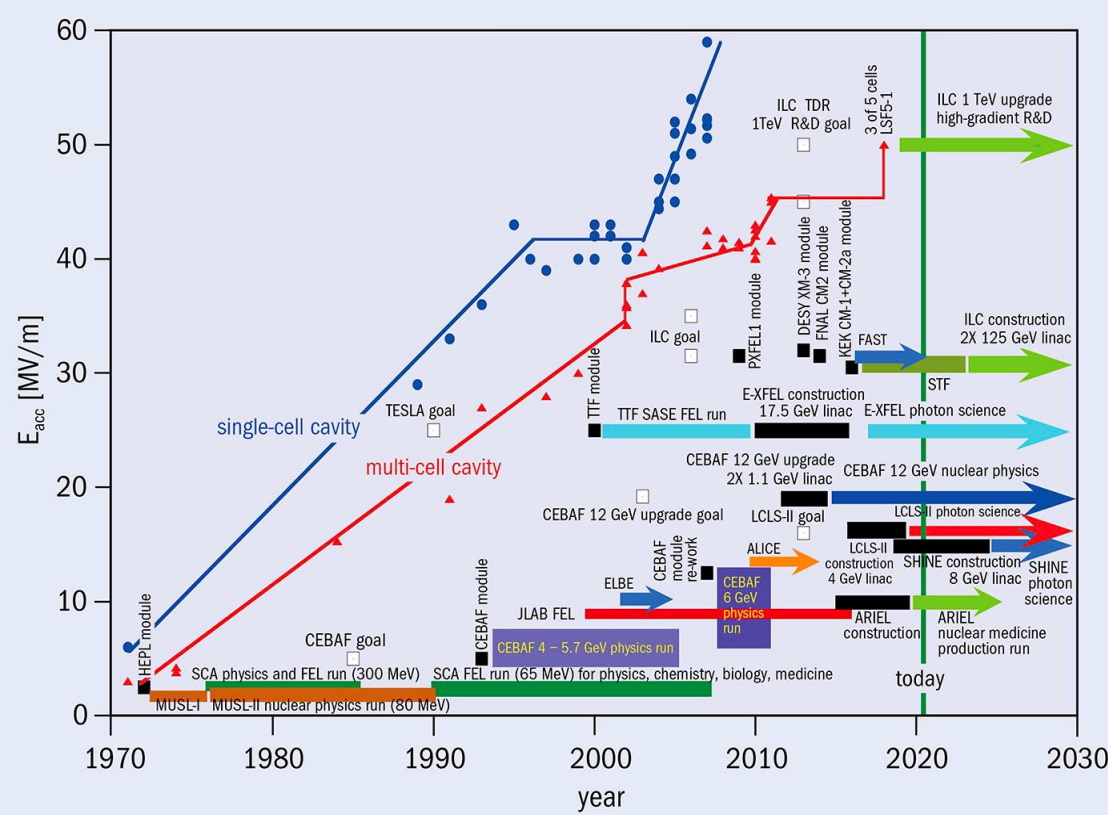
SRF gradient achievements and applications

Gradient growth SRF linac accelerating gradient achievements and application specifications since 1970.

CW SRF Linacs – SCA: Stanford Superconducting Accelerator; MUSL: Illinois Microtron Using a Superconducting Linac; CEBAF: Continuous Electron Beam Accelerator Facility; JLab FEL: JLab Free Electron Laser; ELBE: HZDR Electron Linear accelerator with high Brilliance and Low Emittance; ALICE: STFC Accelerators and Lasers In Combined Experiments; ARIEL: TRIUMF Advanced Rare Isotope Laboratory; LCLS-II:

Linac Coherence Light Source extension; SHINE: Shanghai High Brightness Photon Facility.

Pulsed SRF Linacs – FAST: Fermilab Accelerator Science and Technology Facility; STF: KEK Superconducting RF Test Facility; E-XFEL: European X-ray Free Electron Laser; ILC: International Linear Collider.



Credit: Source: R Geng/ORNL



Fermilab PIP-II



(115 Cavities)



Half Wave



Spokes

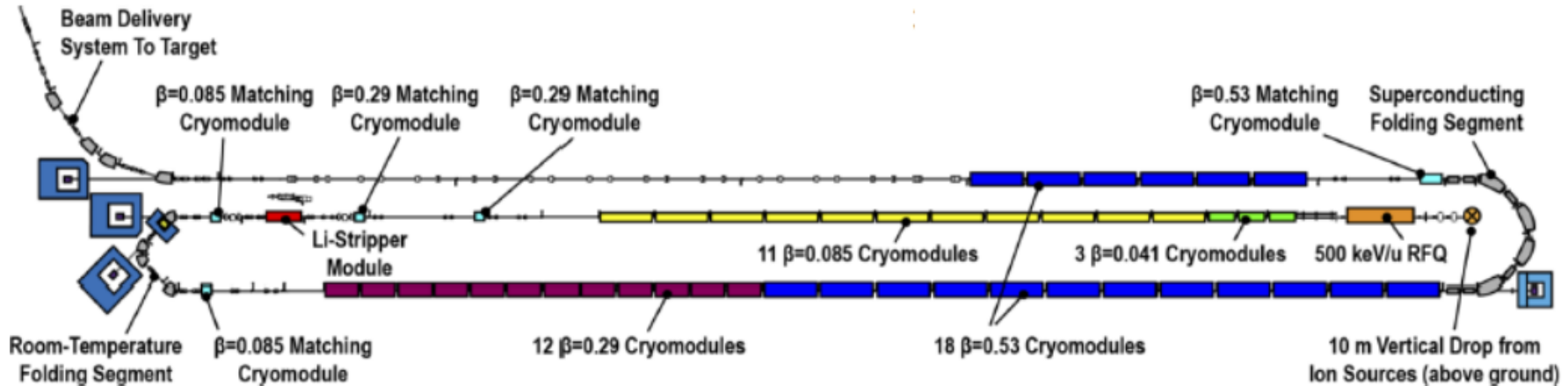


Medium-Beta Elliptical Cavities

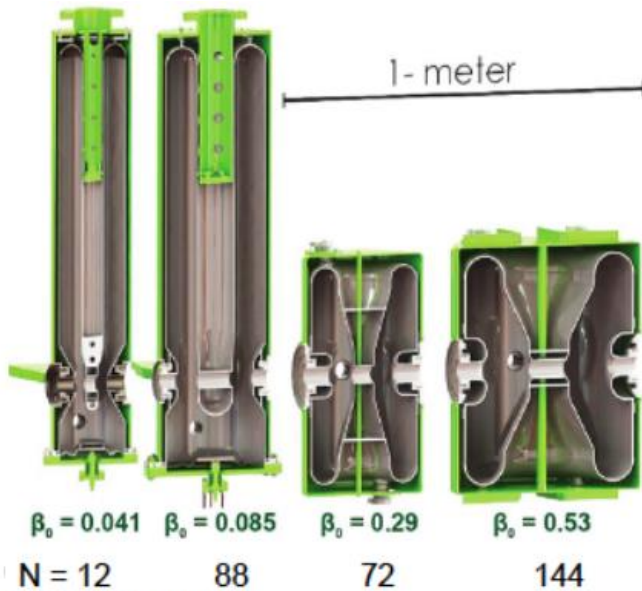


High-Beta Elliptical Cavities

Architecture of Facility for Rare Isotope Beams (FRIB, MSU)



K. Saito, September 2014 LINAC14 TH10A02,



316 cavities need,
total 347 including matching module, spares

FRIB cavities

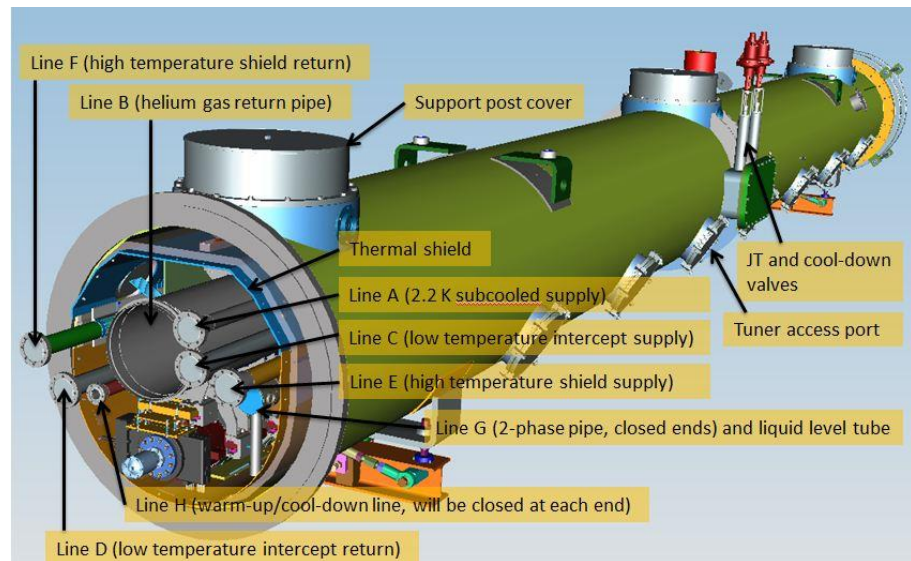
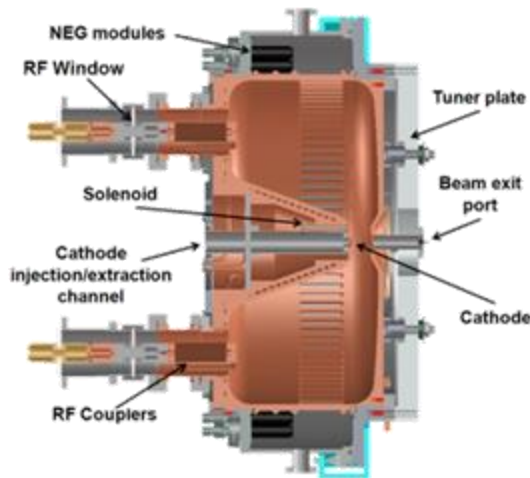
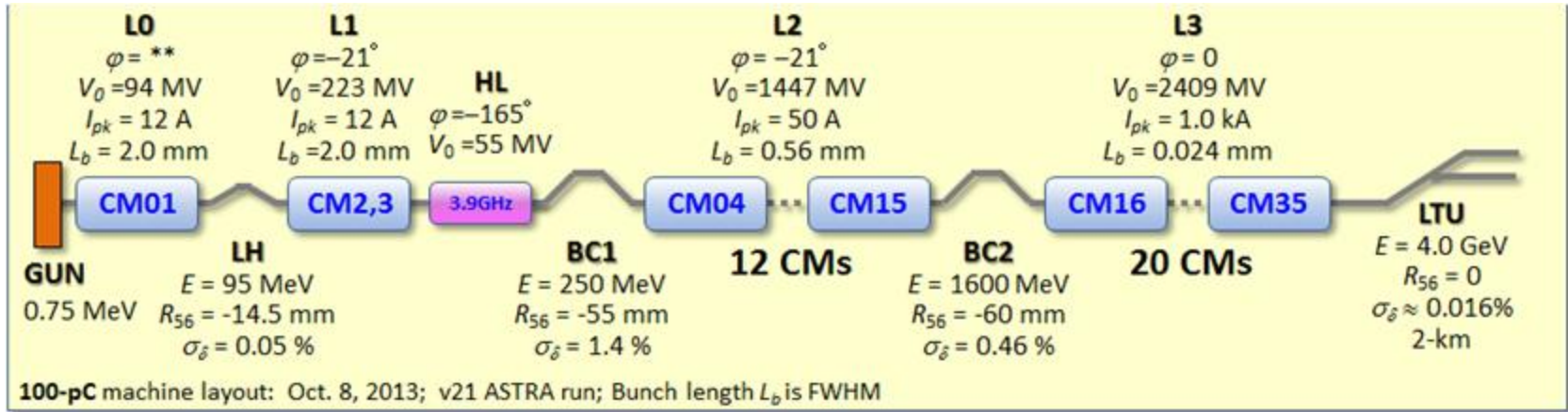
Cavity Type	QWR	QWR	HWR	HWR
β_0	0.041	0.085	0.285	0.53
f [MHz]	80.5	80.5	322	322
V_a [MV]	0.810	1.80	2.09	3.70
E_{acc} [MV/m]	5.29	5.68	7.89	7.51
E_p/E_{acc}	5.82	5.89	4.22	3.53
B_p/E_{acc} [mT/(MV/m)]	10.3	12.1	7.55	8.41
R/Q [Ω]	402	455	224	230
G [Ω]	15.3	22.3	77.9	107
Aperture [m]	0.036	0.036	0.040	0.040
$L_{eff} \equiv \beta\lambda$ [m]	0.153	0.317	0.265	0.493
Lorenz detuning [Hz/(MV/m) ²]	< 4	< 4	< 4	< 4
Specific $Q_0@VT$	1.4E+9	2.0E+9	5.5e+9	9.2E+9
Q_L	6.3E+6	1.9E+6	5.6E+6	9.7E+6

FRIB linac is successfully commissioned in May 2021!



5-cell 644 MHz cavities for FRIB upgrade

Architecture of LCLS II (electron SRF Linac for FEL)



European XFEL

View Along the 1 km Long
Superconducting Accelerator

- EXFEL is the world largest SRF application at 17.5 GeV (800 cavities).
- Operating gradient is 23.5 MV/m.
- Construction is complete, commissioning has started.
- First lasing in May 2017!

IPAC '17, 15.5.2017
Winnli Decking, DESY



Επιλογος

“The internal machinery of life, the chemistry of the parts, is something beautiful. And it turns out that all life is interconnected with all other life.”
Richard Feynman

- ❖ SRF Liner Accelerator is self-consistent system, parts of it strongly depend on each other. Deep understanding and careful analysis of subsystems and components as well as their interaction are necessary to achieve required beam parameters and facility reliability at minimal capital and operation cost.
- ❖ The design process will never be reduced to just a few simple rules or recipes. Using an existing design as a base for developing a new system is OK and can shorten the new system development time, but the system designers should be aware that even seemingly small changes could bring big consequences.
- ❖ As accelerator application demands continue to increase (higher energy, higher luminosity, brighter beams, more efficient accelerators, ...) there will be no shortage of new challenges to tackle in the future.
- ❖ The field of RF superconductivity is very active. The SRF technology is the technology of choice for many types of accelerators.

There will always be ample opportunities for imagination, originality, and common sense.



The end

Appendixes

Appendix 1, Vector calculus

$\nabla \cdot (\psi \mathbf{A}) = \mathbf{A} \cdot \nabla \psi + \psi \nabla \cdot \mathbf{A}$	$\text{div}(\psi \mathbf{A}) = \mathbf{A} \cdot \text{grad} \psi + \psi \text{div} \mathbf{A}$
$\nabla \times (\psi \mathbf{A}) = \nabla \psi \times \mathbf{A} + \psi \nabla \times \mathbf{A}$	$\text{curl}(\psi \mathbf{A}) = \text{grad} \psi \times \mathbf{A} + \psi \text{curl} \mathbf{A}$
$\nabla(\mathbf{A} \cdot \mathbf{B}) = (\mathbf{A} \cdot \nabla) \mathbf{B} + (\mathbf{B} \cdot \nabla) \mathbf{A} + \mathbf{A} \times (\nabla \times \mathbf{B}) + \mathbf{B} \times (\nabla \times \mathbf{A})$	$\text{grad}(\mathbf{A} \cdot \mathbf{B}) = (\mathbf{A} \cdot \nabla) \mathbf{B} + (\mathbf{B} \cdot \nabla) \mathbf{A} + \mathbf{A} \times \text{curl} \mathbf{B} + \mathbf{B} \times \text{curl} \mathbf{A}$
$\frac{1}{2} \nabla A^2 = \mathbf{A} \times (\nabla \times \mathbf{A}) + (\mathbf{A} \cdot \nabla) \mathbf{A}$	$\frac{1}{2} \text{grad} A^2 = \mathbf{A} \times (\text{curl} \mathbf{A}) + (\mathbf{A} \cdot \nabla) \mathbf{A}$
$\nabla \cdot (\mathbf{A} \times \mathbf{B}) = \mathbf{B} \cdot \nabla \times \mathbf{A} - \mathbf{A} \cdot \nabla \times \mathbf{B}$	$\text{div}(\mathbf{A} \times \mathbf{B}) = \mathbf{B} \cdot \text{curl} \mathbf{A} - \mathbf{A} \cdot \text{curl} \mathbf{B}$
$\nabla \times (\mathbf{A} \times \mathbf{B}) = \mathbf{A}(\nabla \cdot \mathbf{B}) - \mathbf{B}(\nabla \cdot \mathbf{A}) + (\mathbf{B} \cdot \nabla) \mathbf{A} - (\mathbf{A} \cdot \nabla) \mathbf{B}$	$\text{curl}(\mathbf{A} \times \mathbf{B}) = \mathbf{A}(\text{div} \mathbf{B}) - \mathbf{B}(\text{div} \mathbf{A}) + (\mathbf{B} \cdot \nabla) \mathbf{A} - (\mathbf{A} \cdot \nabla) \mathbf{B}$

P.W.

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Appendix 1. Vector calculus

$$\nabla\psi = \text{grad } \psi$$

$$\nabla \cdot \mathbf{A} = \text{div } \mathbf{A}$$

$$\nabla \times \mathbf{A} = \text{curl } \mathbf{A} \equiv \text{rot } \mathbf{A}$$

$$\Delta = \nabla^2 = \nabla \cdot \nabla$$

$$\nabla \times (\nabla\psi) = 0$$

$$\text{curl}(\text{grad } \psi) = 0$$

$$\nabla \cdot (\nabla \times \mathbf{A}) = 0$$

$$\text{div}(\text{curl } \mathbf{A}) = 0$$

$$\Delta \psi = \nabla \cdot (\nabla\psi) = \nabla^2 \psi$$

$$\Delta \psi = \text{div}(\text{grad } \psi)$$

$$\nabla \times \nabla \times \mathbf{A} = \nabla(\nabla \cdot \mathbf{A}) - \nabla^2 \mathbf{A}$$

$$\text{curl}(\text{curl } \mathbf{A}) = \text{grad}(\text{div } \mathbf{A}) - \Delta \mathbf{A}$$

$$\nabla = \frac{\partial}{\partial x} \vec{i} + \frac{\partial}{\partial y} \vec{j} + \frac{\partial}{\partial z} \vec{k},$$

$$\int \text{curl } \mathbf{F} \cdot d\mathbf{S} = \oint \mathbf{F} \cdot d\mathbf{l}, \quad \int \text{div } \mathbf{F} dV = \int \mathbf{F} \cdot d\mathbf{S}$$

Stokes theorem

Gauss theorem

Appendix 1. Vector calculus

Differential operators in different coordinates:

Cartesian

Cylindrical

Spherical

\mathbf{A}	$A_x \hat{\mathbf{x}} + A_y \hat{\mathbf{y}} + A_z \hat{\mathbf{z}}$	$A_\rho \hat{\boldsymbol{\rho}} + A_\varphi \hat{\boldsymbol{\varphi}} + A_z \hat{\mathbf{z}}$	$A_r \hat{\mathbf{r}} + A_\theta \hat{\boldsymbol{\theta}} + A_\varphi \hat{\boldsymbol{\varphi}}$
∇f	$\frac{\partial f}{\partial x} \hat{\mathbf{x}} + \frac{\partial f}{\partial y} \hat{\mathbf{y}} + \frac{\partial f}{\partial z} \hat{\mathbf{z}}$	$\frac{\partial f}{\partial \rho} \hat{\boldsymbol{\rho}} + \frac{1}{\rho} \frac{\partial f}{\partial \varphi} \hat{\boldsymbol{\varphi}} + \frac{\partial f}{\partial z} \hat{\mathbf{z}}$	$\frac{\partial f}{\partial r} \hat{\mathbf{r}} + \frac{1}{r} \frac{\partial f}{\partial \theta} \hat{\boldsymbol{\theta}} + \frac{1}{r \sin \theta} \frac{\partial f}{\partial \varphi} \hat{\boldsymbol{\varphi}}$
$\nabla \cdot \mathbf{A}$	$\frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}$	$\frac{1}{\rho} \frac{\partial (\rho A_\rho)}{\partial \rho} + \frac{1}{\rho} \frac{\partial A_\varphi}{\partial \varphi} + \frac{\partial A_z}{\partial z}$	$\frac{1}{r^2} \frac{\partial (r^2 A_r)}{\partial r} + \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (A_\theta \sin \theta) + \frac{1}{r \sin \theta} \frac{\partial A_\varphi}{\partial \varphi}$
$\nabla \times \mathbf{A}$	$\begin{aligned} &\left(\frac{\partial A_z}{\partial y} - \frac{\partial A_y}{\partial z} \right) \hat{\mathbf{x}} + \\ &\left(\frac{\partial A_x}{\partial z} - \frac{\partial A_z}{\partial x} \right) \hat{\mathbf{y}} + \\ &\left(\frac{\partial A_y}{\partial x} - \frac{\partial A_x}{\partial y} \right) \hat{\mathbf{z}} \end{aligned}$	$\begin{aligned} &\left(\frac{1}{\rho} \frac{\partial A_z}{\partial \varphi} - \frac{\partial A_\varphi}{\partial z} \right) \hat{\boldsymbol{\rho}} + \\ &\left(\frac{\partial A_\rho}{\partial z} - \frac{\partial A_z}{\partial \rho} \right) \hat{\boldsymbol{\varphi}} + \\ &\frac{1}{\rho} \left(\frac{\partial (\rho A_\varphi)}{\partial \rho} - \frac{\partial A_\rho}{\partial \varphi} \right) \hat{\mathbf{z}} \end{aligned}$	$\begin{aligned} &\frac{1}{r \sin \theta} \left(\frac{\partial}{\partial \theta} (A_\varphi \sin \theta) - \frac{\partial A_\theta}{\partial \varphi} \right) \hat{\mathbf{r}} + \\ &\frac{1}{r} \left(\frac{1}{\sin \theta} \frac{\partial A_r}{\partial \varphi} - \frac{\partial}{\partial r} (r A_\varphi) \right) \hat{\boldsymbol{\theta}} + \\ &\frac{1}{r} \left(\frac{\partial}{\partial r} (r A_\theta) - \frac{\partial A_r}{\partial \theta} \right) \hat{\boldsymbol{\varphi}} \end{aligned}$
$\Delta f = \nabla^2 f$	$\frac{\partial^2 f}{\partial x^2} + \frac{\partial^2 f}{\partial y^2} + \frac{\partial^2 f}{\partial z^2}$	$\frac{1}{\rho} \frac{\partial}{\partial \rho} \left(\rho \frac{\partial f}{\partial \rho} \right) + \frac{1}{\rho^2} \frac{\partial^2 f}{\partial \varphi^2} + \frac{\partial^2 f}{\partial z^2}$	$\frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{\partial f}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial f}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 f}{\partial \varphi^2}$
$\Delta \mathbf{A}$	$\Delta A_x \hat{\mathbf{x}} + \Delta A_y \hat{\mathbf{y}} + \Delta A_z \hat{\mathbf{z}}$	$\begin{aligned} &\left(\Delta A_\rho - \frac{A_\rho}{\rho^2} - \frac{2}{\rho^2} \frac{\partial A_\varphi}{\partial \varphi} \right) \hat{\boldsymbol{\rho}} + \\ &\left(\Delta A_\varphi - \frac{A_\varphi}{\rho^2} + \frac{2}{\rho^2} \frac{\partial A_\rho}{\partial \varphi} \right) \hat{\boldsymbol{\varphi}} + \\ &(\Delta A_z) \hat{\mathbf{z}} \end{aligned}$	$\begin{aligned} &\left(\Delta A_r - \frac{2A_r}{r^2} - \frac{2}{r^2 \sin \theta} \frac{\partial (A_\theta \sin \theta)}{\partial \theta} - \frac{2}{r^2 \sin \theta} \frac{\partial A_\varphi}{\partial \varphi} \right) \hat{\mathbf{r}} + \\ &\left(\Delta A_\theta - \frac{A_\theta}{r^2 \sin^2 \theta} + \frac{2}{r^2} \frac{\partial A_r}{\partial \theta} - \frac{2 \cos \theta}{r^2 \sin^2 \theta} \frac{\partial A_\varphi}{\partial \varphi} \right) \hat{\boldsymbol{\theta}} + \\ &\left(\Delta A_\varphi - \frac{A_\varphi}{r^2 \sin^2 \theta} + \frac{2}{r^2 \sin \theta} \frac{\partial A_r}{\partial \varphi} + \frac{2 \cos \theta}{r^2 \sin^2 \theta} \frac{\partial A_\theta}{\partial \varphi} \right) \hat{\boldsymbol{\varphi}} \end{aligned}$

Appendix 2

□ Focusing. Panofsky-Wenzel theorem

Let's consider the particle transverse momentum change caused by the cavity RF field. The particle moves on the trajectory $z=vt$ parallel to the axis, but is displaced from it by \vec{r}_\perp .

Force acting on the particle is

$$\vec{F}(\vec{r}) = e \left[\vec{E}(\vec{r}) + \vec{v} \times \vec{B}(\vec{r}) \right] e^{i\omega t}.$$

From Maxwell equation $\text{curl} \vec{E}(\vec{r}) = -i\omega \vec{B}(\vec{r})$ one has:

$$\vec{F}(\vec{r}) = e \left[\vec{E}(\vec{r}) + \vec{v} \times \frac{i}{\omega} \text{curl} \vec{E}(\vec{r}) \right] e^{i\omega t} = e \left[\vec{E}(\vec{r}) + \frac{i}{\omega} \vec{\nabla} (\vec{v} \cdot \vec{E}(\vec{r})) - (\vec{v} \cdot \vec{\nabla}) \cdot \vec{E}(\vec{r}) \right] e^{i\omega t}.$$

If $\vec{v} = \vec{i}_z v$, then

$$F_z(\vec{r}) = e E_z(\vec{r}) e^{i\omega t}$$

$$F_\perp(\vec{r}) = e \left[\vec{E}_\perp(\vec{r}) + v \frac{i}{\omega} \left(\vec{\nabla}_\perp E_z(\vec{r}) - \frac{\partial \vec{E}_\perp(\vec{r})}{\partial z} \right) \right] e^{i\omega t}$$

The differential operator $\vec{\nabla}_\perp$ acts on the transverse coordinates \vec{r}_\perp only.

Appendix 2

If the particle velocity and particle transverse coordinates do not change significantly in the cavity, longitudinal and transverse momentum changes are:

$$\Delta p_{\parallel} = e \int_{-\infty}^{\infty} E_z(\vec{r}) e^{i\omega t} dt \Big|_{t=z/v} = \frac{e}{v} \int_{-\infty}^{\infty} E_z(\vec{r}) e^{i\omega z/v} dz;$$

$$\Delta \vec{p}_{\perp} = e \int_{-\infty}^{\infty} \left[\vec{E}_{\perp}(\vec{r}) + \frac{iv}{\omega} \vec{\nabla}_{\perp} E_z(\vec{r}) - \frac{iv}{\omega} \frac{\partial \vec{E}_{\perp}(\vec{r})}{\partial z} \right] e^{i\omega t} dt \Big|_{t=z/v}$$

However,

$$\frac{iv}{\omega} \int_{-\infty}^{\infty} \frac{\partial \vec{E}_{\perp}(\vec{r})}{\partial z} e^{i\omega t} dt \Big|_{t=z/v} = \frac{i}{\omega} \int_{-\infty}^{\infty} \frac{\partial \vec{E}_{\perp}(\vec{r})}{\partial z} e^{i\omega z/v} dz = \frac{1}{v} \int_{-\infty}^{\infty} \vec{E}_{\perp}(\vec{r}) e^{i\omega z/v} dz = \int_{-\infty}^{\infty} \vec{E}_{\perp}(\vec{r}) e^{i\omega t} dt \Big|_{t=z/v}$$

and

$$\Delta \vec{p}_{\perp} = e \frac{iv}{\omega} \int_{-\infty}^{\infty} \vec{\nabla}_{\perp} E_z(\vec{r}) e^{i\omega t} dt \Big|_{t=z/v} = e \frac{i}{\omega} \int_{-\infty}^{\infty} \vec{\nabla}_{\perp} E_z(\vec{r}) e^{i\omega z/v} dz$$

Finally, we have:

$$\Delta \vec{p}_{\perp} = \frac{iv}{\omega} \vec{\nabla}_{\perp} (\Delta p_{\parallel}).$$

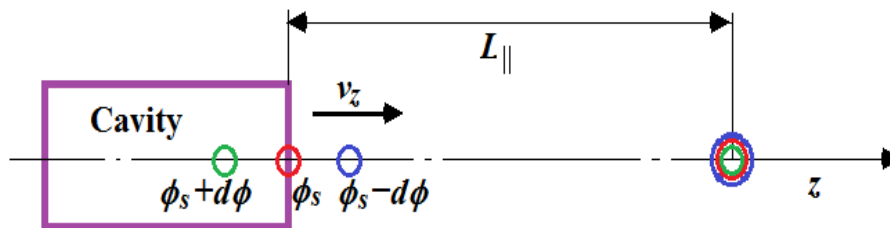
This relation between transverse and longitudinal momentum changes in an RF field is known as Panofsky – Wenzel theorem.

Appendix 2.

□ Bunching

Because a particle velocity depends on its energy, the cavity RF field provides the beam bunching. If the particle in the bunch center has the RF phase of ϕ_s (ϕ_s is a so-called synchronous phase), the particles next to the bunch center will have the same longitudinal coordinate at the same time at the distance L_z from the cavity (see Figure below), if

$$d(v(\phi) \cdot t(\phi))/d\phi|_{\phi=\phi_s} = 0. \quad (1)$$



Here $v(\phi)$ is the particle velocity after it leaves the cavity, and $t(\phi)$ is time necessary for the particle to reach the distance $L_{||}$ – “longitudinal focusing distance”:

$$t(\phi) = L_{||}/v(\phi_s) - (\phi - \phi_s)/\omega. \quad (2)$$

On the other hand, one has (Lecture 8, Slide 23):

$$\Delta W(r) = e \left| V_{0,z} \left(\frac{k}{\beta} \right) \right| I_0 \left(\frac{kr}{\beta\gamma} \right) \cos \phi, \quad (3)$$

Appendix 2.

From (1), (2) and (3) it follows for the particle next to the axis ($kr/\beta\gamma \ll 1$) that

$$L_{||} = \left[\frac{v(\phi)^2}{\omega} \cdot \frac{dW/dv}{dW/d\phi} \right]_{\phi=\phi_s} = - \frac{v(\phi_s)^2}{\omega} \cdot \frac{dW/dv|_{\phi=\phi_s}}{\Delta W_{max}(0) \cdot \sin\phi_s} \quad (4)$$

For a linear accelerator one has

$$\frac{dW}{dv} = \frac{d(mc^2\gamma)}{dv} = mv\gamma^3, \quad (5)$$

where m is the particle mass, and therefore

$$L_{||} = - \frac{(mc^2/e)\beta(\phi_s)^3 \gamma(\phi_s)^3}{k\Delta W_{max}(0) \cdot \sin\phi_s} = - \frac{(mc^2/e)\beta(\phi_s)^3 \gamma(\phi_s)^3}{kV_{max}(0) \cdot \sin\phi_s} \quad (6)$$

For bunching one needs $L_{||} > 0$, and therefore, $\phi_s < 0$. Note that $L_{||} \sim 1/k$. Note that for small energy (and therefore small β) the bunching may be too strong, and low RF frequency is to be used for acceleration.

Appendix 3. Eigen modes properties:

Eigenmodes in a cavity.

$$\begin{aligned} \text{curl} \vec{E} &= -i\omega\mu_0 \vec{H}, \text{curl} \vec{H} = i\omega\varepsilon_0 \vec{E}, \\ \text{curlcurl} \vec{E} - k^2 \vec{E} &= 0, \text{curlcurl} \vec{H} - k^2 \vec{H} = 0, \end{aligned} \quad (1)$$

Here $k^2 = \omega^2 \varepsilon_0 \mu_0$

Boundary conditions: $\vec{E}_t = 0, \vec{H}_n = 0$ or $\vec{n} \times \vec{E} = 0, \vec{n} \cdot \vec{H} = 0$.

Equations (1) has non-trivial solutions only for defined k_m^2 , eigenvalues.

Corresponding solutions $\vec{E}_m(x, y, z)$ and $\vec{H}_m(x, y, z)$ -eigenfunctions. There are infinite number of eigenvalues.

Eigenvalues are **real** and **positive**. From (1) and vector theorem (App.1)

$$\text{div}(\vec{A} \times \vec{B}) = \vec{B} \cdot \text{curl} \vec{A} - \vec{A} \cdot \text{curl} \vec{B} \quad (2)$$

one has:

$$\text{div}(\vec{H}_m^* \times \text{curl} \vec{H}_m) = \text{curl} \vec{H}_m \cdot \text{curl} \vec{H}_m^* - \vec{H}_m^* \cdot \text{curlcurl} \vec{H}_m = |\text{curl} \vec{H}_m|^2 - k_m^2 |\vec{H}_m|^2$$

$$k_m^2 \int_V |\vec{H}_m|^2 dV = \int_V |\text{curl} \vec{H}_m|^2 dV - \oint_S (\vec{H}_m^* \times \text{curl} \vec{H}_m) \cdot \vec{n} dS. \quad \Rightarrow \quad k_m^2 = \frac{\int_V |\text{curl} \vec{H}_m|^2 dV}{\int_V |\vec{H}_m|^2 dV}$$

Appendix 3. Eigen modes properties

Eigenmodes are orthogonal:

$$\text{curlcurl}\vec{E}_m - k_m^2\vec{E}_m = 0, \quad (3)$$

$$\text{curlcurl}\vec{E}_n - k_n^2\vec{E}_n = 0, \quad k_m^2 \neq k_n^2$$

Let's calculate using (2)

$$\begin{aligned} \text{div}(\vec{E}_n \times \text{curl}\vec{E}_m) - \text{div}(\vec{E}_m \times \text{curl}\vec{E}_n) &= \text{curl}\vec{E}_m \cdot \text{curl}\vec{E}_n - \vec{E}_n \text{curlcurl}\vec{E}_m - \\ \text{curl}\vec{E}_n \cdot \text{curl}\vec{E}_m + \vec{E}_m \text{curlcurl}\vec{E}_n &= \vec{E}_m \text{curlcurl}\vec{E}_n - \vec{E}_n \text{curlcurl}\vec{E}_m. \end{aligned}$$

Using (3) we have:

$$\text{div}(\vec{E}_n \times \text{curl}\vec{E}_m) - \text{div}(\vec{E}_m \times \text{curl}\vec{E}_n) = (k_n^2 - k_m^2) \vec{E}_m \cdot \vec{E}_n \rightarrow 0$$

$$(k_n^2 - k_m^2) \int_V \vec{E}_m \cdot \vec{E}_n dV = \oint_S \left(\text{div}(\vec{E}_n \times \text{curl}\vec{E}_m) - \text{div}(\vec{E}_m \times \text{curl}\vec{E}_n) \right) \vec{n} dS.$$

$$(k_n^2 - k_m^2) \int_V \vec{E}_m \cdot \vec{E}_n dV = 0, \quad \leftarrow k_m^2 \neq k_n^2$$

$$\int_V \vec{E}_m \cdot \vec{E}_n dV = 0,$$

$$\int_V \vec{H}_m \cdot \vec{H}_n dV = 0,$$

Appendix 3. Eigen modes properties

Let's consider again the equation we got

$$k_m^2 \int_V |\vec{H}_m|^2 dV = \int_V |\text{curl} \vec{H}_m|^2 dV$$

Taking into account $\text{curl} \vec{H}_m = i\omega_m \epsilon_0 \vec{E}_m$ and $k_m^2 = \omega_m^2 \epsilon_0 \mu_0$, we have:

$$\frac{1}{4} \int_V \mu_0 |\vec{H}_m|^2 dV = \frac{1}{4} \int_V \epsilon_0 |\vec{E}_m|^2 dV$$

The time-average electrical stored energy is equal to the time-average magnetic stored energy.

Appendix 3. Eigen modes properties:

Variation properties of the eigenmodes:

Let's consider expression for k^2 :

$$k^2 = \frac{\int_V |\text{curl}\vec{H}|^2 dV}{\int_V |\vec{H}|^2 dV}. \quad (1)$$

Variation of (1) gives:

$$\delta k^2 \int_V |\vec{H}|^2 dV + 2k^2 \int_V \vec{H} \cdot \delta\vec{H} dV = 2 \int_V \text{curl}\vec{H} \cdot \text{curl}\delta\vec{H} dV$$

Using the vector theorem (2)

$\text{div}(\delta\vec{H} \times \text{curl}\vec{H}) = \text{curl}\vec{H} \cdot \text{curl}\delta\vec{H} - \delta\vec{H} \cdot \text{curlcurl}\vec{H}$, we have

$$\delta k^2 \int_V |\vec{H}|^2 dV = 2 \int_V (\text{curlcurl}\vec{H} - k^2\vec{H}) \cdot \delta\vec{H} dV + 2 \int_V \text{div}(\delta\vec{H} \times \text{curl}\vec{H}) dV.$$

Using Gauss theorem, we get

$$\delta k^2 \int_V |\vec{H}|^2 dV = 2 \int_V (\text{curlcurl}\vec{H} - k^2\vec{H}) \cdot \delta\vec{H} dV + 2 \oint_S (\delta\vec{H} \times \text{curl}\vec{H}) \cdot \vec{n} dS.$$

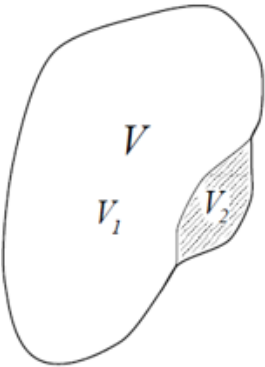
Taking into account that $\text{curlcurl}\vec{H} - k^2\vec{H} = 0$ and boundary condition

$\vec{n} \times \text{curl}\vec{H} = 0$ on S , and we finally get

$$\delta k^2 = 0$$

Appendix 3.

Small perturbations of the cavity geometry:



$$k^2 = \frac{\int_{V_1} |\text{curl} \vec{H}^{(2)}|^2 dV}{\int_{V_1} |\vec{H}^{(2)}|^2 dV} \approx \frac{\int_{V_1} |\text{curl} \vec{H}^{(1)}|^2 dV}{\int_{V_1} |\vec{H}^{(1)}|^2 dV} = \frac{\int_V |\text{curl} \vec{H}^{(1)}|^2 dV}{\int_V |\vec{H}^{(1)}|^2 dV} - \frac{\int_{V_2} |\text{curl} \vec{H}^{(1)}|^2 dV}{\int_{V_2} |\vec{H}^{(1)}|^2 dV} =$$

$$= k_1^2 \left\{ 1 + \frac{\int_{V_2} |\vec{H}^{(1)}|^2 dV}{\int_V |\vec{H}^{(1)}|^2 dV} - \frac{\int_{V_2} |\text{curl} \vec{H}^{(1)}|^2 dV}{\int_V |\text{curl} \vec{H}^{(1)}|^2 dV} \right\} = k_1^2 \left(1 + \frac{\Delta W_E - \Delta W_H}{W_0} \right)$$

$$\frac{k^2 - k_1^2}{k_1^2} = \frac{\Delta W_E - \Delta W_H}{W_0} \approx \frac{2\Delta\omega}{\omega_1} = \frac{(w_E - w_H) \cdot \Delta V}{W_0} \quad \leftarrow \text{Slater theorem}$$

On the other hand, $\frac{1}{2}(w_H - w_E)\Delta V = p \Delta V = -\Delta W_0$ and

$$\frac{\Delta\omega}{\omega_0} = \frac{\Delta W_0}{W_0},$$



$$\boxed{\frac{W_0}{\omega_0} = \text{const}} \quad (\text{compare to } E = \nu h)$$

Appendix 4: Accelerating voltage and transit time factor:

For arbitrary axial distribution of the axisymmetric accelerating field the voltage

$V(\varphi)$ at arbitrary phase φ is the following:

$$\begin{aligned} V(\varphi) &= \operatorname{Re} \int_{-\infty}^{\infty} E_z(\rho = 0, z) e^{i(k_z z + \varphi)} dz = \\ &= \int_{-\infty}^{\infty} \operatorname{Re}[E_z(\rho = 0, z)] \cos(k_z z + \varphi) dz - \int_{-\infty}^{\infty} \operatorname{Im}[E_z(\rho = 0, z)] \sin(k_z z + \varphi) dz = \\ &= \int_{-\infty}^{\infty} \operatorname{Re}[E_z(\rho = 0, z)] \cos(k_z z + \varphi) dz - \int_{-\infty}^{\infty} \operatorname{Im}[E_z(\rho = 0, z)] \sin(k_z z + \varphi) dz = \\ &= \cos(\varphi) \left(\int_{-\infty}^{\infty} \operatorname{Re}[E_z(\rho = 0, z)] \cos(k_z z) dz - \int_{-\infty}^{\infty} \operatorname{Im}[E_z(\rho = 0, z)] \sin(k_z z) dz \right) - \\ &\quad - \sin(\varphi) \left(\int_{-\infty}^{\infty} \operatorname{Re}[E_z(\rho = 0, z)] \sin(k_z z) dz - \int_{-\infty}^{\infty} \operatorname{Im}[E_z(\rho = 0, z)] \cos(k_z z) dz \right) \end{aligned}$$

Maximal voltage V is, therefore,

$$\begin{aligned} V &= \left[\left(\int_{-\infty}^{\infty} \operatorname{Re}[E_z(\rho = 0, z)] \cos(k_z z) dz - \int_{-\infty}^{\infty} \operatorname{Im}[E_z(\rho = 0, z)] \sin(k_z z) dz \right)^2 \right. \\ &\quad \left. + \left(\int_{-\infty}^{\infty} \operatorname{Re}[E_z(\rho = 0, z)] \sin(k_z z) dz - \int_{-\infty}^{\infty} \operatorname{Im}[E_z(\rho = 0, z)] \cos(k_z z) dz \right)^2 \right]^{1/2} \end{aligned}$$

If $E_z(\rho, z)$ is real,
$$V = \left[\left(\int_{-\infty}^{\infty} E_z(\rho = 0, z) \cos(k_z z) dz \right)^2 + \left(\int_{-\infty}^{\infty} E_z(\rho = 0, z) \sin(k_z z) dz \right)^2 \right]^{1/2}$$

Appendix 5. Modes in a pillbox cavity:

In a pillbox cavities resonance field satisfies wave equations:

$$\Delta \vec{E} + k^2 \vec{E} = 0, \Delta \vec{H} + k^2 \vec{H} = 0, \text{ where } k = \frac{\omega}{c} - \text{wavenumber.}$$

For an ideally conductive wall components of electric field tangential to the surface is zero. The pillbox cavity may be considered as a part of a waveguide having circular cross section, shortened at both ends. The fields in this waveguide may be described in cylindrical coordinates, (r, φ, z) . In cylindrical coordinates longitudinal field components satisfy scalar wave equations:

$$\Delta E_z + k^2 E_z = 0, \Delta H_z + k^2 H_z = 0 \quad (1)$$

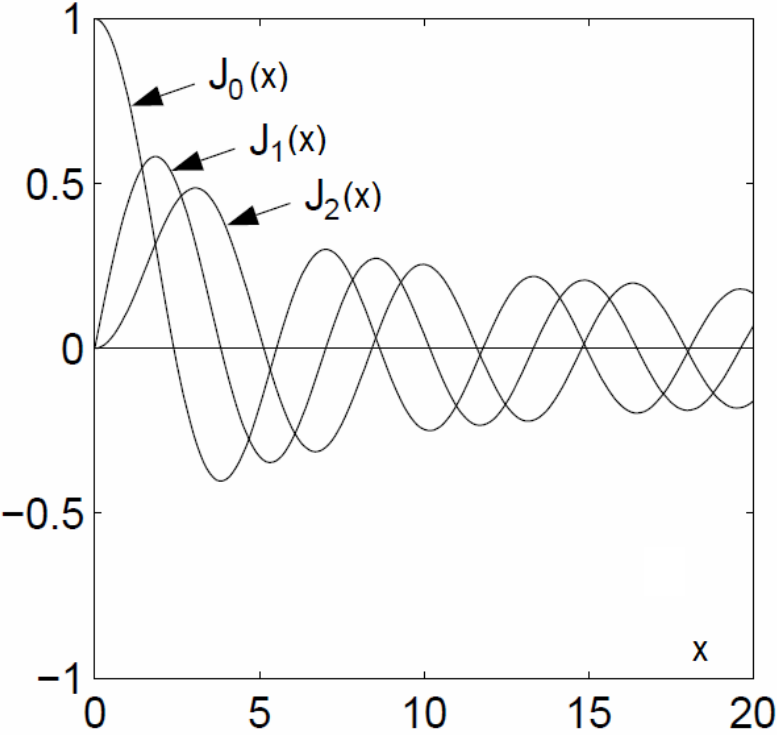
For the waveguide, the fields have translation symmetry along \vec{z} , i.e., in two points having the same transverse coordinate, but different z , the fields differ by phase $\psi = k_z z$; i.e., $\vec{E}, \vec{H} \sim e^{ik_z z}$. In this case:

- Equations (1) have solution

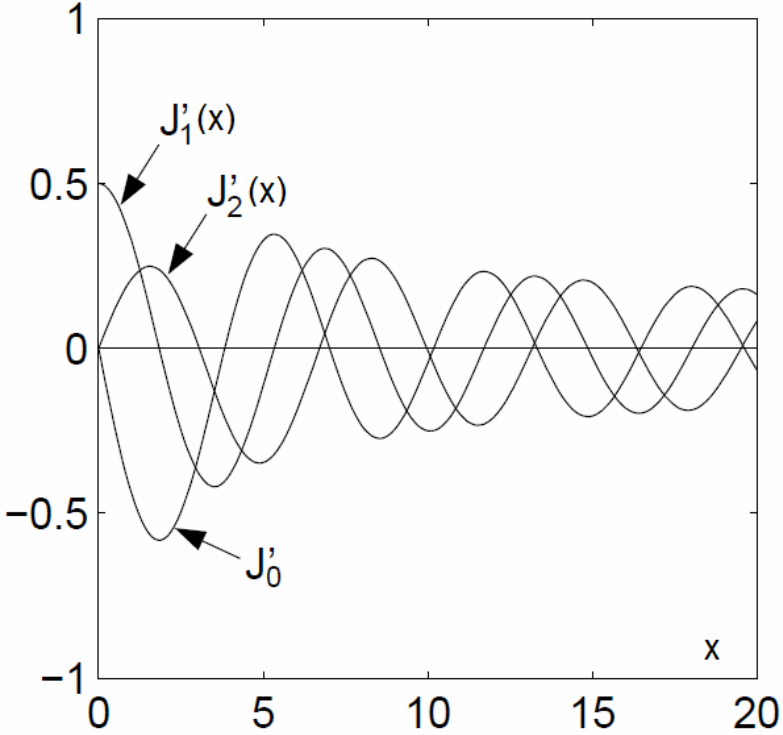
$$E_z(r, \varphi, z), H_z(r, \varphi, z) = J_m(k_r r) e^{im\varphi} e^{ik_z z}; J_m(k_r r) \text{ are Bessel functions;}$$

- $k_r^2 + k_z^2 = k^2$;
- All transverse components (E_r, E_φ, H_r and H_φ) may be expressed through the longitudinal field components, E_z and H_z ;
- At $r=b$ (b is the waveguide radius) $E_z=0$ and $\frac{\partial H_z}{\partial n} = 0$. \vec{n} is normal to the waveguide surface.

Appendix 5.



(a)



(b)

Bessel functions and their derivatives

Appendix 5.

E_z and H_z satisfy the same equation, but have different boundary conditions, and therefore, different k_r :

Electric field:

$$\Delta E_z + k^2 E_z = 0$$

Equation: $E_z(r, \varphi, z) = 0, r = a$;

$$J_m(k_r b) = 0;$$

or
and $k_r = \frac{\nu_{mn}}{b}; J_m(\nu_{mn}) = 0$;

Magnetic field:

$$\Delta H_z + k^2 H_z = 0$$

$$\partial H_z(r, \varphi, z) / \partial r = 0, r = a$$

$$J'_m(k_r b) = 0,$$

$$k_r = \frac{\mu_{mn}}{b}; J'_m(\mu_{mn}) = 0.$$

For the pillbox cavity having end walls at $z=0$ and $z=d$; therefore $k_z d = \pi p$ and

$$E_z = C J_m(k_r r) e^{im\varphi} \cos(\pi p z / d); H_z = J_m(k_r r) e^{im\varphi} \sin(\pi p z / d),$$

and resonant frequencies are:

$$\frac{\omega}{c} = \sqrt{\left(\frac{\nu_{mn}}{b}\right)^2 + \left(\frac{\pi p}{d}\right)^2};$$

$$m = 0, 1, \dots, \infty;$$

$$n = 1, 2, \dots, \infty;$$

$$p = 0, 1, \dots, \infty;$$

TM_{mnp}-modes

$$\frac{\omega}{c} = \sqrt{\left(\frac{\mu_{mn}}{b}\right)^2 + \left(\frac{\pi p}{d}\right)^2}$$

$$m = 0, 1, \dots, \infty;$$

$$n = 1, 2, \dots, \infty;$$

$$p = 1, 2, \dots, \infty;$$

TE_{mnp}-modes

Appendix 5.

Roots of $J'_m(x) = 0$.

m	μ_{m1}	μ_{m2}	μ_{m3}	μ_{m4}
0	3.832	7.016	10.174	13.324
1	1.841	5.331	8.536	11.706
2	3.054	6.706	9.970	13.170
3	4.201	8.015	11.346	14.586
4	5.318	9.282	12.682	15.964
5	6.416	10.520	13.987	17.313

Roots of $J_m(x) = 0$.

m	ν_{m1}	ν_{m2}	ν_{m3}	ν_{m4}
0	2.405	5.520	8.654	11.792
1	3.832	7.016	10.174	13.324
2	5.135	8.417	11.620	14.796
3	6.380	9.761	13.015	16.223
4	7.588	11.065	14.373	17.616
5	8.771	12.339	15.700	18.980

$$\mu_{0n} = \nu_{1n}!$$

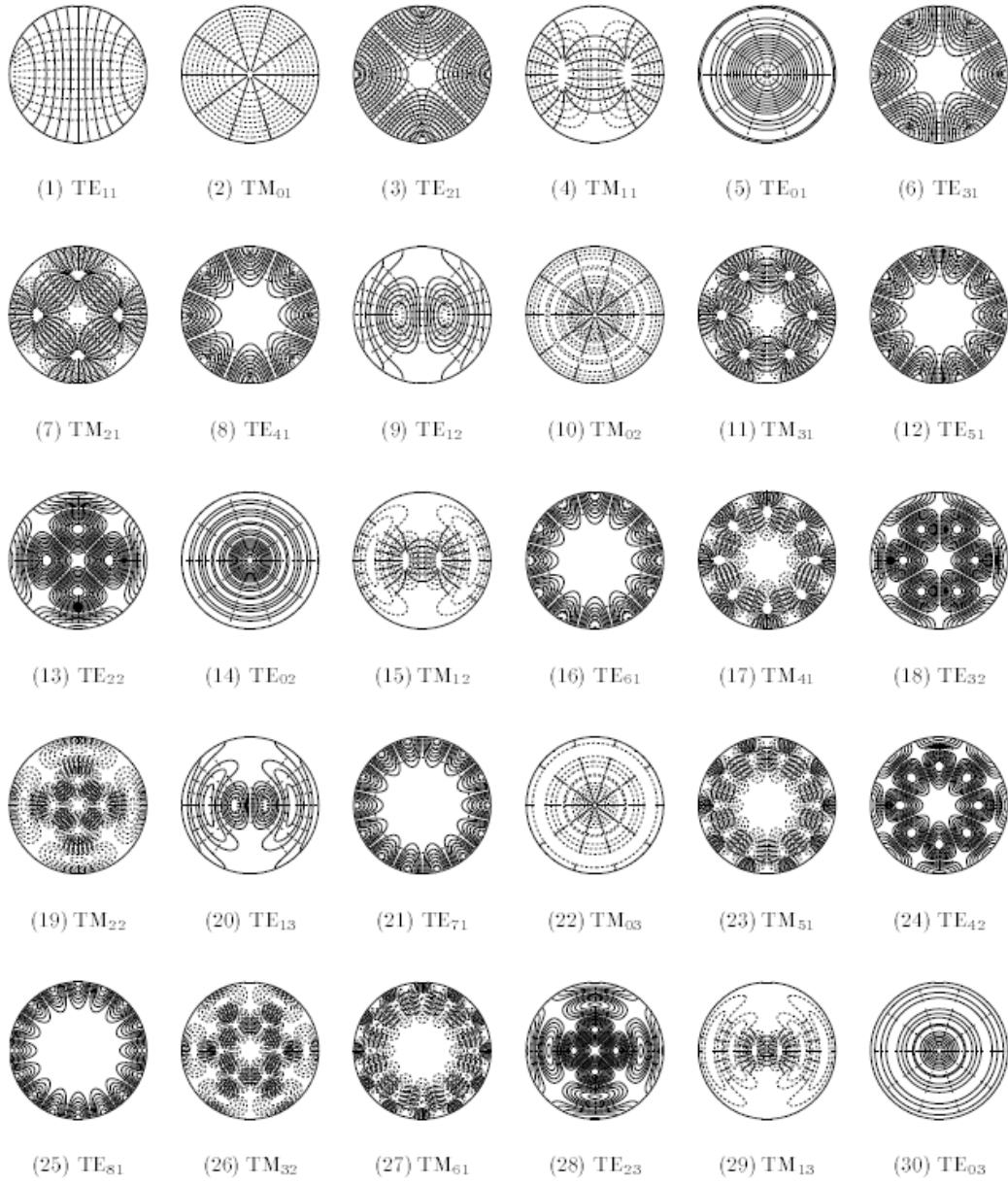
TM_{1np} and TE_{0np}
are degenerated!

$n=1,2,\dots,\infty$;

$p=1,2,\dots,\infty^*$

*Note that TE_{0n0} does not exist because of boundary conditions for magnetic field on the end walls.

Appendix 5.



Field plots for the pillbox modes

Appendix 6: RF cavity excitation by the beam:

- RF cavity having eigenmodes: eigen fields satisfy Maxwell equations:

$$\text{curl} \mathbf{E}_s = -i\omega_s \mu \mathbf{H}_s, \quad \text{curl} \mathbf{H}_s = i\omega_s \varepsilon \mathbf{E}_s. \quad (1)$$

- The field excited by the beam:

$$\text{curl} \mathbf{E} = -i\omega \mu \mathbf{H}, \quad (2)$$

$$\text{curl} \mathbf{H} = i\omega \varepsilon \mathbf{E} + \mathbf{J}_e$$

- The excited field may be expanded over the eigenmodes:

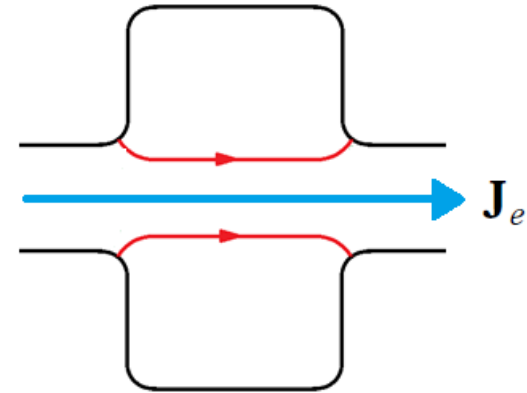
$$\mathbf{E} = \sum A_s \mathbf{E}_s - \text{grad} \varphi_e, \quad \mathbf{H} = \sum B_s \mathbf{H}_s \quad (3)$$

Here φ_e is space charge potential, typically its impact is small.

- From (1) and (2) one has:

$$\begin{aligned} \text{div} (\mathbf{E}_s^* \times \mathbf{H}) &= \mathbf{H} \cdot (i\omega_s \mu \mathbf{H}_s^*) - \mathbf{E}_s^* \cdot (i\omega \varepsilon \mathbf{E} + \mathbf{J}_e) = \\ &= i\omega_s \mu \mathbf{H} \cdot \mathbf{H}_s^* - i\omega \varepsilon \mathbf{E} \cdot \mathbf{E}_s^* - \mathbf{J}_e \cdot \mathbf{E}_s^*, \end{aligned} \quad (4)$$

$$\begin{aligned} \text{div} (\mathbf{E} \times \mathbf{H}_s^*) &= \mathbf{H}_s^* \cdot (-i\omega \mu \mathbf{H}) - \mathbf{E} \cdot (-i\omega_s \varepsilon \mathbf{E}_s^*) = \\ &= -i\omega \mu \mathbf{H} \cdot \mathbf{H}_s^* + i\omega_s \varepsilon \mathbf{E} \cdot \mathbf{E}_s^*. \end{aligned}$$



\mathbf{J}_e – the beam current density spectrum component oscillating at the frequency ω

All the fields have zero tangential electric field components on the wall.

Appendix 6:

Substituting (3) to (4) one has:

$$A_s = -\frac{\omega}{i(\omega^2 - \omega_s^2)} \cdot \frac{\int_V \mathbf{J}_e \mathbf{E}_s^* dV}{\mu \int_V \mathbf{H}_s \cdot \mathbf{H}_s^* dV}, \quad B_s = -\frac{\omega_s}{i(\omega^2 - \omega_s^2)} \cdot \frac{\int_V \mathbf{J}_e \mathbf{E}_s^* dV}{\mu \int_V \mathbf{H}_s \cdot \mathbf{H}_s^* dV},$$

Note that

$$\mu \int_V \mathbf{H}_s \cdot \mathbf{H}_s^* dV = 2W_s$$

If there are wall losses, $\omega_s^2 \rightarrow \omega_s^2 \left(1 + \frac{i}{Q_0}\right)$

and for thin beam having the average current I_0 on the axis one has:

$$A_s = -\frac{i\omega}{\omega^2 - \omega_s^2 - i\frac{\omega\omega_s}{Q_s}} \times \frac{\int_V \mathbf{J}_e \mathbf{E}_s dV}{2W_s} = -\frac{i\omega}{\omega^2 - \omega_s^2 - i\frac{\omega\omega_s}{Q_s}} \times \frac{I_0 \left| \int_{-\infty}^{\infty} E_{sz}(z) e^{ikz} dz \right|}{W_s}, \quad k = \frac{\omega_s}{c} \quad (5)$$

From (5) and (3) one has for the cavity voltage on the axis for the s^{th} mode:

$$V_s = \frac{\Delta p_{\parallel} c}{e} = \int_{-\infty}^{\infty} E_{sz} e^{ikz} dz \approx \frac{i\omega_s^2}{\omega^2 - \omega_s^2 - i\frac{\omega\omega_s}{Q_s}} \cdot \frac{I_0 \left| \int_{-\infty}^{\infty} E_{sz}(z) e^{ikz} dz \right|^2}{\omega_s W_s} = \frac{i\omega_s^2}{\omega^2 - \omega_s^2 - i\frac{\omega\omega_s}{Q_s}} \cdot I_0 \left(\frac{R}{Q} \right)_s$$

Appendix 6:

here

$$\left(\frac{R}{Q}\right)_s = \frac{\left|\int_{-\infty}^{\infty} E_{sz}(z)e^{ikz} dz\right|^2}{\omega_s W_s}$$

At the resonance one has

$$V_s = -Q_s I_0 \left(\frac{R}{Q}\right)_s = -I_0 R_s$$

where

$$R_s = Q_s \left(\frac{R}{Q}\right)_s$$

is a shunt impedance of the s^{th} mode.

Appendix 6:

- This coincides to the voltage excited by the AC current $I = -2I_0$ in a parallel resonance circuit.
- Note that for a short bunch the beam current spectrum is
$$I(\omega) \approx I_0 + 2I_0 \sum \delta(\omega_s),$$
 ω_s is the bunch sequence frequency; the equivalent circuit describes the cavity excitation by imaginary current, it gives the sign “-”.
- From Kirchhoff theorem one has

$$I_C + I_L + I_R = iV\omega C + V/R + V/i\omega L = -2I_0,$$

and taking into account that $\omega_s = (LC)^{-1/2}$, we get

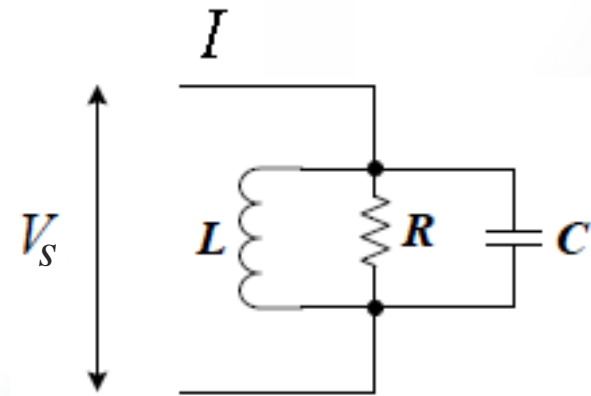
$$V_s = \frac{i\omega_s^2}{\omega^2 - \omega_s^2 - i\frac{\omega\omega_s}{Q_s}} \cdot I_0 \left(\frac{R}{Q} \right)_s,$$

if $\omega_s \approx \omega$. Here are the equivalent circuit parameters:

$$L = (R/Q)_s / 2\omega;$$

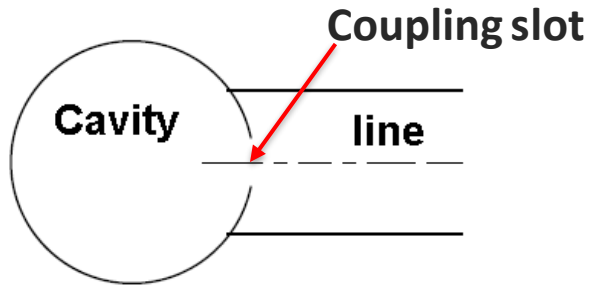
$$C = 2/\omega(R/Q)_s;$$

$$R = (R/Q)_s Q_s / 2.$$

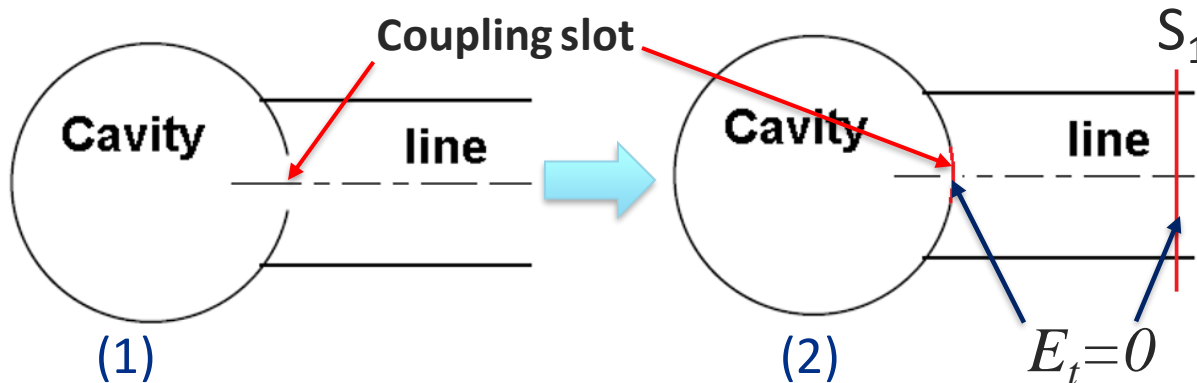


Appendix 7: The cavity coupled to the line.

Let's consider the cavity coupled to the feeding line:



Consider another problem – the cavity coupled to the line shortened by a perfectly conducting plane S_1 placed such a way, that the electric field at the coupling slot has no tangential component:



- The eigenfrequency of the new problem will be the same as for uncoupled cavity;
- The fields inside the cavity will be the same as for uncoupled cavity;
- The magnetic field on S_1 will be proportional to WG magnetic transverse eigenfunction h , $H_t = ikh(x, y)$; k is coefficient (real for convenience). Tangential electric field is zero.

Appendix 7:

For the cavity excited by the line (problem 1) one has on S_1 for transverse fields:

$$E_t = U \cdot \mathbf{e}(x, y), H_t = I \cdot \mathbf{h}(x, y), \quad (1)$$

where $\mathbf{e}(x, y)$ and $\mathbf{h}(x, y)$ are the electric and magnetic WG transverse eigenfunction, $\int_{S_1} (\mathbf{e} \times \mathbf{h}) d\mathbf{S} = 1$.

The field in the cavity \mathbf{H} is proportional to the eigenfunction \mathbf{H}_s of the cavity coupled to the shortened line (see previous slide):

$$\mathbf{H} = B \cdot \mathbf{H}_s = ikB \mathbf{h}(x, y) \quad (2)$$

From (1) and (2) one can find that

$$I = ikB \quad (3)$$

Following the procedure from Appendix 11, we have, see Formulas 1, 2 and 3 from this Appendix 11 and (1-3):

$$B = \frac{i\omega_s}{\omega^2 - \omega_s^2} \cdot \frac{\int_{S_1} (\vec{E} \times \vec{H}_s^*) \cdot d\mathbf{S}}{2W_s} = - \frac{\omega_s}{\omega^2 - \omega_s^2} \cdot \frac{Uk}{2W_s}$$

and

$$I = - \frac{i\omega_s}{\omega^2 - \omega_s^2} \cdot \frac{Uk^2}{2W_s}$$

(4)

Appendix 7:

If there is wall loss in the cavity, $\omega_s^2 \rightarrow \omega_s^2 \left(1 + \frac{i}{Q_0}\right)$ and

$$I = -\frac{i\omega_s}{\omega^2 - \omega_s^2 - \frac{i\omega_s^2}{Q_0}} \cdot \frac{Uk^2}{2W_s}$$

The cavity impedance at S_1 is therefore

$$Z_1 = \frac{U}{I} = -R_1 Q_0 \left(\frac{\omega^2 - \omega_s^2 - \frac{i\omega_s^2}{Q_0}}{i\omega_s^2} \right) \approx R_1 (1 + iQ_0 x),$$

where

$$R_1 = \frac{2\omega_s W_s}{Q_0 k^2}, \quad x = \frac{\omega^2}{\omega_s^2} - 1 \approx \frac{2(\omega - \omega_s)}{\omega_s}$$

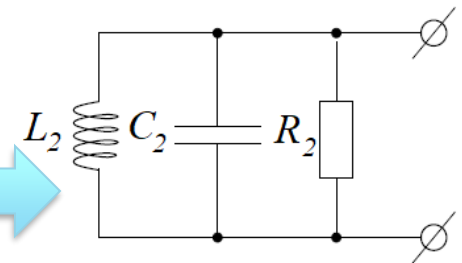
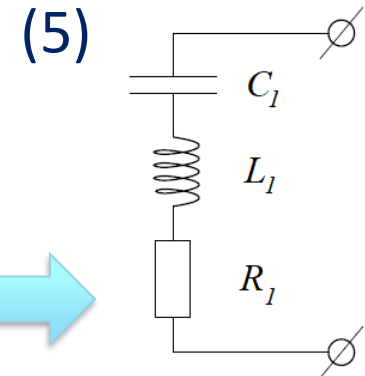
The impedance (5) coincides to the impedance of a serial resonance circuit . At the distance of

$\Lambda/4$ (Λ is wavelength in the WG) the cavity impedance is

$$Z_2 = \frac{Z_0^2}{Z_1} = \frac{Z_0^2}{R_1 (1 + iQ_0 x)} = \frac{R_2}{(1 + iQ_0 x)}, \quad R_2 = \frac{Z_0^2 Q_0 k^2}{2\omega_s W_s} \quad (6)$$

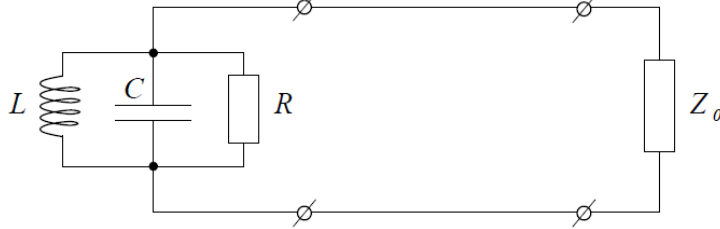
It is the impedance of a parallel resonance circuit

(Z_0 is the WG impedance).



Appendix 7:

If the line is matched, the equivalent circuit is



- Power P_R dissipated in R corresponds to the Ohmic losses in the cavity walls;
- Power P_{Z_0} dissipated in Z_0 corresponds to radiation in the line.

One can see:

$$\frac{P_{Z_0}}{P_R} = \frac{R}{Z_0} = \frac{Q_0}{Q_{ext}},$$

External quality factor, Q_{ext} describes radiation to the line:

$$Q_{ext} = \frac{\omega_s W_s}{P_{Z_0}}$$

Note that $Q_0 = \frac{\omega_s W_s}{P_R}$. The total loss is described by the loaded quality factor, Q_{load}

$$\frac{1}{Q_{load}} = \frac{P_{Z_0} + P_R}{\omega_s W_s} = \frac{1}{Q_{ext}} + \frac{1}{Q_0} \quad (7)$$

Ratio of Q_0 to Q_{ext} is called coupling, β :

$$\frac{Q_0}{Q_{ext}} = \frac{R}{Z_0} \equiv \beta \quad (8)$$

Appendix 7:

Let's estimate the reflection coefficient Γ of the parallel resonance circuit connected to the line:

$$\Gamma = \frac{Z - Z_0}{Z + Z_0},$$

According to (6)

$$Z = \frac{R}{(1 + iQ_0x)}$$

At resonance ($x=0$)

$$\Gamma_0 = \frac{R - Z_0}{R + Z_0} = \frac{\frac{R}{Z_0} - 1}{\frac{R}{Z_0} + 1} = \frac{\beta - 1}{\beta + 1}.$$

For $x \neq 0$ one has

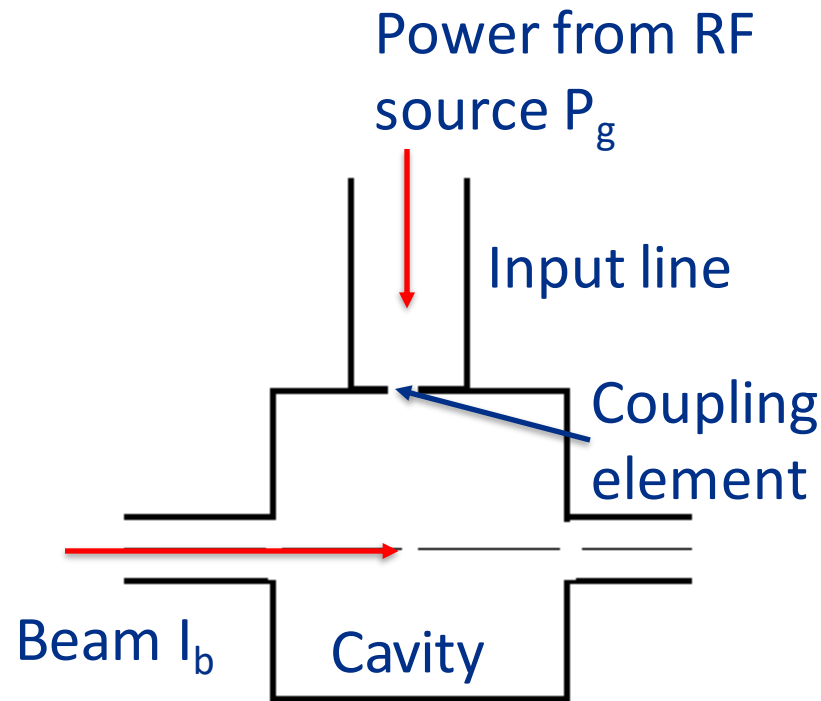
$$\Gamma = \frac{R - Z_0(1 + iQ_0x)}{R + Z_0(1 + iQ_0x)} = \frac{\Gamma_0 + 1}{1 + iQ_0x/(1 + \beta)} - 1 = \frac{\Gamma_0 + 1}{1 + iQ_{load}x} - 1$$

(from (7) and (8) it follows that $Q_0/(1 + \beta) = Q_{load}$).

The power P dissipated in the cavity excited by the input power P_{in} is the following:

$$P = P_{in}(1 - |\Gamma|^2) = P_{in} \frac{1 - \Gamma_0^2}{1 + Q_{load}^2 x^2}.$$

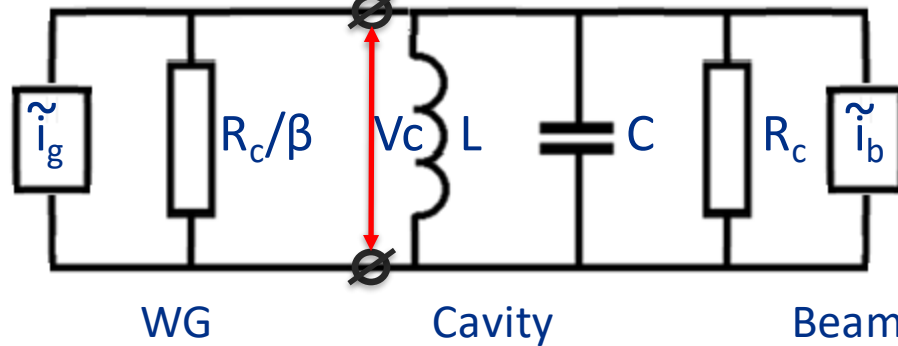
Appendix 8: Beam Loading



- RF source and beam $\omega_g = \omega_b = \omega$;
- Cavity: ω_0
- Cavity voltage : V_c
- Shunt impedance: R_{sh}
- Losses: $P_c = V_c^2 / R_{sh} = V_c^2 / (Q_0 \cdot R / Q)$
- Radiation to the line: $V_c^2 / (Q_{ext} \cdot R / Q)$
- Coupling: $\beta = Q_0 / Q_{ext}$
- Loaded Q: $Q_L = Q_0 / (1 + \beta)$
- Average beam current: I_b
- Synchronous phase: φ
- Power consumed by the beam: $P_b = I_b V_c \cos \varphi$
- Input power P_g
- Reflected power: $P_r = P_g - P_c - P_b$

Appendix 8:

Equivalent circuit for the cavity excited by a WG and loaded by the beam (transformed to the cavity):



$$L = R/Q / (2\omega_0)$$

$$C = 2 / (R/Q \cdot \omega_0)$$

$$R_c = R/Q \cdot Q_0/2$$

$$\tilde{i}_b = -2I_b$$

$$\beta = Q_0/Q_{ext}$$

- The WG impedance transformed to the cavity is $Z_{WG} = R_c/\beta$ (radiated power is $\frac{V_c^2}{R_c/\beta}$);
 - The WG is terminated by the cavity impedance* $Z_0 = \frac{R_c}{1+iQ_0x}$ in parallel to the beam impedance $Z_b = \frac{V_c}{I_b e^{i\varphi}}$, φ is the beam phase versus the voltage V_c .
 - The total load impedance is $Z = \left(\frac{1}{Z_0} + \frac{1}{Z_b} \right)^{-1} = \left(\frac{1+iQ_0x}{R_c} + \frac{I_b e^{i\varphi}}{V_c} \right)^{-1}$.
 - Reflection for this load is $\Gamma = \frac{V_{ref}}{V_{forw}} = \frac{Z - Z_{WG}}{Z + Z_{WG}} = \frac{1 - Z_{WG}/Z}{1 + Z_{WG}/Z}$.
 - The cavity voltage is $V_c = V_{ref} + V_{forw} = V_{forw}(1 + \Gamma) = V_{forw} \frac{2}{1 + Z_{WG}/Z}$ and
- $$V_{forw} = \frac{V_c}{(1+\Gamma)} = \frac{V_c}{2} \left(1 + Z_{WG}/Z \right).$$

*See Formulas (5-6), Appendix 7

Appendix 8:

Therefore, the input power is:

$$P_g = \frac{|V_{forw}|^2}{Z_{WG}} = \frac{V_c^2}{4Z_{WG}} \left| 1 + Z_{WG}/Z \right|^2 =$$
$$\approx \frac{V_c^2(1+\beta)^2}{4\left(\frac{R}{Q}\right)\beta Q_0} \left[\left(1 + \frac{I_b \cos\varphi \left(\frac{R}{Q}\right) Q_0}{(1+\beta)V_c} \right)^2 + \left(\frac{Q_0}{1+\beta} \cdot \frac{2\Delta f}{f} + \frac{I_b \sin\varphi \left(\frac{R}{Q}\right) Q_0}{1+\beta} \right)^2 \right].$$

The formula works next to resonance: approximation is used for x :

$$x = \frac{\omega^2}{\omega_0^2} - 1 \approx \frac{2(\omega - \omega_0)}{\omega_0} = \frac{2(f - f_0)}{f_0} \approx \frac{2(f - f_0)}{f} = \frac{2\Delta f}{f}.$$

Note that $P_g = \frac{|V_{forw}|^2}{Z_{WG}}$ does not contain factor of 2 in the denominator because of the cavity impedance $\left(\frac{R}{Q}\right)$ definition.

Appendix 9: Transverse impedance:

Let's consider a cavity excited by a beam current I_0 having offset x_0 . If \vec{E} is a dipole eigenmode, the field \vec{E} in the cavity in one-mode approximation may be expressed as $\vec{E} \approx A(\omega, \omega_0)\vec{E}$, where ω is the bunch sequence harmonic frequency, ω_0 is the resonant frequency, and

$$A = -\frac{i\omega}{\omega^2 - \omega_0^2 - i\frac{\omega\omega_0}{Q}} \times \frac{\int \vec{j} \cdot \vec{E} dV}{2W} = -\frac{i\omega}{\omega^2 - \omega_0^2 - i\frac{\omega\omega_0}{Q}} \times \frac{I_0 \left| \int_{-\infty}^{\infty} E_z(x, y, z) e^{ikz} dz \right|}{W}.$$

where $W = \frac{\epsilon_0}{2} \int |\vec{E}|^2 dV$ and I_0 is an average current

Appendix 10:

RF-kick at $x=x_0$ and $y=0$ may be obtained using Panofsky-Wenzel theorem:

$$\begin{aligned}
 U_{kick} &= \frac{\Delta p_{\perp} c}{e} = \frac{ic}{\omega} \left| \nabla_{\perp} \int_{-\infty}^{\infty} E_z e^{ikz} dz \right| \approx \frac{c}{x_0 \left(\omega^2 - \omega_0^2 - i \frac{\omega \omega_0}{Q} \right)} \times \frac{I_0 \left| \int_{-\infty}^{\infty} E_z(x_0, 0, z) e^{ikz} dz \right|^2}{W} \\
 &\approx \frac{c \omega_0}{\left(\omega^2 - \omega_0^2 - i \frac{\omega \omega_0}{Q} \right)} \times \frac{x_0 I_0 \left| \int_{-\infty}^{\infty} \left(\frac{\partial E_z(x, 0, z)}{\partial x} \right)_{x=x_0} e^{ikz} dz \right|^2}{W \omega_0} = \frac{\omega_0^2}{\left(\omega^2 - \omega_0^2 - i \frac{\omega \omega_0}{Q} \right)} \times \left(\frac{x_0}{k} \right) I_0 \left(\frac{r_{\parallel}}{Q} \right)
 \end{aligned}$$

where $\left(\frac{r_{\parallel}}{Q} \right) \equiv \frac{\left| \int_{-\infty}^{\infty} \left(\frac{\partial E_z(x, 0, z)}{\partial x} \right)_{x=x_0} e^{ikz} dz \right|^2}{W \omega_0}$

is dipole longitudinal impedance. and $k = \omega_0/c$. **Dipole $\left(\frac{r_{\parallel}}{Q} \right)$ is measured in Ohm/m².**

Appendix 10:

RF-kick may be expressed through the transverse impedance:

$$U_{kick} = \frac{\omega_0^2}{\left(\omega^2 - \omega_0^2 - i\frac{\omega\omega_0}{Q}\right)} \times x_0 I_0 \left(\frac{r_{\perp}}{Q}\right),$$

where $\left(\frac{r_{\perp}}{Q}\right) \equiv \left(\frac{r_{\parallel}}{Q}\right) \times \frac{1}{k} = \frac{\left| \int_{-\infty}^{\infty} \left(\frac{\partial E_z(x, 0, z)}{\partial x}\right)_{x=x_0} e^{ikz} dz \right|^2}{kW\omega_0}$.

Note that $\left(\frac{r_{\perp}}{Q}\right)$ is measured in Ohm/m.

At resonance

$$U_{kick} = i \left(\frac{x_0}{k}\right) I_0 Q \left(\frac{r_{\parallel}}{Q}\right) = ix_0 I_0 Q \left(\frac{r_{\perp}}{Q}\right).$$

Appendix 10:

Sometimes they use other transverse impedance, that is determined as:

$$\left(\frac{r_{\perp}}{Q}\right)_1 = \frac{|U_{kick}|^2}{\omega_0 W_0},$$

where W_0 is the energy, stored in the cavity at the RF field amplitude which provides given transverse kick U_{kick} , $W_0 = |A|^2 W$. At the resonance one has

$$|A|^2 = \frac{\left| x_0 I_0 Q \int_{-\infty}^{\infty} \left(\frac{\partial E_z(x, 0, z)}{\partial x} \right)_{x=x_0} e^{ikz} dz \right|^2}{(W \omega_0)^2} = \left(\frac{r_{\parallel}}{Q}\right) \frac{(x_0 I_0 Q)^2}{W \omega_0}$$

and

$$\omega_0 W_0 = \omega_0 W |A|^2 = \left(\frac{r_{\parallel}}{Q}\right) (x_0 I_0 Q)^2. \text{ On the other hand, } |U_{kick}|^2 = \frac{1}{k^2} \left(x_0 I_0 Q \left(\frac{r_{\parallel}}{Q}\right) \right)^2$$

and

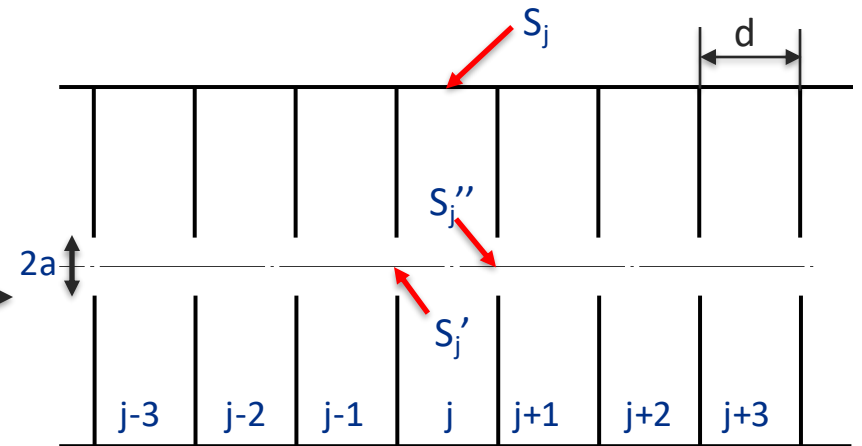
$$\left(\frac{r_{\perp}}{Q}\right)_1 = \frac{|U_{kick}|^2}{\omega_0 W_0} = \left(\frac{r_{\parallel}}{Q}\right) \times \frac{1}{k^2}.$$

$\left(\frac{r_{\perp}}{Q}\right)_1$ is measured in **Ohm**. Note that

$$U_{kick} = i(kx_0) I_0 Q \left(\frac{r_{\perp}}{Q}\right)_1$$

Appendix 11: Travelling–Wave acceleration structures:

- Each previous cell exits EM field in a current cell, which in turn excites the field in the next cell.
- Pillbox cells with thin walls



S_i is the cell metallic surface, S_j' and S_j'' are the coupling holes; V is the cell volume.

\vec{E}_j, \vec{H}_j - fields in the j^{th} pillbox cell oscillate at frequency ω ;

$\vec{E}_{j,n}, \vec{H}_{j,n}$ - eigenmodes in the j^{th} pillbox cell oscillate at frequency ω_n ;

All the fields satisfy Maxwell Equations. Boundary conditions:

$$\vec{E}_j \times n = 0 \text{ on } S_j; \vec{E}_{j,n} \times n = 0 \text{ on } S_j + S_j' + S_j''.$$

Appendix 11: Travelling–Wave acceleration structures.

We consider the following:

$$\begin{aligned} \int_{V_j} \vec{\nabla} \cdot (\vec{E}_j \times \vec{H}_{j,n}^*) dV &= \int_{V_j} [\vec{H}_{j,n}^* \cdot (\vec{\nabla} \times \vec{E}_j) - \vec{E}_j \cdot (\vec{\nabla} \times \vec{H}_{j,n}^*)] dV \\ &= \int_{V_j} [-i\omega\mu_0 \vec{H}_j \cdot \vec{H}_{j,n}^* - i\omega_n \epsilon_0 \cdot \vec{E}_{j,n}^* \cdot \vec{E}_j] dV \end{aligned}$$

Using Gauss theorem and boundary conditions we have

$$\omega_n \epsilon_0 \int_{V_j} \vec{E}_{j,n}^* \cdot \vec{E}_j dV - \omega \mu_0 \int_{V_j} \vec{H}_j \cdot \vec{H}_{j,n}^* dV = \frac{1}{i} \int_{S'_j + S''_j} (\vec{E}_j \times \vec{H}_{j,n}^*) \cdot d\vec{S} \quad (1)$$

Similarly, by considering

$$\int_{V_j} \vec{\nabla} \cdot (\vec{H}_j \times \vec{E}_{j,n}^*) dV$$

we have

$$\omega \epsilon_0 \int_{V_j} \vec{E}_{j,n}^* \cdot \vec{E}_j dV - \omega_n \mu_0 \int_{V_j} \vec{H}_j \cdot \vec{H}_{j,n}^* dV = 0 \quad (2)$$

The eigenmode expansion:

$$\begin{aligned} \vec{E}_j &= \sum X_{j,n} \vec{E}_{j,n} \\ \vec{H}_j &= \sum Y_{j,n} \vec{H}_{j,n} \end{aligned}$$

Appendix 11: Travelling–Wave acceleration structures:

By using (1) and (2) we obtain

$$X_{j,n} = -\frac{i\omega_n}{\omega^2 - \omega_n^2} \cdot \frac{\int_{S'_j+S''_j} (\vec{E}_j \times \vec{H}_{j,n}^*) \cdot dS}{2W_{j,n}} \quad (3)$$

$$Y_{j,n} = \frac{i\omega}{\omega^2 - \omega_n^2} \cdot \frac{\int_{S'_j+S''_j} (\vec{E}_j \times \vec{H}_{j,n}^*) \cdot dS}{2W_{j,n}}$$

The eigenmode amplitudes are determined by tangential electric field on the holes, E_{rj}

How to find E_{rj} for small holes?

1. Quasi-static approximation:

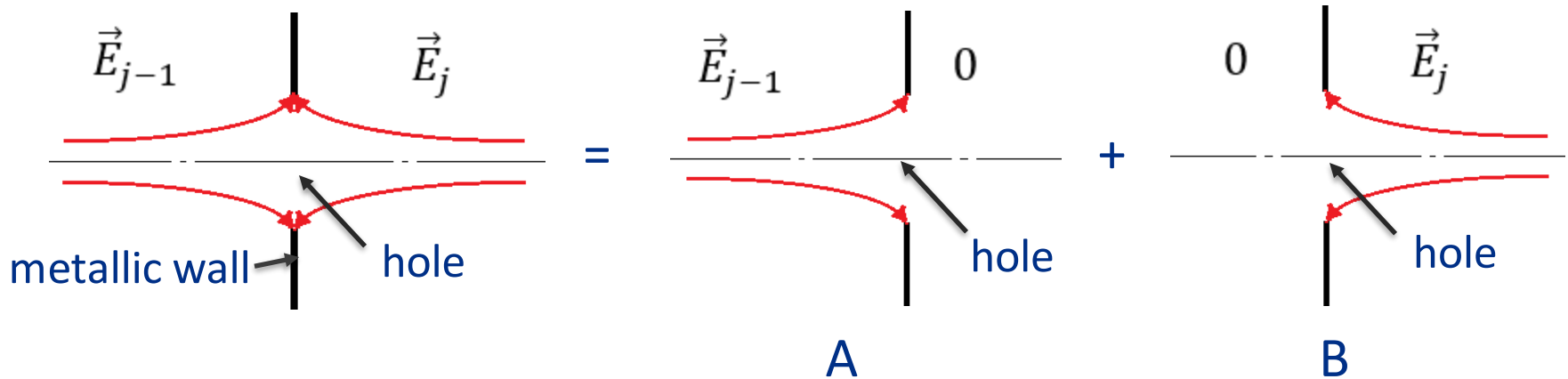
We have: $\Delta \vec{E}_j + k^2 \vec{E}_j = 0$, $k^2 = \frac{\omega^2}{c^2}$

For a small hole, $a \ll \lambda_n$. It means that $\Delta \vec{E}_j \sim \frac{\vec{E}_j}{a^2} \gg k^2 \vec{E}_j = \left(\frac{\omega}{c}\right)^2 \vec{E}_j$ and $\Delta \vec{E}_j \approx 0$
i.e., it means that $\vec{E}_j = \vec{\nabla} \Phi$, $\Delta \Phi = 0$ or electric field is quasi-static.

Far from the holes electric field has only longitudinal (accelerating) component!

Appendix 11: Travelling–Wave acceleration structures:

2. Superposition:



3. Electrostatic problem: conducting sheet at $z=0$ with a circular hole of the radius a , at $z=-\infty$ the field is homogeneous, $E_z=E_0$; at $z=\infty$ the field $E=0$: problem A above. The problem has analytical solution.*

We need to define the radial electric field at the hole, or at $z=0$.

* W.R. Smythe, Static and Dynamic Electricity, 1939, p. 159

Appendix 11: Travelling-Wave acceleration structures:

Oblate spheroidal coordinates:

$$\frac{z^2}{\zeta^2} + \frac{r^2}{1 + \zeta^2} = a^2 \quad \text{spheroids}$$

$$-\frac{z^2}{\xi^2} + \frac{r^2}{1 - \xi^2} = a^2 \quad \text{hyperboloids}$$

$$r = a[(1 + \zeta^2)(1 - \xi^2)]^{1/2}$$

$$z = a\zeta\xi$$

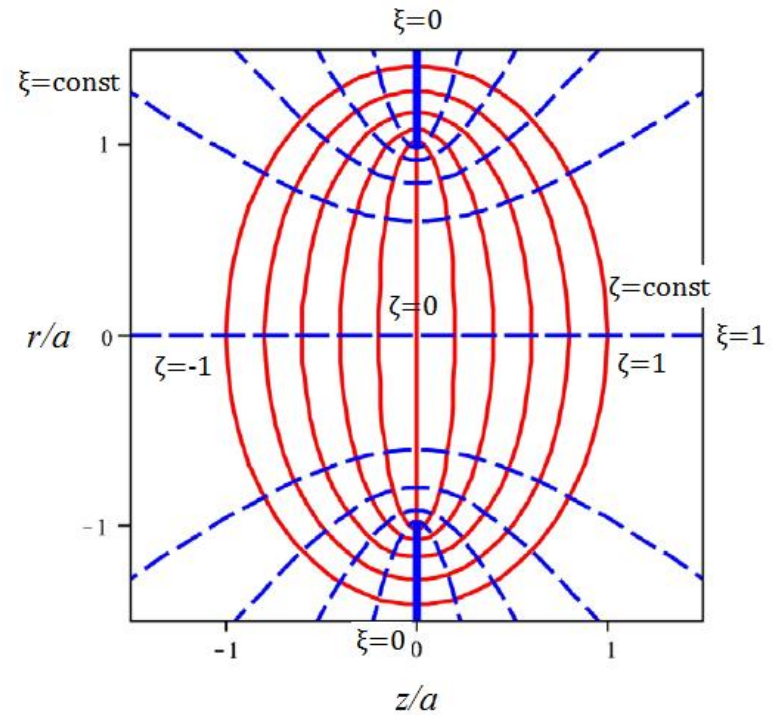
$$\Delta\Phi = 0$$



$$\Phi(\xi, \zeta) = aE_0\xi \left[\zeta - \frac{1}{\pi}(\zeta \cot^{-1}\zeta - 1) \right]$$

$$E_r(r, 0) = \frac{1}{h_1} \nabla_\xi(\Phi(\xi, \zeta)|_{\zeta=0}) = E_0 \frac{(1 - \xi^2)^{1/2}}{\pi\xi} = E_0 \frac{r}{\pi(a^2 - r^2)^{1/2}}$$

$$E_z(r, 0) = \frac{1}{h_2} \nabla_\zeta(\Phi(\xi, \zeta)|_{\zeta=0}) = \frac{1}{2}E_0$$



Lamet coefficients

$$h_1 = a \left(\frac{\xi^2 - \zeta^2}{1 - \xi^2} \right)^{1/2}$$

$$h_2 = a \left(\frac{\xi^2 - \zeta^2}{1 + \zeta^2} \right)^{1/2}$$

Appendix 11: Travelling–Wave acceleration structures:

For TM_{010} mode in a pillbox near the axis (see slide 110):

$$E_{j,r}(r)|_{S'_j} = E_0 \frac{r}{\pi(a^2 - r^2)^{1/2}} [X_{j-1} - X_j]$$

$$E_{j,r}(r)|_{S''_j} = E_0 \frac{r}{\pi(a^2 - r^2)^{1/2}} [X_j - X_{j+1}]$$

$$H_{j,\varphi}(r) \approx -iE_0 \frac{(\frac{\omega_0}{c})r}{2Z_0}$$

and from (3), slide 108, we have:

$$X_j \left[1 - (1 + K) \frac{\omega_0^2}{\omega^2} \right] + \frac{1}{2} K \frac{\omega_0^2}{\omega^2} [X_{j-1} + X_{j+1}] = 0 \quad (4)$$

where K is the coupling, dimensionless parameter:

$$K = \frac{2E_0^2 a^3}{3Z_0 W_0 c} = \frac{2}{3} \cdot \frac{R/Q}{Z_0} \cdot \frac{k_0 a^3}{d^2 T^2} \quad k_0 = \frac{\omega_0}{c}$$

$K \sim a^3$. For a thick wall $K \sim a^\eta$, $\eta > 3$ (field decays in the coupling hole).

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Theory of Diffraction by Small Holes

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(Received January 26, 1942)

Appendix 11: Travelling–Wave acceleration structures:

In the infinite chain of cavities equation (4) has solution (travelling wave):

$$X_j = X e^{ij\varphi}$$

and

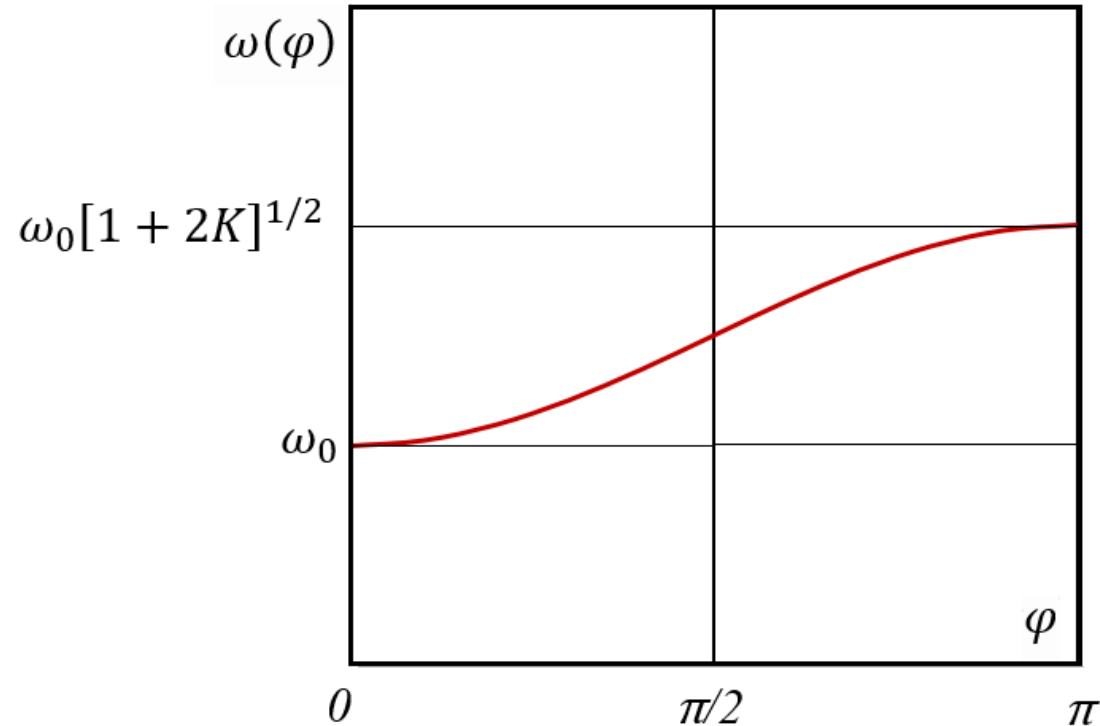
$$\omega(\varphi) = \omega_0 [1 + K(1 - \cos\varphi)]^{1/2}$$

For small K we have:

$$\omega(\varphi) \approx \omega_0 \left[1 + \frac{1}{2} K (1 - \cos\varphi) \right]$$

One can see that

$$K = \frac{\omega(\pi) - \omega(0)}{\omega(0)}$$



Appendix 11: Travelling–Wave acceleration structures.

The 2d Bell theorem, illustration:

For a pillbox structure:

The fields on the hole are equal to

$$E_{j,r}(r)|_{S'_j} = E_0 \frac{r}{\pi(a^2 - r^2)^{1/2}} [X_{j-1} - X_j]$$

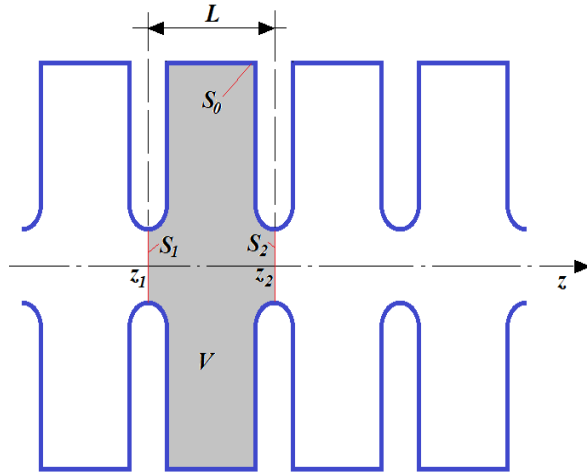
$$H_{j,\phi}(r)|_{S'_j} \approx -iE_0 \frac{\left(\frac{\omega_0}{c}\right) r}{4Z_0} [X_{j-1} + X_j]$$

(E_z and H_ϕ on the hole are two times smaller than in the cell center, see slides 110-111).

Therefore, we have

$$\begin{aligned} v_{gr} &= \frac{P}{\frac{|X_j|^2 W_0}{d}} = \frac{\frac{1}{2} \text{Re} \int E_r H_\phi^* dS}{\frac{|X_j|^2 W_0}{d}} = \frac{\omega_0 E_0^2 a^3 d}{6Z_0 W_0 c} \text{Re}[i(X_{j-1} - X_j)(X_{j-1}^* + X_j^*)] = \\ &= \frac{\omega_0 E_0^2 a^3 d}{3Z_0 W_0 c} \sin(\varphi) = \frac{\omega_0 d}{2} K \sin(\varphi) = c \frac{\pi K d}{\lambda} \sin(\varphi) = \frac{d\omega}{dk_z} \end{aligned}$$

Bell theorems for periodic acceleration structures, proof



I. Floquet Theorem

“For a given mode of propagation in a periodic system at given steady-state frequency the fields at one cross section differ from that one period away only by complex constant”.

$$\begin{aligned}\vec{E}(x, y, z_2) &= \vec{E}(x, y, z_1)e^{-ik_z L} \\ \vec{H}(x, y, z_2) &= \vec{H}(x, y, z_1)e^{-ik_z L} \\ z_2 &= z_1 + L\end{aligned}\quad (1)$$

II. 1st Bell Theorem

“The time-average electrical stored energy per period is equal to the time-average magnetic stored energy per period in the passband”.

Consider the periodic structure to be divided by a series of surfaces perpendicular to the axis spaced by the periodic distance L . One cell of the structure having the volume V is surrounded by these surfaces S_1 and S_2 and the ideal metal boundary S_0 . Let’s consider the integral over the surface surrounding the cell, which equals to zero:

$$\oint \vec{E} \times \vec{H}^* \cdot d\vec{s} = \int_{S_0} \vec{E} \times \vec{H}^* \cdot d\vec{s} + \int_{S_1} \vec{E} \times \vec{H}^* \cdot d\vec{s} + \int_{S_2} \vec{E} \times \vec{H}^* \cdot d\vec{s} = 0 \quad (2)$$

This is because we have

$$\int_{S_0} \vec{E} \times \vec{H}^* \cdot d\vec{s} = 0 \text{ because } E_t = 0 \text{ on } S_0$$

$$\text{and } \int_{S_1} \vec{E} \times \vec{H}^* \cdot d\vec{s} + \int_{S_2} \vec{E} \times \vec{H}^* \cdot d\vec{s} = 0 \text{ because of (1) and } d\vec{s}_2 = -d\vec{s}_1$$

Bell theorems for periodic acceleration structures, proof

From Maxwell equations

$$\text{curl} \vec{E} = -i\omega\mu_0 \vec{H} \quad (3)$$

$$\text{curl} \vec{E}^* = i\omega\mu_0 \vec{H}^*$$

and (2) we have

$$\oint \vec{E} \times \text{curl} \vec{E}^* \cdot d\vec{s} = 0$$

Applying Gauss's theorem, we get:

$$\int_V \text{div}(\vec{E} \times \text{curl} \vec{E}^*) dv = 0. \quad (4)$$

Using the vector theorem (see Appendix 1)

$$\text{div}(\vec{A} \times \vec{B}) = \vec{B} \cdot \text{curl} \vec{A} - \vec{A} \cdot \text{curl} \vec{B}, \quad (5)$$

we get:

$$\int_V (\text{curl} \vec{E}^*) \cdot (\text{curl} \vec{E}) dv - \int_V \vec{E} \cdot (\text{curl}(\text{curl} \vec{E}^*)) dv = 0.$$

Using Maxwell's equations (3) and the homogenous wave equation derived therefrom (see Lecture 10)

$$\text{curl}(\text{curl} \vec{E}^*) = \omega^2 \mu_0 \varepsilon_0 \vec{E}^* \quad (6)$$

we get

$$\int_V (-i\omega\mu_0 \vec{H}) \cdot (i\omega\mu_0 \vec{H}^*) dv - \int_V \vec{E} \cdot (\omega^2 \mu_0 \varepsilon_0 \vec{E}^*) dv = 0$$

Dividing through $4\omega^2 \mu_0$ yields:

$$\frac{1}{4} \int_V \mu_0 |\vec{H}|^2 dv = \frac{1}{4} \int_V \varepsilon_0 |E|^2 dv = W/2,$$

quod erat demonstrandum.

Here W is total energy of electromagnetic field per period,

$$W = \frac{1}{2} \int_V \varepsilon_0 \vec{E} \cdot \vec{E}^* dv = \frac{1}{2} \int_V \mu_0 \vec{H} \cdot \vec{H}^* dv$$

Bell theorems for periodic acceleration structures, proof

III. 2^d Bell Theorem

“The time-average power flow in the pass band is equal to the group velocity times time-average electro-magnetic stored energy per period divided by the period.”

Consider (4) wherein \vec{E} and \vec{E}^* are functions of frequency ω . Differentiate with respect to frequency:

$$\frac{\partial}{\partial \omega} \int_V \operatorname{div}(\vec{E} \times \operatorname{curl} \vec{E}^*) dv = 0.$$

It gives:

$$\int_V \operatorname{div} \left(\frac{\partial \vec{E}}{\partial \omega} \times \operatorname{curl} \vec{E}^* \right) dv + \int_V \operatorname{div} \left(\vec{E} \times \operatorname{curl} \frac{\partial \vec{E}^*}{\partial \omega} \right) dv = 0.$$

Using the vector identity equation (5), we get

$$\begin{aligned} & \int_V (\operatorname{curl} \vec{E}^*) \cdot \left(\operatorname{curl} \frac{\partial \vec{E}}{\partial \omega} \right) dv - \int_V \frac{\partial \vec{E}}{\partial \omega} \cdot (\operatorname{curl}(\operatorname{curl} \vec{E}^*)) dv + \\ & + \int_V (\operatorname{curl} \vec{E}) \cdot \left(\operatorname{curl} \frac{\partial \vec{E}^*}{\partial \omega} \right) dv - \int_V \vec{E} \cdot \left(\operatorname{curl}(\operatorname{curl} \frac{\partial \vec{E}^*}{\partial \omega}) \right) dv = 0 \end{aligned} \quad (7)$$

Differentiation (6) with respect to ω gives

$$\operatorname{curl}(\operatorname{curl} \frac{\partial \vec{E}^*}{\partial \omega}) = 2\omega \mu_0 \epsilon_0 \vec{E}^* + \omega^2 \mu_0 \epsilon_0 \frac{\partial \vec{E}^*}{\partial \omega}$$

Using this in the second and fourth integrals of (7)

$$\begin{aligned} & \int_V (\operatorname{curl} \vec{E}^*) \cdot \left(\operatorname{curl} \frac{\partial \vec{E}}{\partial \omega} \right) dv - \omega^2 \mu_0 \epsilon_0 \int_V \frac{\partial \vec{E}}{\partial \omega} \cdot \vec{E}^* dv + \\ & + \int_V (\operatorname{curl} \vec{E}) \cdot \left(\operatorname{curl} \frac{\partial \vec{E}^*}{\partial \omega} \right) dv - 2\omega \mu_0 \epsilon_0 \int_V \vec{E} \cdot \vec{E}^* dv - \omega^2 \mu_0 \epsilon_0 \int_V \frac{\partial \vec{E}^*}{\partial \omega} \cdot \vec{E} dv = 0 \end{aligned}$$

which is

$$2\operatorname{Re} \left\{ \int_V (\operatorname{curl} \vec{E}^*) \cdot \left(\operatorname{curl} \frac{\partial \vec{E}}{\partial \omega} \right) dv - \int_V \frac{\partial \vec{E}}{\partial \omega} \cdot (\operatorname{curl}(\operatorname{curl} \vec{E}^*)) dv \right\} - 2\omega \mu_0 \epsilon_0 \int_V \vec{E} \cdot \vec{E}^* dv = 0.$$

Bell theorems for periodic acceleration structures, proof

Using (5) in reverse,

$$2\text{Re} \left\{ \int_V \text{div} \left(\frac{\partial \vec{E}}{\partial \omega} \times \text{curl} \vec{E}^* \right) dv \right\} - 2\omega \mu_0 \epsilon_0 \int_V \vec{E} \cdot \vec{E}^* dv = 0.$$

and using Gauss's theorem on the first term,

$$2\text{Re} \left\{ \oint \frac{\partial \vec{E}}{\partial \omega} \times \text{curl} \vec{E}^* \cdot d\vec{s} \right\} - 2\omega \mu_0 \epsilon_0 \int_V \vec{E} \cdot \vec{E}^* dv = 0. \quad (8)$$

Integral $\int_{S_0} \frac{\partial \vec{E}}{\partial \omega} \times \text{curl} \vec{E}^* \cdot d\vec{s} = 0$ because of boundary conditions on the ideal metal surface.

From Floquet theorem (1) the following relation hold:

$$\begin{aligned} \vec{E}(x, y, z_2) &= \vec{E}(x, y, z_1) e^{-ik_z L} \\ \text{curl} \vec{E}^*(x, y, z)|_{z=z_2} &= \text{curl} \vec{E}^*(x, y, z)|_{z=z_1} e^{ik_z L} \end{aligned} \quad (9)$$

$$\vec{E}^*(x, y, z_2) = \vec{E}^*(x, y, z_1) e^{ik_z L}$$

$$\frac{\partial \vec{E}(x, y, z_2)}{\partial \omega} = \frac{\partial \vec{E}(x, y, z_1)}{\partial \omega} e^{-ik_z L} - iL \frac{dk_z}{d\omega} \vec{E}(x, y, z_1) e^{-ik_z L}$$

Separating the surface integral of (8)

$$2\text{Re} \left\{ \int_{S_1} \frac{\partial \vec{E}(x, y, z_1)}{\partial \omega} \times \text{curl} \vec{E}^*(x, y, z)|_{z=z_1} \cdot d\vec{s} + \int_{S_2} \frac{\partial \vec{E}(x, y, z_2)}{\partial \omega} \times \text{curl} \vec{E}^*(x, y, z)|_{z=z_2} \cdot d\vec{s} \right\} -$$

$$2\omega \mu_0 \epsilon_0 \int_V \vec{E} \cdot \vec{E}^* dv = 0$$

and substituting equations (9)

$$2\text{Re} \left\{ \int_{S_1} \frac{\partial \vec{E}(x, y, z_1)}{\partial \omega} \times \text{curl} \vec{E}^*(x, y, z)|_{z=z_1} \cdot d\vec{s} + \right.$$

$$\left. \int_{S_2} \frac{\partial \vec{E}(x, y, z_1)}{\partial \omega} \times \text{curl} \vec{E}^*(x, y, z)|_{z=z_1} \cdot d\vec{s} - iL \frac{dk_z}{d\omega} \int_{S_2} \vec{E}(x, y, z_1) \times \text{curl} \vec{E}^*(x, y, z)|_{z=z_1} \cdot d\vec{s} \right\} -$$

$$2\omega \mu_0 \epsilon_0 \int_V \vec{E} \cdot \vec{E}^* dv = 0$$

Bell theorems for periodic acceleration structures, proof

Since $d\vec{s}_2 = -d\vec{s}_1$ the first two integrals cancel. Using Maxwell equation (3) and condition s (9) we get

$$2\text{Re} \left\{ \omega \mu_0 L \frac{dk_z}{d\omega} \int_{S_2} \vec{E}(x, y, z_2) \times \vec{H}^*(x, y, z_2) \cdot d\vec{s} \right\} - 2\omega \mu_0 \varepsilon_0 \int_V \vec{E} \cdot \vec{E}^* dv = 0$$

Multiplying by $\frac{1}{4\omega \mu_0 L} \frac{d\omega}{dk_z}$ we finally have

$$\frac{1}{2} \text{Re} \left\{ \int_{S_2} \vec{E}(x, y, z_2) \times \vec{H}^*(x, y, z_2) \cdot d\vec{s} \right\} = \frac{d\omega}{dk_z} \cdot \frac{1}{L} \cdot \frac{1}{2} \int_V \varepsilon_0 \vec{E} \cdot \vec{E}^* dv = 0$$

or

$$P = v_{gr} w,$$

here

$P = \frac{1}{2} \text{Re} \left\{ \int_{S_2} \vec{E}(x, y, z_2) \times \vec{H}^*(x, y, z_2) \cdot d\vec{s} \right\}$ is the time averaged power flow in the passbands;

$v_{gr} = \frac{d\omega}{dk_z}$ is a group velocity;

$w = \frac{1}{L} \cdot \frac{1}{2} \int_V \varepsilon_0 \vec{E} \cdot \vec{E}^* dv = \frac{1}{L} \cdot \frac{1}{2} \int_V \mu_0 \vec{H} \cdot \vec{H}^* dv = \frac{W}{L}$ is the time-averaged stored electromagnetic energy per unit length.

[1] J.S. Bell, "Group velocity and energy velocity in periodic waveguides," Harwell, AERE-T-R-858 (1952)

[2] D.A. Watkins, Topics in Electromagnetic Theory, John Willey & Sons, Inc. London, 1958

[3] E. A. Burshtein, and G. B. Voskresensky, The Intensive Beam Electron Linear Accelerators, Atomizdat, Moscow, 1970.

Appendix 12: Standing –Wave acceleration structures.

Perturbation theory.

In matrix form Eq(1), see Lecture 8, Slide 31:

$$M\hat{X} - \frac{\omega_0^2}{\omega^2}\hat{X} = 0$$

here $M_{jj} = 1; j = 0, 1, \dots, N;$

$$M_{jj-1} = \frac{K}{2W(j)}; j = 1, 2, \dots, N; M_{jj+1} = \frac{K}{2W(j)}; j = 0, 1, \dots, N - 1.$$

and $W(j) = 1, j = 1, 2, \dots, N - 1$ $W(j) = \frac{1}{2}, j = 0, N$

Eigenvectors and eigenvalues:

$$\hat{X}_j^q = \cos \frac{\pi q j}{N}; \omega_q^2 = \frac{\omega_0^2}{1 + K \cos \frac{\pi q}{N}}, q = 0, 1, \dots, N$$

Orthogonality:

$$\hat{X}^q \cdot \hat{X}^r \equiv \sum_{j=0}^N W(j) \hat{X}_j^q \hat{X}_j^r = \frac{N \delta_{qr}}{2W(q)}, \delta_{qq} = 1, \text{ and } \delta_{qr} = 0, \text{ if } q \neq r$$

Appendix 12: Standing –Wave acceleration structures.

Perturbation theory

- Perturbation of the cell resonance frequencies causes perturbation of the mode resonance frequencies $\delta\omega_q$;
- the field distribution $\delta\hat{X}^q$.

$$\omega_{0j}^{2'} = \omega_0^2 + \delta\omega_{0j}^2 \rightarrow \hat{X}^{q'} = \hat{X}^q + \delta\hat{X}^q, \quad \hat{X}^q \cdot \delta\hat{X}^q$$

Variation of the equation (1), see previous

slide, gives $M\delta\hat{X}^q = \frac{\omega_0^2}{\omega_q^2} \left[\delta\hat{X}^q + \Omega\hat{X}^q - \frac{\delta\omega_q^2}{\omega_q^2} \hat{X}^q \right]$, where $\Omega = \begin{bmatrix} \frac{\delta\omega_{01}^2}{\omega_0^2} & \dots & 0 \\ \vdots & \ddots & \vdots \\ 0 & \dots & \frac{\delta\omega_{0N}^2}{\omega_0^2} \end{bmatrix}$

$$\frac{\delta\omega_q^2}{\omega_q^2} = [2W(q)/N] \cdot \hat{X}^q \Omega \hat{X}^q;$$

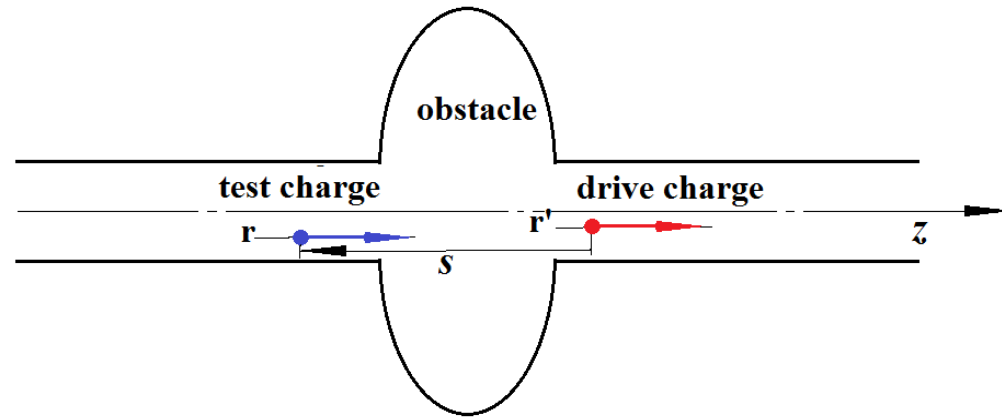
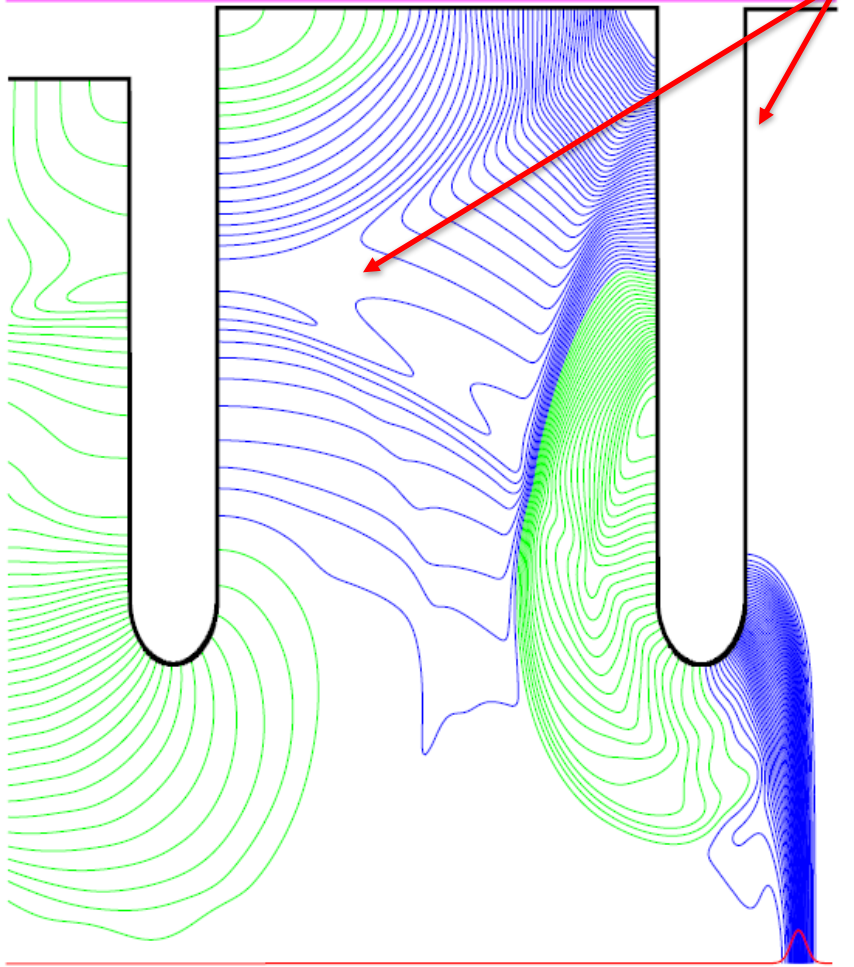
$$\delta\hat{X}^q = \sum_{q' \neq q} \frac{2W(q') \hat{X}^q \Omega \hat{X}^q}{N \left(\frac{\omega_q^2}{\omega_{q'}^2} - 1 \right)} \hat{X}^{q'}$$

$$|\delta\hat{X}^q| \sim \frac{|\delta\omega_{0j}|_{av}}{|\omega_q - \omega_{q\pm 1}|}$$

Appendix 13: Wake potentials

Acceleration cells

Radiation fields in the TW acceleration structure



$$W_z(\vec{r}, \vec{r}', s) = -\frac{1}{q} \int_{z_1}^{z_2} dz [E_z(\vec{r}, z, t)]_{t=(z+s)/c}$$

$$\vec{W}_\perp(\vec{r}, \vec{r}', s) = \frac{1}{q} \int_{z_1}^{z_2} dz [\vec{E}_\perp + c(\hat{z} \times \vec{B})]_{t=(z+s)/c}$$

A. Novokhatski

Blue – deceleration
Green - acceleration

bunch

$$W_z=0, W_\perp=0 \text{ for } s < 0$$



Appendix 13: Wake potentials

Loss and kick distribution along the bunch $V_z(s)$ and $V_{\perp}(s)$:

$$V_z(s) = \int_0^{\infty} ds' \lambda(s-s') W_z(s') = \int_{-\infty}^s ds' \lambda(s') W_z(s-s'),$$

$$\vec{V}_{\perp}(s) = \int_0^{\infty} ds' \lambda(s-s') \vec{W}_{\perp}(s') = \int_{-\infty}^s ds' \lambda(s') \vec{W}_{\perp}(s-s').$$

$\lambda(s)$ is the charge distribution along the bunch.

Total losses and kick:

$$\Delta U = \int_{-\infty}^{\infty} ds \lambda(s) V_z(s), \quad k_{\ell} = \frac{\Delta U}{q^2} \equiv \frac{1}{q^2} \int_{-\infty}^{\infty} ds \lambda(s) V_z(s) \quad k_{\ell} - \text{loss factor}$$

$$k_{HOM} = k_{\ell} - \frac{1}{4} R/Q \cdot \omega |_{acc. mode}$$

$$\vec{p}_{\perp} = q^2 \vec{k}_{\perp} / c \quad \vec{k}_{\perp} \equiv \frac{1}{q^2} \int_{-\infty}^{\infty} ds \lambda(s) \vec{V}_{\perp}(s) \quad \vec{k}_{\perp} - \text{kick factor}$$

Appendix 13: Wake potentials

Panofsky-Wenzel theorem for wakes:

$$\frac{\partial \vec{W}_\perp}{\partial s} = \frac{c}{eq} \frac{\partial \vec{p}_\perp}{\partial s} = -\frac{1}{q} \int_0^L dz \left[\vec{\nabla}_\perp E_z(z, t) \right]_{t=(s+z)/c} = \vec{\nabla}_\perp W_z .$$

Relation between wake and impedance:

$$Z(\omega) = \int_0^\infty W_z(\tau) \exp\{-i\omega\tau\} d\tau \equiv \widetilde{W}_z(\tau) ,$$

$$s = c\tau = ct - z$$

$$Z(k) = \int_0^\infty W_z(s) \exp\{-iks\} ds \equiv \widetilde{W}_z(s) = c [Z(\omega)]_{\omega=kc}$$

and

$$k_\ell = \frac{\Delta U}{q^2} = \frac{1}{\pi q^2} \int_0^\infty Z_R(\omega) I^2(\omega) d\omega .$$